

Data Sheet: Technical Data

CCFC2012BC Microcontroller Data Sheet(V3.0)

Features

- Single issue, 32-bit CPU core complex (c2003)
 - Compliant with the Power Architecture® technology embedded category
 - Enhanced instruction set allowing variable length encoding (VLE) for code size footprint reduction. With the optional encoding of mixed 16-bit and 32-bit instructions, it is possible to achieve significant code size footprint reduction.
- Up to 1.5 MB on-chip code flash memory supported with the flash memory controller
- 128 (4 × 16+64) KB on-chip data flash memory with ECC
- Up to 128 KB on-chip SRAM
- Memory protection unit (MPU) with 8 region descriptors and 32-byte region granularity on certain family members (Refer to [Table 1](#) for details.)
- Interrupt controller (INTC) capable of handling 231 selectable-priority interrupt sources
- Frequency modulated phase-locked loop (FMPLL)
- Crossbar switch architecture for concurrent access to peripherals, Flash, or RAM from multiple bus masters
- 16-channel eDMA controller with multiple transfer request sources using DMA multiplexer
- Boot assist module (BAM) supports internal Flash programming via a serial link (CAN or SCI)
- Timer supports I/O channels providing a range of 16-bit input capture, output compare, and pulse width modulation functions (eMIOS)
- 2 analog-to-digital converters (ADC): one 10-bit and one 12-bit
- Cross Trigger Unit to enable synchronization of ADC conversions with a timer event from the eMIOS or PIT
- Up to 6 serial peripheral interface (DSPI) modules
- Up to 10 serial communication interface (LINFlex) modules
- Up to 8 enhanced full CAN (FlexCAN) modules with configurable buffers, each can interface compatible with canfd interface.
- 1 inter-integrated circuit (I²C) interface module
- Up to 149 configurable general purpose pins supporting input and output operations (package dependent)
- Real-Time Counter (RTC)
- Clock source from internal 128 kHz or 16 MHz oscillator supporting autonomous wakeup with 1 ms resolution with maximum timeout of 2 seconds
- Optional support for RTC with clock source from external 32 kHz crystal oscillator, supporting wakeup with 1 sec resolution and maximum timeout of 1 hour
- Up to 8 periodic interrupt timers (PIT) with 32-bit counter resolution
- Nexus development interface (NDI) per IEEE-ISTO 5001-2003 Class Two Plus
- Device/board boundary scan testing supported per Joint Test Action Group (JTAG) of IEEE (IEEE 1149.1)
- On-chip voltage regulator (VREG) for regulation of input supply for all internal levels

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1 Introduction

This document describes the features of the family and options available within the family members, and highlights important electrical and physical characteristics of the device.

1.1 Description

This family of 32-bit system-on-chip (SoC) microcontrollers is the latest achievement in integrated automotive application controllers. It belongs to an expanding family of automotive-focused products designed to address the next wave of body electronics applications within the vehicle.

The advanced and cost-efficient c2003 host processor core of this automotive controller family complies with the Power Architecture technology and only implements the VLE (variable-length encoding) APU (Auxiliary Processor Unit), providing improved code density. It operates at speeds of up to 120 MHz and offers high performance processing optimized for low power consumption. It capitalizes on the available development infrastructure of current Power Architecture devices and is supported with software drivers, operating systems and configuration code to assist with users implementations.

Table 1-1 CCFC2012BC family comparison

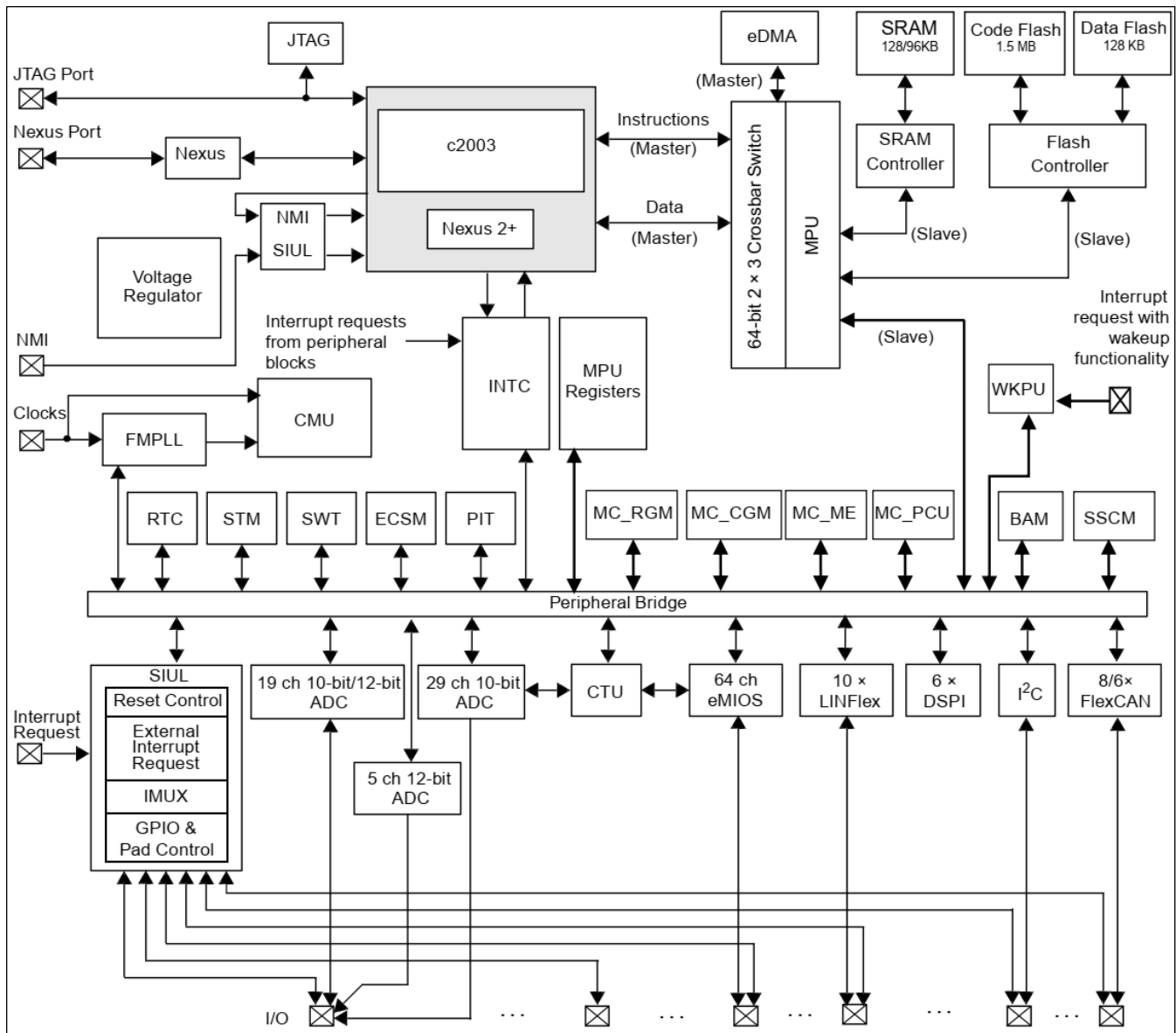
Function	CCFC2012BC 60L7	CCFC2012BC 60L5	CCFC2012BC 60L1	CCFC2012BC 50L5	CCFC2012BC 50L3	CCFC2012BC 50L1
CPU	C2003 (compatible with e200z3)	C2003 (compatible with e200z3)	C2003 (compatible with e200z3)	C2003 (compatible with e200z3)	C2003 (compatible with e200z3)	C2003 (compatible with e200z3)
Frequency	120MHz	120MHz	120MHz	120MHz	120MHz	120MHz
CFLASH	1.5MB	1.5MB	1MB	1MB	1MB	1MB
DFLASH	128KB	128KB	128KB	128KB	128KB	128KB
SRAM	128KB	128KB	128KB	96KB	128KB/96KB	96KB
PLL	FMPLL	FMPLL	FMPLL	FMPLL	FMPLL	FMPLL
eDMA	16	16	16	16	16	16
FlexCAN (CanFD)	8	6	8	6	6	3
DSPI	6	5	3	5	3	2
LINFlex	10	8	6	8	6	6
eMIOS	64	64	19	64	37	19
ADC	10bit+12bit/53ch	10bit+12bit/39ch	10bit+12bit/16ch	10bit+12bit/39ch	10bit+12bit/33ch	10bit+12bit/16ch
PIT	8	8	8	8	8	8
STM/SWT	5	5	5	5	5	5
I2C	1	1	1	1	1	1
MPU	8	8	8	8	8	8
GPIO	149	121	45	123	79	45
Package	LQFP176	LQFP144	LQFP64	LQFP144	LQFP100	LQFP64
Pin	SPC560B64L7C (B)6E0X	SPC560B60L5C 6E0	SPC560B50L5*	CCFC2002BC/ SPC560B50L5C 6E0	SPC560B50L3*	SPC560B50L1*
Replaceability	SPC560*L7*/ MPC5607B*/SP C5744B(无以太 网)	SPC560B60L5C 6E0 /MPC5606B*	SPC560B50L5* /MPC5604B*	CCFC2002BC/ SPC560*L5*	SPC560*L3*	SPC560*L1*
Reliability	AEC-Q100	AEC-Q100	AEC-Q100	AEC-Q100	AEC-Q100	AEC-Q100

1. If the ADC module is used and the frequency of system is greater than 80MHz, peripheral set 3 frequency should be configured by software to lower the frequency of FADC below 16MHz in **CCFC2012BC50XX**.

2. The size of SRAM in **CCFC2012BC50L3** can be configured by software.

2 Block Diagram

Figure 1 shows a top-level block diagram of the CCFC2012BC.



Abbreviation

ADC	Analog-to-Digital Converter	MC_CGM	Clock Generation Module
BAM	Boot Assist Module	MC_ME	Mode Entry Module
CMU	Clock Monitor Unit	MC_PCU	Power Control Unit
CTU	Cross Triggering Unit	MC_RGM	Reset Generation Module
DSPI	Deserial Serial Peripheral Interface	MPU	Memory Protection Unit
ECSM	Error Correction Status Module	NMI	Non-Maskable Interrupt
eDMA	Enhanced Direct Memory Access	PIT	Periodic Interrupt Timer
eMIOS	Enhanced Modular Input Output System	RTC	Real-Time Clock
Flash	Flash memory	SIUL	System Integration Unit Lite
FlexCAN	Controller Area Network	SRAM	Static Random-Access Memory
FMPLL	Frequency-Modulated Phase-Locked Loop	SSCM	System Status Configuration Module
GPIO	General-purpose input/output	STM	System Timer Module
I ² C	Inter-Integrated Circuit bus	SWT	Software Watchdog Timer
IMUX	Internal Multiplexer	VREG	Voltage regulator
INTX	Interrupt Controller	WKPU	Wakeup Unit
JTAG	JTAG controller	XBAR	Crossbar switch
LINFlex	Serial Communication Interface (LIN support)		

Figure 2-1. CCFC2012BC block diagram

Table 2-1 summarizes the functions of the blocks present on the CCFC2012BC.

Table 2-1. CCFC2012BC series block summary

Block	Function
Analog-to-digital converter (ADC)	Converts analog voltages to digital values
Boot assist module (BAM)	A block of read-only memory containing VLE code which is executed according to the boot mode of the device
Clock generation module (MC_CGM)	Provides logic and control required for the generation of system and peripheral clocks
Clock monitor unit (CMU)	Monitors clock source (internal and external) integrity
Cross triggering unit (CTU)	Enables synchronization of ADC conversions with a timer event from the eMIOS or from the PIT
Crossbar switch (XBAR)	Supports simultaneous connections between two master ports and three slave ports. The crossbar supports a 32-bit address bus width and a 64-bit data bus width.
Deserial serial peripheral interface (DSPI)	Provides a synchronous serial interface for communication with external devices
Enhanced direct memory access (eDMA)	Performs complex data transfers with minimal intervention from a host processor via “n” programmable channels
Enhanced modular input output system (eMIOS)	Provides the functionality to generate or measure events
Error correction status module (ECSM)	Provides a myriad of miscellaneous control functions for the device including program-visible information about configuration and revision levels, a reset status register, wakeup control for exiting sleep modes, and optional features such as information on memory errors reported by error-correcting codes
Flash memory	Provides non-volatile storage for program code, constants and variables
FlexCAN (controller area network)	Supports the standard CAN communications protocol
Frequency-modulated phase-locked loop (FMPLL)	Generates high-speed system clocks and supports programmable frequency modulation
Inter-integrated circuit (I ² C) bus	Two-wire bidirectional serial bus that provides a simple and efficient method of data exchange between devices
Internal multiplexer (IMUX) SIU subblock	Allows flexible mapping of peripheral interface on the different pins of the device
Interrupt controller (INTC)	Provides priority-based preemptive scheduling of interrupt requests
JTAG controller (JTAGC)	Provides the means to test chip functionality and connectivity while remaining transparent to system logic when not in test mode
LINFlex controller	Manages a high number of LIN (Local Interconnect Network protocol) messages efficiently with a minimum of CPU load
Memory protection unit (MPU)	Provides hardware access control for all memory references generated in a device
Mode entry module (MC_ME)	Provides a mechanism for controlling the device operational mode and mode transition sequences in all functional states; also manages the power control unit, reset generation module and clock generation module, and holds the configuration, control and status registers accessible for applications
Non-maskable interrupt (NMI)	Handles external events that must produce an immediate response, such as power down detection
Periodic interrupt timer (PIT)	Produces periodic interrupts and triggers
Power control unit (MC_PCU)	Reduces the overall power consumption by disconnecting parts of the device from the power supply via a power switching device; device components are grouped into sections called “power domains” which are controlled by the PCU
Real-time counter (RTC)	A free running counter used for time keeping applications, the RTC can be configured to generate an interrupt at a predefined interval independent of the mode of operation (run mode or low-power mode)
Reset generation module (MC_RGM)	Centralizes reset sources and manages the device reset sequence of the device
Static random-access memory (SRAM)	Provides storage for program code, constants, and variables
System integration unit lite (SIUL)	Provides control over all the electrical pad controls and up 32 ports with 16 bits of

Block	Function
	bidirectional, general-purpose input and output signals and supports up to 32 external interrupts with trigger event configuration
System status and configuration module (SSCM)	Provides system configuration and status data (such as memory size and status, device mode and security status), device identification data, debug status port enable and selection, and bus and peripheral abort enable/disable
System timer module (STM)	Provides a set of output compare events to support AUTOSAR (Automotive Open System Architecture) and operating system tasks
Software watchdog timer (SWT)	Provides protection from runaway code
Wakeup unit (WKPU)	The wakeup unit supports up to 27 external sources that can generate interrupts or wakeup events, of which 1 can cause non-maskable interrupt requests or wakeup events.

3 Package pinouts and signal descriptions

3.1 Package pinouts

The available LQFP pinouts and the ballmap are provided in the following figures. For pin signal descriptions, please see Table 3-3. Functional port pin descriptions.

Figure 3-1 shows the CCFC2012BC60L7 in the 176 LQFP package.

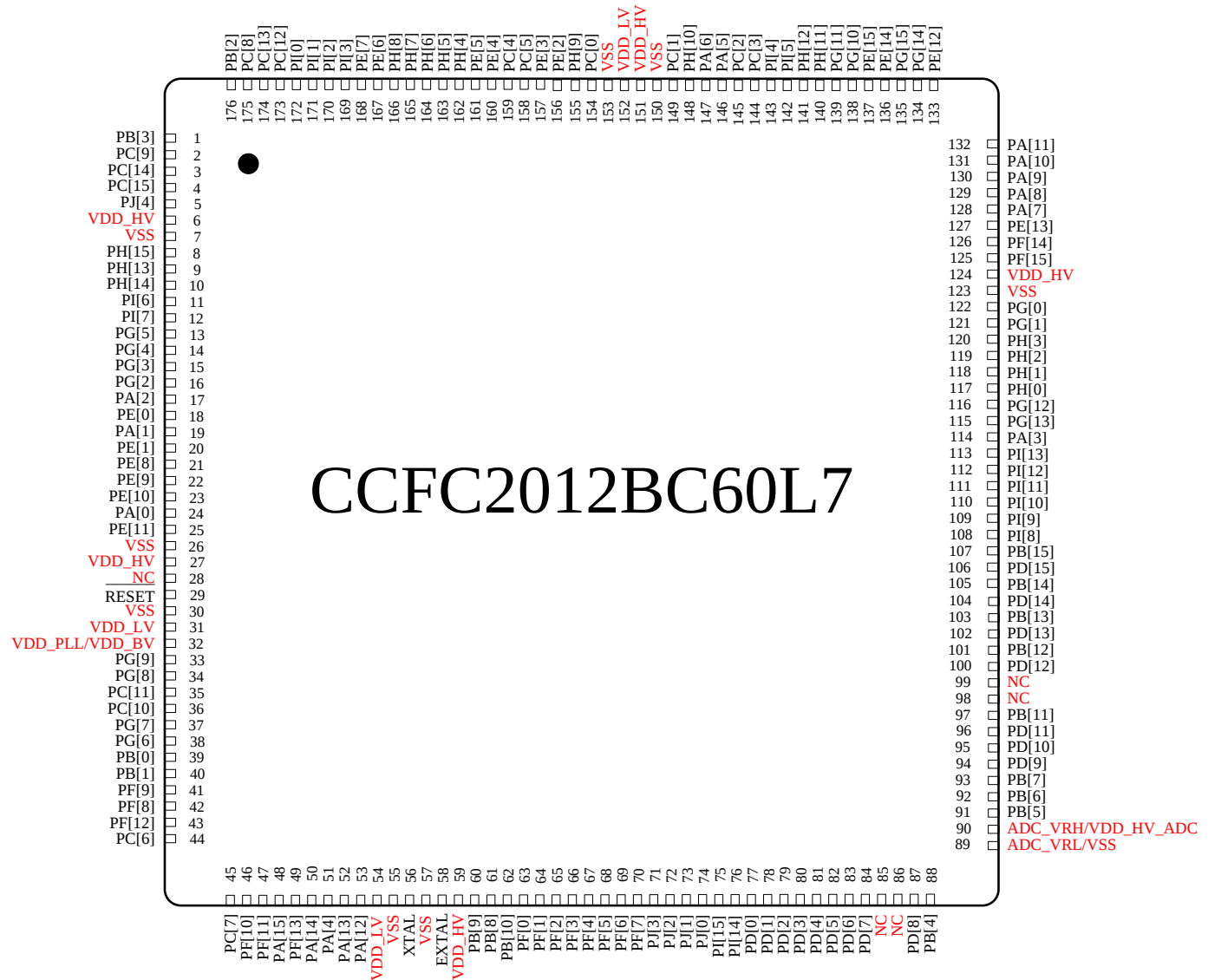


Figure 3-1. CCFC2012BC60L7 176 LQFP pin configuration

Figure 3-2. CCFC2012BC60L5 144 LQFP pin configuration

Figure 3-3 shows the CCFC2012BC60L1 in the 64 LQFP package.

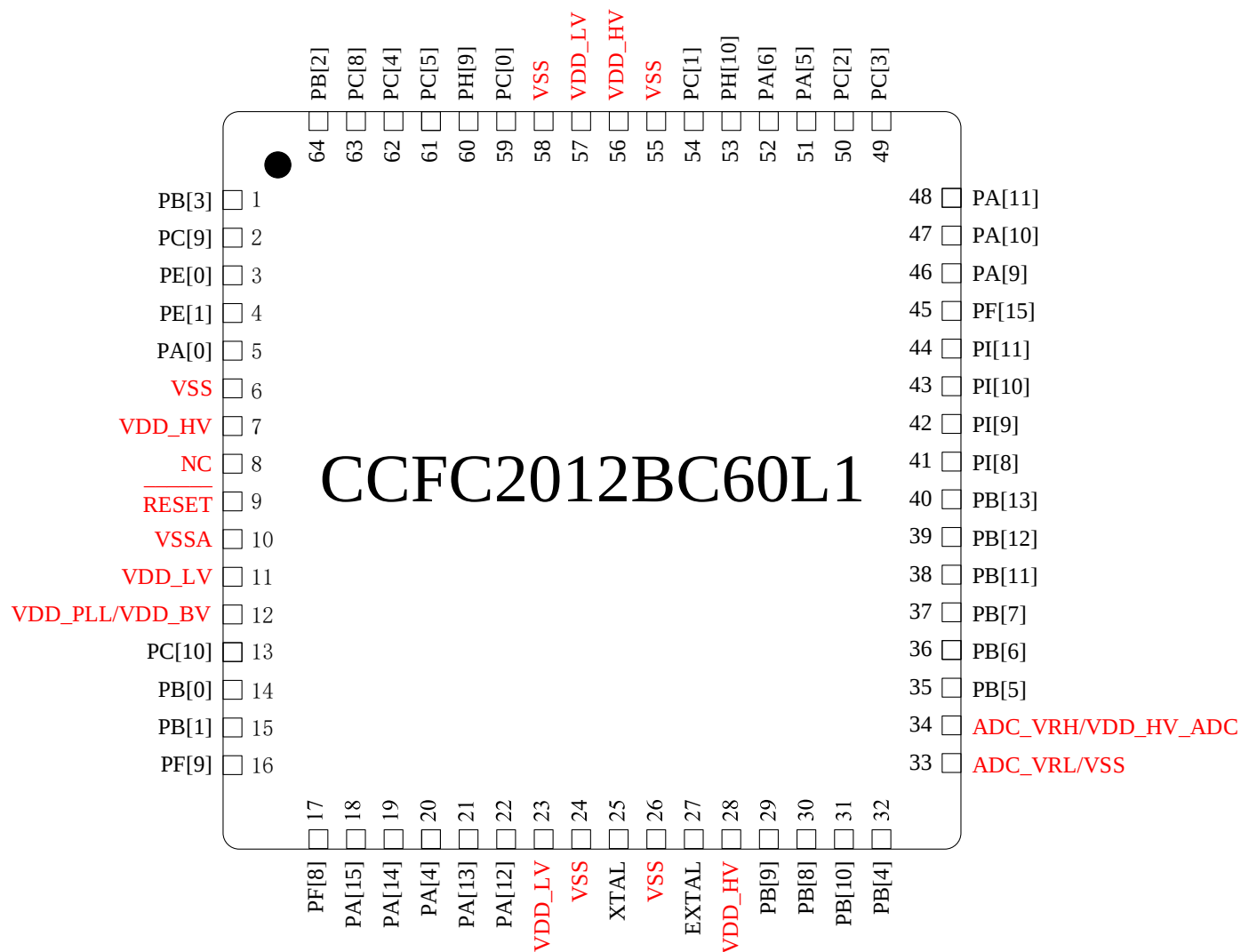


Figure 3-3. CCFC2012BC60L1 64 LQFP pin configuration

Figure 3-4 shows the CCFC2012BC50L5 in the 144 LQFP package.

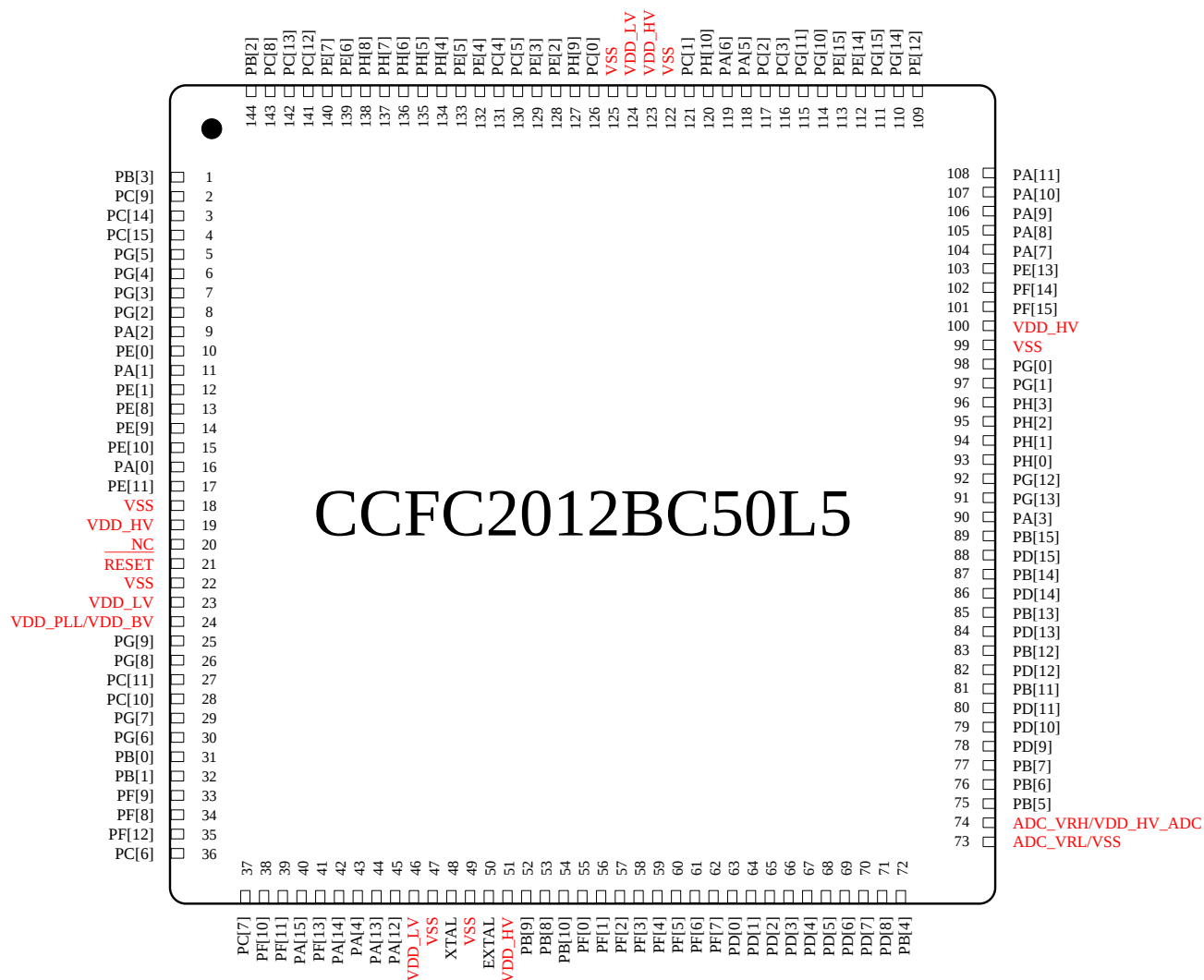


Figure 3-4. CCFC2012BC50L5 144 LQFP pin configuration

Figure 3-5 shows the CCFC2012BC50L3 in the 100 LQFP package.

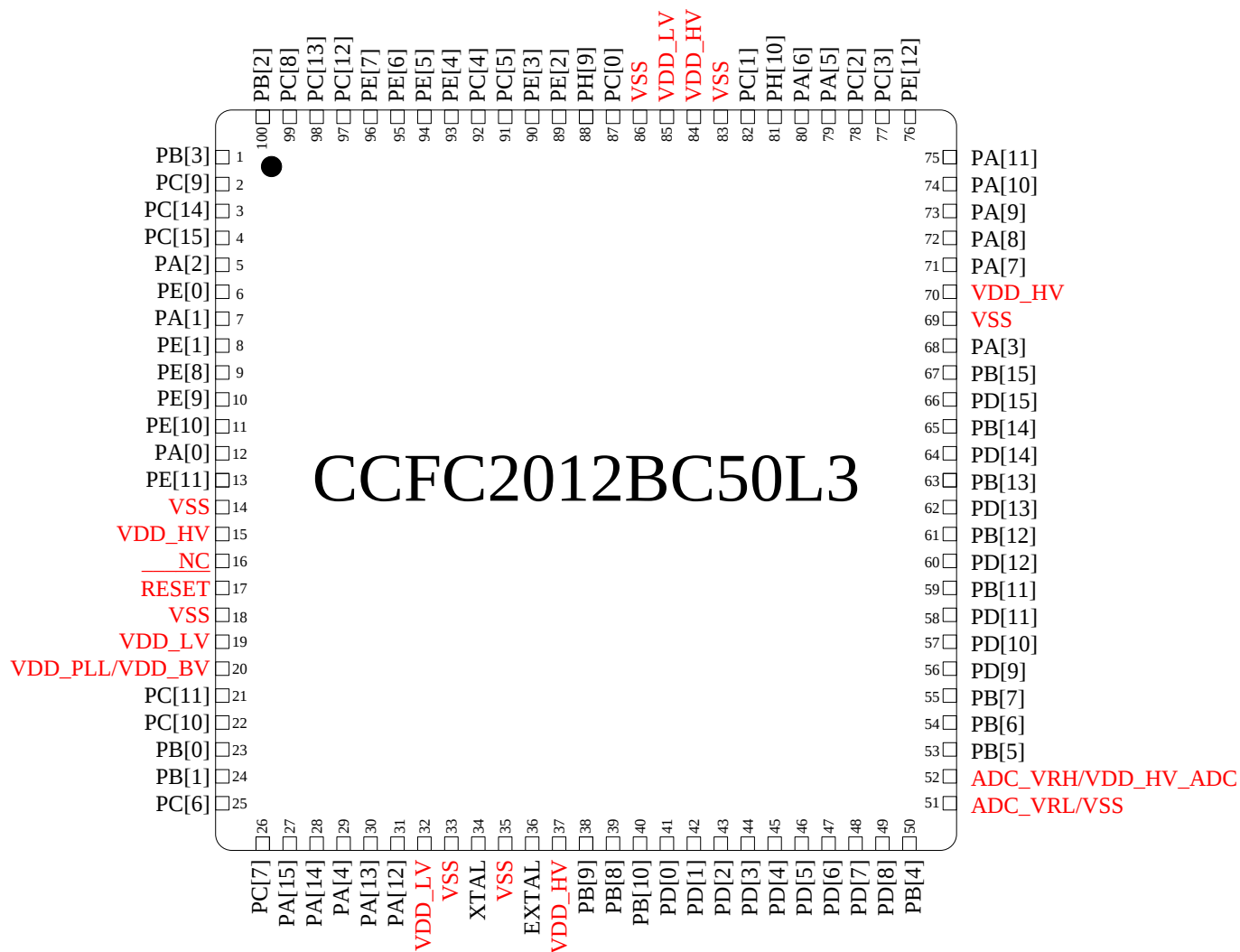


Figure 3-5. CCFC2012BC50L3 100 LQFP pin configuration

Figure 3-6 shows the CCFC2012BC50L1 in the 64 LQFP package.

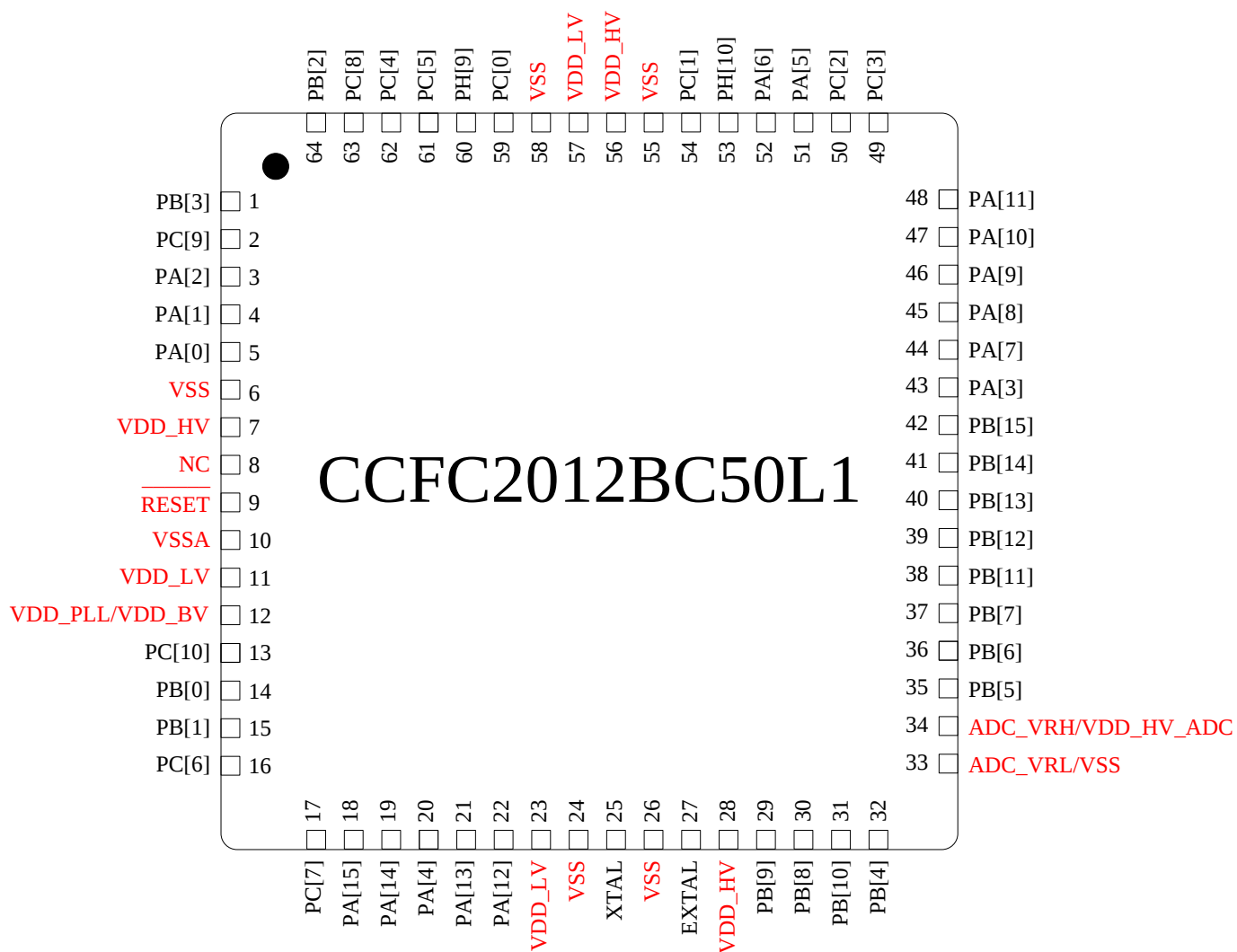


Figure 3-6. CCFC2012BC50L1 64 LQFP pin configuration

3.2 Pad configuration during reset phases

All pads have a fixed configuration under reset.

During the power-up phase, all pads are forced to tristate.

After power-up phase, all pads are tristate with the following exceptions:

- PA[9] (FAB) is pull-down. Without external strong pull-up the device starts fetching from flash.
- PA[8], PC[0] and PH[9:10] are in input weak pull-up when out of reset.
- RESET pad is driven low by the device till 40 FIRC clock cycles after phase2 completion. Minimum phase3 duration is 40 FIRC cycles.
- Nexus output pads (MDO[n], MCKO, EVTO, MSEO) are forced to output.

3.3 Pad configuration during standby mode exit

Pad configuration (input buffer enable, pull enable) for low-power wakeup pads is controlled by both the SIUL and WKPU modules. During standby exit, all low power pads PA[0,1,2,4,15], PB[1,3,8,9,10], PC[7,9,11], PD[0,1], PE[0,9,11], PF[9,11,13], PG[3,5,7,9], PI[1,3] are configured according to their respective configuration done in the WKPU module. All other pads will have the same configuration as expected after a reset.

The TDO pad has been moved into the STANDBY domain in order to allow low-power debug handshaking in STANDBY mode. However, no pull-resistor is active on the TDO pad while in STANDBY mode. At this time the pad is configured as an input. When no debugger is connected the TDO pad is floating causing additional current consumption.

To avoid the extra consumption TDO must be connected. An external pull-up resistor in the range of 47–100 kOhms should be added between the TDO pin and VDD. Only if the TDO pin is used as an application pin and a pull-up cannot be used should a pull-down resistor with the same value be used instead between the TDO pin and GND.

3.4 Voltage supply pins

Voltage supply pins are used to provide power to the device. Three dedicated pins are used for 1.2V regulator stabilization.

Table 3-1. Voltage supply pin descriptions

Port pin	Function	CCFC2012BC Pin number					
		60L7 LQFP176	60L5 LQFP144	60L1 LQFP64	50L5 LQFP144	50L3 LQFP 100	50L1 LQFP64
VDD_HV	Digital supply voltage	6, 27, 59, 90, 124,151	19, 51, 74, 100, 123	7, 28, 34, 56	19, 51, 74, 100, 123	5, 37, 52, 70, 84	7, 28, 34, 56
VSS	Digital ground	7, 26,30, 55, 57, 89, 123,150,153	18, 22,47, 49,73,99, 122, 125	6, 10, 24, 26, 33, 55, 58	18, 22, 47, 49, 73, 99, 122, 125	14, 18, 33, 35, 51, 69, 83, 86	6, 10, 24, 26, 33, 55, 58
VDD_LV	1.2V decoupling pins. Decoupling capacitor must be connected between these pins and the nearest V _{SS_LV} pin.	31, 54, 152	23, 46, 124	11, 23, 57	23, 46, 124	19, 32, 85	11, 23, 57
VDD_BV	Internal regulator supply voltage	32	24	12	24	20	12
ADC_VRL	Reference ground and analog ground for the A/D converter	89	73	33	73	51	33
ADC_VRH	Reference voltage and analog supply for the A/D converter	90	74	34	74	52	34

3.5 Pad types

In the device the following types of pads are available for system pins and functional port pins:

S = Slow¹

M = Medium^{1 2}

F = Fast^{1 2}

I = Input only with analog feature¹

J = Input/Output ('S' pad) with analog feature

X = Oscillator

3.6 System pins

The system pins are listed in Table 3-2.

Table 3-2. System pin descriptions

Port pin	Function	I/O Direction	Pad type	RESET config.	Pin number					
					60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
RESET	Bidirectional reset with Schmitt-Trigger characteristics and noise filter.	I/O	M	Input weak pull-up after RGM PHASE2 and 40 FIRC cycles	29	21	9	21	17	9
EXTAL	Analog input of the oscillator amplifier circuit. Needs to be grounded if oscillator bypass mode is used.	I	X	Tristate	58	50	27	50	36	27
XTAL	Analog output of the oscillator amplifier circuit, when the oscillator is not in bypass mode. Analog input for the clock generator when the oscillator is in bypass mode.	I/O	X	Tristate	56	48	25	48	34	25

3.7 Functional port pins

The functional port pins are listed in Table 3-3.

Table 3-3. Functional port pin descriptions

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port A													
PA[0]	PCR[0]	AF0	GPIO[0]	SIUL	I/O	M	Tristate	24	16	5	16	12	5
		AF1	E0UC[0]	eMIOS_0	I/O								
		AF2	CLKOUT	MC_CGM	O								
		AF3	E0UC[13]	eMIOS_0	I/O								
		—	WKPU[19] ⁵	WKPU	I								
PA[1]	PCR[1]	AF0	GPIO[1]	SIUL	I/O	S	Tristate	19	11	—	11	7	4
		AF1	E0UC[1]	eMIOS_0	I/O								
		AF2	NMI ⁶	WKPU	I								
		AF3	—	—	—								
		—	WKPU[2] ⁵	WKPU	I								
PA[2]	PCR[2]	AF0	GPIO[2]	SIUL	I/O	S	Tristate	17	9	—	9	5	3
		AF1	E0UC[2]	eMIOS_0	I/O								
		AF2	—	—	—								
		AF3	MA[2]	ADC_0	O								
		—	WKPU[3] ⁵	WKPU	I								
PA[3]	PCR[3]	AF0	GPIO[3]	SIUL	I/O	J	Tristate	114	90	—	90	68	43
		AF1	E0UC[3]	eMIOS_0	I/O								
		AF2	LIN5TX	LINFlex_5	O								
		AF3	CS4_1	DSPI_1	O								
		—	EIRQ[0]	SIUL	I								
—	ADC1_S[0]	ADC_1	I										
PA[4]	PCR[4]	AF0	GPIO[4]	SIUL	I/O	S	Tristate	51	43	20	43	29	20
		AF1	E0UC[4]	eMIOS_0	I/O								
		AF2	—	—	—								
		AF3	CS0_1	DSPI_1	I/O								
		—	LIN5RX	LINFlex_5	I								
—	WKPU[9] ⁵	WKPU	I										
PA[5]	PCR[5]	AF0	GPIO[5]	SIUL	I/O	M	Tristate	146	118	51	118	79	51
		AF1	E0UC[5]	eMIOS_0	I/O								
		AF2	LIN4TX	LINFlex_4	O								
		AF3	—	—	—								
		—	—	—	—								
PA[6]	PCR[6]	AF0	GPIO[6]	SIUL	I/O	S	Tristate	147	119	52	119	80	52
		AF1	E0UC[6]	eMIOS_0	I/O								
		AF2	—	—	—								
		AF3	CS1_1	DSPI_1	O								
		—	EIRQ[1]	SIUL	I								
—	LIN4RX	LINFlex_4	I										
PA[7]	PCR[7]	AF0	GPIO[7]	SIUL	I/O	J	Tristate	128	104	—	104	71	44
		AF1	E0UC[7]	eMIOS_0	I/O								
		AF2	LIN3TX	LINFlex_3	O								
		AF3	—	—	—								
		—	EIRQ[2]	SIUL	I								
—	ADC1_S[1]	ADC_1	I										

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PA[8]	PCR[8]	AF0 AF1 AF2 AF3 — N/A ⁷ —	GPIO[8] E0UC[8] E0UC[14] — EIRQ[3] ABS[0] LIN3RX	SIUL eMIOS_0 eMIOS_0 — SIUL BAM LINFlex_3	I/O I/O I/O — I I I	S	Input, weak pull-up	129	105	46	105	72	45
PA[9]	PCR[9]	AF0 AF1 AF2 AF3 N/A ⁷ —	GPIO[9] E0UC[9] — CS2_1 FAB	SIUL eMIOS_0 — DSPI_1 BAM	I/O I/O — O I	S	Pull- down	130	106	47	106	73	46
PA[10]	PCR[10]	AF0 AF1 AF2 AF3 —	GPIO[10] E0UC[10] SDA LIN2TX ADC1_S[2]	SIUL eMIOS_0 I ² C_0 LINFlex_2 ADC_1	I/O I/O I/O O I	J	Tristate	131	107	48	107	74	47
PA[11]	PCR[11]	AF0 AF1 AF2 AF3 — — —	GPIO[11] E0UC[11] SCL — EIRQ[16] LIN2RX ADC1_S[3]	SIUL eMIOS_0 I ² C_0 — SIUL LINFlex_2 ADC_1	I/O I/O I/O — I I I	J	Tristate	132	108	—	108	75	48
PA[12]	PCR[12]	AF0 AF1 AF2 AF3 — —	GPIO[12] — E0UC[28] CS3_1 EIRQ[17] SIN_0 —	SIUL — eMIOS_0 DSPI_1 SIUL DSPI_0	I/O — I/O O I I	S	Tristate	53	45	22	45	31	22
PA[13]	PCR[13]	AF0 AF1 AF2 AF3	GPIO[13] SOUT_0 E0UC[29] —	SIUL DSPI_0 eMIOS_0 —	I/O O I/O —	M	Tristate	52	44	21	44	30	21
PA[14]	PCR[14]	AF0 AF1 AF2 AF3 —	GPIO[14] SCK_0 CS0_0 E0UC[0] EIRQ[4]	SIUL DSPI_0 DSPI_0 eMIOS_0 SIUL	I/O I/O I/O I/O I	M	Tristate	50	42	19	42	28	19
PA[15]	PCR[15]	AF0 AF1 AF2 AF3 —	GPIO[15] CS0_0 SCK_0 E0UC[1] WKPU[10] ⁵	SIUL DSPI_0 DSPI_0 eMIOS_0 WKPU	I/O I/O I/O I/O I	M	Tristate	48	40	18	40	27	18

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port B													
PB[0]	PCR[16]	AF0 AF1 AF2 AF3	GPIO[16] CAN0TX E0UC[30] LIN0TX	SIUL FlexCAN_0 eMIOS_0 LINFlex_0	I/O O I/O O	M	Tristate	39	31	14	31	23	14
PB[1]	PCR[17]	AF0 AF1 AF2 AF3 — — —	GPIO[17] — E0UC[31] — WKPU[4] ⁵ CAN0RX LIN0RX	SIUL — eMIOS_0 — WKPU FlexCAN_0 LINFlex_0	I/O — I/O — I I I	S	Tristate	40	32	15	32	24	15
PB[2]	PCR[18]	AF0 AF1 AF2 AF3	GPIO[18] LIN0TX SDA E0UC[30]	SIUL LINFlex_0 I ² C_0 eMIOS_0	I/O O I/O I/O	M	Tristate	176	144	64	144	100	64
PB[3]	PCR[19]	AF0 AF1 AF2 AF3 — —	GPIO[19] E0UC[31] SCL — WKPU[11] ⁵ LIN0RX	SIUL eMIOS_0 I ² C_0 — WKPU LINFlex_0	I/O I/O I/O — I I	S	Tristate	1	1	1	1	1	1
PB[4]	PCR[20]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[0] ADC1_P[0] GPIO[20]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	88	72	32	72	50	32
PB[5]	PCR[21]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[1] ADC1_P[1] GPIO[21]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	91	75	35	75	53	35
PB[6]	PCR[22]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[2] ADC1_P[2] GPIO[22]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	92	76	36	76	54	36
PB[7]	PCR[23]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[3] ADC1_P[3] GPIO[23]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	93	77	37	77	55	37

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PB[8]	PCR[24]	AF0 AF1 AF2 AF3 — — — —	GPIO[24] — — — OSC32K_XTAL ⁸ WKPU[25] ⁵ ADC0_S[0] ADC1_S[4]	SIUL — — — OSC32K WKPU ADC_0 ADC_1	I — — — — I ⁹ I I	I	—	61	53	30	53	39	30
PB[9]	PCR[25]	AF0 AF1 AF2 AF3 — — — —	GPIO[25] — — — OSC32K_EXT ⁸ WKPU[26] ⁵ ADC0_S[1] ADC1_S[5]	SIUL — — — OSC32K WKPU ADC_0 ADC_1	I — — — — I ⁹ I I	I	—	60	52	29	52	38	29
PB[10]	PCR[26]	AF0 AF1 AF2 AF3 — — — —	GPIO[26] — — — WKPU[8] ⁵ ADC0_S[2] ADC1_S[6]	SIUL — — — WKPU ADC_0 ADC_1	I/O — — — I I I	J	Tristate	62	54	31	54	40	31
PB[11]	PCR[27]	AF0 AF1 AF2 AF3 — — — —	GPIO[27] E0UC[3] — CS0_0 ADC0_S[3]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — I/O I	J	Tristate	97	—	38	—	59	38
PB[12]	PCR[28]	AF0 AF1 AF2 AF3 — — — —	GPIO[28] E0UC[4] — CS1_0 ADC0_X[0]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	101	83	39	83	61	39
PB[13]	PCR[29]	AF0 AF1 AF2 AF3 — — — —	GPIO[29] E0UC[5] — CS2_0 ADC0_X[1]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	103	85	40	85	63	40
PB[14]	PCR[30]	AF0 AF1 AF2 AF3 — — — —	GPIO[30] E0UC[6] — CS3_0 ADC0_X[2]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	105	87	—	87	65	41
PB[15]	PCR[31]	AF0 AF1 AF2 AF3 — — — —	GPIO[31] E0UC[7] — CS4_0 ADC0_X[3]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	107	89	—	89	67	42

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port C													
PC[0] ¹⁰	PCR[32]	AF0 AF1 AF2 AF3	GPIO[32] — TDI —	SIUL — JTAGC —	I/O — I —	M	Input, weak pull-up	154	126	59	126	87	59
PC[1] ¹⁰	PCR[33]	AF0 AF1 AF2 AF3	GPIO[33] — TDO —	SIUL — JTAGC —	I/O — O —	F ¹¹	Tristate	149	121	54	121	82	54
PC[2]	PCR[34]	AF0 AF1 AF2 AF3 —	GPIO[34] SCK_1 DSPI_1 CAN4TX DEBUG[0] EIRQ[5] SIUL	SIUL DSPI_1 FlexCAN_4 SSCM O I	I/O I/O O O I	M	Tristate	145	117	50	117	78	50
PC[3]	PCR[35]	AF0 AF1 AF2 AF3 — — —	GPIO[35] CS0_1 MA[0] DEBUG[1] EIRQ[6] SIUL CAN1RX CAN4RX	SIUL DSPI_1 ADC_0 SSCM O I I I	I/O I/O O O I I I	S	Tristate	144	116	49	116	77	49
PC[4]	PCR[36]	AF0 AF1 AF2 AF3 — — —	GPIO[36] E1UC[31] — DEBUG[2] EIRQ[18] SIN_1 CAN3RX	SIUL eMIOS_1 — SSCM SIUL DSPI_1 FlexCAN_3	I/O I/O — O I I I	M	Tristate	159	131	62	131	92	62
PC[5]	PCR[37]	AF0 AF1 AF2 AF3 —	GPIO[37] SOUT_1 CAN3TX DEBUG[3] EIRQ[7]	SIUL DSPI_1 FlexCAN_3 SSCM SIUL	I/O O O O I	M	Tristate	158	130	61	130	91	61
PC[6]	PCR[38]	AF0 AF1 AF2 AF3	GPIO[38] LIN1TX E1UC[28] DEBUG[4]	SIUL LINFlex_1 eMIOS_1 SSCM	I/O O I/O O	S	Tristate	44	36	—	36	25	16
PC[7]	PCR[39]	AF0 AF1 AF2 AF3 — —	GPIO[39] — E1UC[29] DEBUG[5] LIN1RX WKPU[12] ⁵	SIUL — eMIOS_1 SSCM LINFlex_1 WKPU	I/O — I/O O I I	S	Tristate	45	37	—	37	26	17
PC[8]	PCR[40]	AF0 AF1 AF2 AF3	GPIO[40] LIN2TX E0UC[3] DEBUG[6]	SIUL LINFlex_2 eMIOS_0 SSCM	I/O O I/O O	S	Tristate	175	143	63	143	99	63

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PC[9]	PCR[41]	AF0 AF1 AF2 AF3 — —	GPIO[41] — E0UC[7] DEBUG[7] WKPU[13] ⁵ LIN2RX	SIUL — eMIOS_0 SSCM WKPU LINFlex_2	I/O — I/O O I I	S	Tristate	2	2	2	2	2	2
PC[10]	PCR[42]	AF0 AF1 AF2 AF3	GPIO[42] CAN1TX CAN4TX MA[1]	SIUL FlexCAN_1 FlexCAN_4 ADC_0	I/O O O O	M	Tristate	36	28	13	28	22	13
PC[11]	PCR[43]	AF0 AF1 AF2 AF3 — — —	GPIO[43] — — MA[2] WKPU[5] ⁵ CAN1RX CAN4RX	SIUL — — ADC_0 WKPU FlexCAN_1 FlexCAN_4	I/O — — O I I I	S	Tristate	35	27	—	27	21	—
PC[12]	PCR[44]	AF0 AF1 AF2 AF3 — —	GPIO[44] E0UC[12] — — EIRQ[19] SIN_2	SIUL eMIOS_0 — — SIUL DSPI_2	I/O I/O — — I I	M	Tristate	173	141	—	141	97	—
PC[13]	PCR[45]	AF0 AF1 AF2 AF3	GPIO[45] E0UC[13] SOUT_2 —	SIUL eMIOS_0 DSPI_2 —	I/O I/O O —	S	Tristate	174	142	—	142	98	—
PC[14]	PCR[46]	AF0 AF1 AF2 AF3 —	GPIO[46] E0UC[14] SCK_2 — EIRQ[8]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	S	Tristate	3	3	—	3	3	—
PC[15]	PCR[47]	AF0 AF1 AF2 AF3 —	GPIO[47] E0UC[15] CS0_2 — EIRQ[20]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	M	Tristate	4	4	—	4	4	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port D													
PD[0]	PCR[48]	AF0 AF1 AF2 AF3 — — —	GPIO[48] — — — WKPU[27] ⁵ ADC0_P[4] ADC1_P[4]	SIUL — — — WKPU ADC_0 ADC_1	I — — — I I I	I	Tristate	77	63	—	63	41	—
PD[1]	PCR[49]	AF0 AF1 AF2 AF3 — — —	GPIO[49] — — — WKPU[28] ⁵ ADC0_P[5] ADC1_P[5]	SIUL — — — WKPU ADC_0 ADC_1	I — — — I I I	I	Tristate	78	64	—	64	42	—
PD[2]	PCR[50]	AF0 AF1 AF2 AF3 — — —	GPIO[50] — — — ADC0_P[6] ADC1_P[6]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	79	65	—	65	43	—
PD[3]	PCR[51]	AF0 AF1 AF2 AF3 — — —	GPIO[51] — — — ADC0_P[7] ADC1_P[7]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	80	66	—	66	44	—
PD[4]	PCR[52]	AF0 AF1 AF2 AF3 — — —	GPIO[52] — — — ADC0_P[8] ADC1_P[8]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	81	67	—	67	45	—
PD[5]	PCR[53]	AF0 AF1 AF2 AF3 — — —	GPIO[53] — — — ADC0_P[9] ADC1_P[9]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	82	68	—	68	46	—
PD[6]	PCR[54]	AF0 AF1 AF2 AF3 — — —	GPIO[54] — — — ADC0_P[10] ADC1_P[10]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	83	69	—	69	47	—
PD[7]	PCR[55]	AF0 AF1 AF2 AF3 — — —	GPIO[55] — — — ADC0_P[11] ADC1_P[11]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	84	70	—	70	48	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PD[8]	PCR[56]	AF0	GPIO[56]	SIUL	I	I	Tristate	87	71	—	71	49	—
		AF1	—	—	—								
		AF2	—	—	—								
		AF3	—	—	—								
		—	ADC0_P[12]	ADC_0	I								
		—	ADC1_P[12]	ADC_1	I								
PD[9]	PCR[57]	AF0	GPIO[57]	SIUL	I	I	Tristate	94	78	—	78	56	—
		AF1	—	—	—								
		AF2	—	—	—								
		AF3	—	—	—								
		—	ADC0_P[13]	ADC_0	I								
		—	ADC1_P[13]	ADC_1	I								
PD[10]	PCR[58]	AF0	GPIO[58]	SIUL	I	I	Tristate	95	79	—	79	57	—
		AF1	—	—	—								
		AF2	—	—	—								
		AF3	—	—	—								
		—	ADC0_P[14]	ADC_0	I								
		—	ADC1_P[14]	ADC_1	I								
PD[11]	PCR[59]	AF0	GPIO[59]	SIUL	I	I	Tristate	96	80	—	80	58	—
		AF1	—	—	—								
		AF2	—	—	—								
		AF3	—	—	—								
		—	ADC0_P[15]	ADC_0	I								
		—	ADC1_P[15]	ADC_1	I								
PD[12]	PCR[60]	AF0	GPIO[60]	SIUL	I/O	J	Tristate	100	—	—	82	60	—
		AF1	CS5_0	DSPI_0	O								
		AF2	E0UC[24]	eMIOS_0	I/O								
		AF3	—	—	I/O								
		—	ADC0_S[4]	ADC_0	—								
					I								
PD[13]	PCR[61]	AF0	GPIO[61]	SIUL	I/O	J	Tristate	102	84	—	84	62	—
		AF1	CS0_1	DSPI_1	I/O								
		AF2	E0UC[25]	eMIOS_0	I/O								
		AF3	—	—	I/O								
		—	ADC0_S[5]	ADC_0	—								
					I								
PD[14]	PCR[62]	AF0	GPIO[62]	SIUL	I/O	J	Tristate	104	86	—	86	64	—
		AF1	CS1_1	DSPI_1	O								
		AF2	E0UC[26]	eMIOS_0	I/O								
		AF3	—	—	I/O								
		—	ADC0_S[6]	ADC_0	—								
					I								
PD[15]	PCR[63]	AF0	GPIO[63]	SIUL	I/O	J	Tristate	106	88	—	88	66	—
		AF1	CS2_1	DSPI_1	O								
		AF2	E0UC[27]	eMIOS_0	I/O								
		AF3	—	—	I/O								
		—	ADC0_S[7]	ADC_0	—								
					I								

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port E													
PE[0]	PCR[64]	AF0 AF1 AF2 AF3 — —	GPIO[64] E0UC[16] — — WKPU[6] ⁵ CAN5RX	SIUL eMIOS_0 — — WKPU FlexCAN_5	I/O I/O — — I I	S	Tristate	18	10	3	10	6	—
PE[1]	PCR[65]	AF0 AF1 AF2 AF3	GPIO[65] E0UC[17] CAN5TX —	SIUL eMIOS_0 FlexCAN_5 —	I/O I/O O —	M	Tristate	20	12	4	12	8	—
PE[2]	PCR[66]	AF0 AF1 AF2 AF3 — —	GPIO[66] E0UC[18] — — EIRQ[21] SIN_1	SIUL eMIOS_0 — — SIUL DSPI_1	I/O I/O — — I I	M	Tristate	156	128	—	128	89	—
PE[3]	PCR[67]	AF0 AF1 AF2 AF3	GPIO[67] E0UC[19] SOUT_1 —	SIUL eMIOS_0 DSPI_1 —	I/O I/O O —	M	Tristate	157	129	—	129	90	—
PE[4]	PCR[68]	AF0 AF1 AF2 AF3 —	GPIO[68] E0UC[20] SCK_1 — EIRQ[9]	SIUL eMIOS_0 DSPI_1 — SIUL	I/O I/O I/O — I	M	Tristate	160	132	—	132	93	—
PE[5]	PCR[69]	AF0 AF1 AF2 AF3	GPIO[69] E0UC[21] CS0_1 MA[2]	SIUL eMIOS_0 DSPI_1 ADC_0	I/O I/O I/O O	M	Tristate	161	133	—	133	94	—
PE[6]	PCR[70]	AF0 AF1 AF2 AF3 —	GPIO[70] E0UC[22] CS3_0 MA[1] EIRQ[22]	SIUL eMIOS_0 DSPI_0 ADC_0 SIUL	I/O I/O O O O I	M	Tristate	167	139	—	139	95	—
PE[7]	PCR[71]	AF0 AF1 AF2 AF3 —	GPIO[71] E0UC[23] CS2_0 MA[0] EIRQ[23]	SIUL eMIOS_0 DSPI_0 ADC_0 SIUL	I/O I/O O O O I	M	Tristate	168	140	—	140	96	—
PE[8]	PCR[72]	AF0 AF1 AF2 AF3	GPIO[72] CAN2TX E0UC[22] CAN3TX	SIUL FlexCAN_2 eMIOS_0 FlexCAN_3	I/O O I/O O	M	Tristate	21	13	—	13	9	—
PE[9]	PCR[73]	AF0 AF1 AF2 AF3 — — —	GPIO[73] — E0UC[23] — WKPU[7] ⁵ CAN2RX CAN3RX	SIUL — eMIOS_0 — WKPU FlexCAN_2 FlexCAN_3	I/O — I/O — I I I	S	Tristate	22	14	—	14	10	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PE[10]	PCR[74]	AF0 AF1 AF2 AF3 —	GPIO[74] LIN3TX CS3_1 E1UC[30] EIRQ[10]	SIUL LINFlex_3 DSPI_1 eMIOS_1 SIUL	I/O O O I/O I	S	Tristate	23	15	—	15	11	—
PE[11]	PCR[75]	AF0 AF1 AF2 AF3 — —	GPIO[75] E0UC[24] CS4_1 — LIN3RX WKPU[14] ⁵	SIUL eMIOS_0 DSPI_1 — LINFlex_3 WKPU	I/O I/O O — I I	S	Tristate	25	17	—	17	13	—
PE[12]	PCR[76]	AF0 AF1 AF2 AF3 — — —	GPIO[76] — E1UC[19] ¹² — EIRQ[11] SIN_2 ADC1_S[7]	SIUL — eMIOS_1 — SIUL DSPI_2 ADC_1	I/O — I/O — I I I	J	Tristate	133	109	—	109	76	—
PE[13]	PCR[77]	AF0 AF1 AF2 AF3	GPIO[77] SOUT_2 E1UC[20] —	SIUL DSPI_2 eMIOS_1 —	I/O O I/O —	S	Tristate	127	103	—	103	—	—
PE[14]	PCR[78]	AF0 AF1 AF2 AF3 —	GPIO[78] SCK_2 E1UC[21] — EIRQ[12]	SIUL DSPI_2 eMIOS_1 — SIUL	I/O I/O I/O — I	S	Tristate	136	112	—	112	—	—
PE[15]	PCR[79]	AF0 AF1 AF2 AF3	GPIO[79] CS0_2 E1UC[22] —	SIUL DSPI_2 eMIOS_1 —	I/O I/O I/O —	M	Tristate	137	113	—	113	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port F													
PF[0]	PCR[80]	AF0 AF1 AF2 AF3 —	GPIO[80] E0UC[10] CS3_1 — ADC0_S[8]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	63	55	—	55	—	—
PF[1]	PCR[81]	AF0 AF1 AF2 AF3 —	GPIO[81] E0UC[11] CS4_1 — ADC0_S[9]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	64	56	—	56	—	—
PF[2]	PCR[82]	AF0 AF1 AF2 AF3 —	GPIO[82] E0UC[12] CS0_2 — ADC0_S[10]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O I/O — I	J	Tristate	65	57	—	57	—	—
PF[3]	PCR[83]	AF0 AF1 AF2 AF3 —	GPIO[83] E0UC[13] CS1_2 — ADC0_S[11]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	66	58	—	58	—	—
PF[4]	PCR[84]	AF0 AF1 AF2 AF3 —	GPIO[84] E0UC[14] CS2_2 — ADC0_S[12]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	67	59	—	59	—	—
PF[5]	PCR[85]	AF0 AF1 AF2 AF3 —	GPIO[85] E0UC[22] CS3_2 — ADC0_S[13]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	68	60	—	60	—	—
PF[6]	PCR[86]	AF0 AF1 AF2 AF3 —	GPIO[86] E0UC[23] CS1_1 — ADC0_S[14]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	69	61	—	61	—	—
PF[7]	PCR[87]	AF0 AF1 AF2 AF3 —	GPIO[87] — CS2_1 — ADC0_S[15]	SIUL — DSPI_1 — ADC_0	I/O — O — I	J	Tristate	70	62	—	62	—	—
PF[8]	PCR[88]	AF0 AF1 AF2 AF3	GPIO[88] CAN3TX CS4_0 CAN2TX	SIUL FlexCAN_3 DSPI_0 FlexCAN_2	I/O O O O	M	Tristate	42	34	17	34	—	—
PF[9]	PCR[89]	AF0 AF1 AF2 AF3 — — —	GPIO[89] E1UC[1] CS5_0 — WKPU[22] ⁵ CAN2RX CAN3RX	SIUL eMIOS_1 DSPI_0 — WKPU FlexCAN_2 FlexCAN_3	I/O I/O O — I I I	S	Tristate	41	33	16	33	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PF[10]	PCR[90]	AF0 AF1 AF2 AF3	GPIO[90] CS1_0 LIN4TX E1UC[2]	SIUL DSPI_0 LINFlex_4 eMIOS_1	I/O O O I/O	M	Tristate	46	38	—	38	—	—
PF[11]	PCR[91]	AF0 AF1 AF2 AF3 — —	GPIO[91] CS2_0 E1UC[3] — WKPU[15] ⁵ LIN4RX	SIUL DSPI_0 eMIOS_1 — WKPU LINFlex_4	I/O O I/O — I I	S	Tristate	47	39	—	39	—	—
PF[12]	PCR[92]	AF0 AF1 AF2 AF3	GPIO[92] E1UC[25] LIN5TX —	SIUL eMIOS_1 LINFlex_5 —	I/O I/O O —	M	Tristate	43	35	—	35	—	—
PF[13]	PCR[93]	AF0 AF1 AF2 AF3 — —	GPIO[93] E1UC[26] — — WKPU[16] ⁵ LIN5RX	SIUL eMIOS_1 — — WKPU LINFlex_5	I/O I/O — — I I	S	Tristate	49	41	—	41	—	—
PF[14]	PCR[94]	AF0 AF1 AF2 AF3	GPIO[94] CAN4TX E1UC[27] CAN1TX	SIUL FlexCAN_4 eMIOS_1 FlexCAN_1	I/O O I/O O	M	Tristate	126	102	—	102	—	—
PF[15]	PCR[95]	AF0 AF1 AF2 AF3 — — —	GPIO[95] E1UC[4] — — EIRQ[13] CAN1RX CAN4RX	SIUL eMIOS_1 — — SIUL FlexCAN_1 FlexCAN_4	I/O I/O — — I I I	S	Tristate	125	101	45	101	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port G													
PG[0]	PCR[96]	AF0 AF1 AF2 AF3	GPIO[96] CAN5TX E1UC[23] —	SIUL FlexCAN_5 eMIOS_1 —	I/O O I/O —	M	Tristate	122	98	—	98	—	—
PG[1]	PCR[97]	AF0 AF1 AF2 AF3 — —	GPIO[97] — E1UC[24] — EIRQ[14] CAN5RX	SIUL — eMIOS_1 — SIUL FlexCAN_5	I/O — I/O — I I	S	Tristate	121	97	—	97	—	—
PG[2]	PCR[98]	AF0 AF1 AF2 AF3	GPIO[98] E1UC[11] SOUT_3 —	SIUL eMIOS_1 DSPI_3 —	I/O I/O O —	M	Tristate	16	8	—	8	—	—
PG[3]	PCR[99]	AF0 AF1 AF2 AF3 —	GPIO[99] E1UC[12] CS0_3 — WKPU[17] ⁵	SIUL eMIOS_1 DSPI_3 — WKPU	I/O I/O I/O — I	S	Tristate	15	7	—	7	—	—
PG[4]	PCR[100]	AF0 AF1 AF2 AF3	GPIO[100] E1UC[13] SCK_3 —	SIUL eMIOS_1 DSPI_3 —	I/O I/O I/O —	M	Tristate	14	6	—	6	—	—
PG[5]	PCR[101]	AF0 AF1 AF2 AF3 — —	GPIO[101] E1UC[14] — — WKPU[18] ⁵ SIN_3	SIUL eMIOS_1 — — WKPU DSPI_3	I/O I/O — — I I	S	Tristate	13	5	—	5	—	—
PG[6]	PCR[102]	AF0 AF1 AF2 AF3	GPIO[102] E1UC[15] LIN6TX —	SIUL eMIOS_1 LINFlex_6 —	I/O I/O O —	M	Tristate	38	30	—	30	—	—
PG[7]	PCR[103]	AF0 AF1 AF2 AF3 — —	GPIO[103] E1UC[16] E1UC[30] — WKPU[20] ⁵ LIN6RX	SIUL eMIOS_1 eMIOS_1 — WKPU LINFlex_6	I/O I/O I/O — I I	S	Tristate	37	29	—	29	—	—
PG[8]	PCR[104]	AF0 AF1 AF2 AF3 —	GPIO[104] E1UC[17] LIN7TX CS0_2 EIRQ[15]	SIUL eMIOS_1 LINFlex_7 DSPI_2 SIUL	I/O I/O O I/O I	S	Tristate	34	26	—	26	—	—
PG[9]	PCR[105]	AF0 AF1 AF2 AF3 — —	GPIO[105] E1UC[18] — SCK_2 WKPU[21] ⁵ LIN7RX	SIUL eMIOS_1 — DSPI_2 WKPU LINFlex_7	I/O I/O — I/O I I	S	Tristate	33	25	—	25	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PG[10]	PCR[106]	AF0 AF1 AF2 AF3 —	GPIO[106] E0UC[24] E1UC[31] — SIN_4	SIUL eMIOS_0 eMIOS_1 — DSPI_4	I/O I/O I/O — I	S	Tristate	138	114	—	114	—	—
PG[11]	PCR[107]	AF0 AF1 AF2 AF3	GPIO[107] E0UC[25] CS0_4 —	SIUL eMIOS_0 DSPI_4 —	I/O I/O I/O —	M	Tristate	139	115	—	115	—	—
PG[12]	PCR[108]	AF0 AF1 AF2 AF3	GPIO[108] E0UC[26] SOUT_4 —	SIUL eMIOS_0 DSPI_4 —	I/O I/O O —	M	Tristate	116	92	—	92	—	—
PG[13]	PCR[109]	AF0 AF1 AF2 AF3	GPIO[109] E0UC[27] SCK_4 —	SIUL eMIOS_0 DSPI_4 —	I/O I/O I/O —	M	Tristate	115	91	—	91	—	—
PG[14]	PCR[110]	AF0 AF1 AF2 AF3	GPIO[110] E1UC[0] LIN8TX —	SIUL eMIOS_1 LINFlex_8 —	I/O I/O O —	S	Tristate	134	110	—	110	—	—
PG[15]	PCR[111]	AF0 AF1 AF2 AF3 —	GPIO[111] E1UC[1] — — LIN8RX	SIUL eMIOS_1 — — LINFlex_8	I/O I/O — — I	M	Tristate	135	111	—	111	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port H													
PH[0]	PCR[112]	AF0 AF1 AF2 AF3 —	GPIO[112] E1UC[2] — — SIN_1	SIUL eMIOS_1 — — DSPI_1	I/O I/O — — I	M	Tristate	117	93	—	93	—	—
PH[1]	PCR[113]	AF0 AF1 AF2 AF3	GPIO[113] E1UC[3] SOUT_1 —	SIUL eMIOS_1 DSPI_1 —	I/O I/O O —	M	Tristate	118	94	—	94	—	—
PH[2]	PCR[114]	AF0 AF1 AF2 AF3	GPIO[114] E1UC[4] SCK_1 —	SIUL eMIOS_1 DSPI_1 —	I/O I/O I/O —	M	Tristate	119	95	—	95	—	—
PH[3]	PCR[115]	AF0 AF1 AF2 AF3	GPIO[115] E1UC[5] CS0_1 —	SIUL eMIOS_1 DSPI_1 —	I/O I/O I/O —	M	Tristate	120	96	—	96	—	—
PH[4]	PCR[116]	AF0 AF1 AF2 AF3	GPIO[116] E1UC[6] — —	SIUL eMIOS_1 — —	I/O I/O — —	M	Tristate	162	134	—	134	—	—
PH[5]	PCR[117]	AF0 AF1 AF2 AF3	GPIO[117] E1UC[7] — —	SIUL eMIOS_1 — —	I/O I/O — —	S	Tristate	163	135	—	135	—	—
PH[6]	PCR[118]	AF0 AF1 AF2 AF3	GPIO[118] E1UC[8] — MA[2]	SIUL eMIOS_1 — ADC_0	I/O I/O — O	M	Tristate	164	136	—	136	—	—
PH[7]	PCR[119]	AF0 AF1 AF2 AF3	GPIO[119] E1UC[9] CS3_2 MA[1]	SIUL eMIOS_1 DSPI_2 ADC_0	I/O I/O O O	M	Tristate	165	137	—	137	—	—
PH[8]	PCR[120]	AF0 AF1 AF2 AF3	GPIO[120] E1UC[10] CS2_2 MA[0]	SIUL eMIOS_1 DSPI_2 ADC_0	I/O I/O O O	M	Tristate	166	138	—	138	—	—
PH[9] ¹⁰	PCR[121]	AF0 AF1 AF2 AF3	GPIO[121] — TCK —	SIUL — JTAGC —	I/O — I —	S	Input, weak pull-up	155	127	60	127	88	60
PH[10] ¹⁰	PCR[122]	AF0 AF1 AF2 AF3	GPIO[122] — TMS —	SIUL — JTAGC —	I/O — I —	M	Input, weak pull-up	148	120	53	120	81	53
PH[11]	PCR[123]	AF0 AF1 AF2 AF3	GPIO[123] SOUT_3 CS0_4 E1UC[5]	SIUL DSPI_3 DSPI_4 eMIOS_1	I/O O I/O I/O	M	Tristate	140	—	—	—	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PH[12]	PCR[124]	AF0 AF1 AF2 AF3	GPIO[124] SCK_3 CS1_4 E1UC[25]	SIUL DSPI_3 DSPI_4 eMIOS_1	I/O I/O O I/O	M	Tristate	141	—	—	—	—	—
PH[13]	PCR[125]	AF0 AF1 AF2 AF3	GPIO[125] SOUT_4 CS0_3 E1UC[26]	SIUL DSPI_4 DSPI_3 eMIOS_1	I/O O I/O I/O	M	Tristate	9	—	—	—	—	—
PH[14]	PCR[126]	AF0 AF1 AF2 AF3	GPIO[126] SCK_4 CS1_3 E1UC[27]	SIUL DSPI_4 DSPI_3 eMIOS_1	I/O I/O O I/O	M	Tristate	10	—	—	—	—	—
PH[15]	PCR[127]	AF0 AF1 AF2 AF3	GPIO[127] SOUT_5 — E1UC[17]	SIUL DSPI_5 — eMIOS_1	I/O O — I/O	M	Tristate	8	—	—	—	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port I													
PI[0]	PCR[128]	AF0 AF1 AF2 AF3	GPIO[128] E0UC[28] LIN8TX —	SIUL eMIOS_0 LINFlex_8 —	I/O I/O O —	S	Tristate	172	—	—	—	—	—
PI[1]	PCR[129]	AF0 AF1 AF2 AF3 — —	GPIO[129] E0UC[29] — — WKPU[24] ⁵ LIN8RX	SIUL eMIOS_0 — — WKPU LINFlex_8	I/O I/O — — I I	S	Tristate	171	—	—	—	—	—
PI[2]	PCR[130]	AF0 AF1 AF2 AF3	GPIO[130] E0UC[30] LIN9TX —	SIUL eMIOS_0 LINFlex_9 —	I/O I/O O —	S	Tristate	170	—	—	—	—	—
PI[3]	PCR[131]	AF0 AF1 AF2 AF3 — —	GPIO[131] E0UC[31] — — WKPU[23] ⁵ LIN9RX	SIUL eMIOS_0 — — WKPU LINFlex_9	I/O I/O — — I I	S	Tristate	169	—	—	—	—	—
PI[4]	PCR[132]	AF0 AF1 AF2 AF3	GPIO[132] E1UC[28] SOUT_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O O —	S	Tristate	143	—	—	—	—	—
PI[5]	PCR[133]	AF0 AF1 AF2 AF3	GPIO[133] E1UC[29] SCK_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	142	—	—	—	—	—
PI[6]	PCR[134]	AF0 AF1 AF2 AF3	GPIO[134] E1UC[30] CS0_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	11	—	—	—	—	—
PI[7]	PCR[135]	AF0 AF1 AF2 AF3	GPIO[135] E1UC[31] CS1_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O O —	S	Tristate	12	—	—	—	—	—
PI[8]	PCR[136]	AF0 AF1 AF2 AF3 —	GPIO[136] CAN6TX — — ADC0_S[16]	SIUL FLEXCAN_6 — — ADC_0	I/O O — — I	J	Tristate	108	—	41	—	—	—
PI[9]	PCR[137]	AF0 AF1 AF2 AF3 —	GPIO[137] CAN7TX — — ADC0_S[17]	SIUL FLEXCAN_7 — — ADC_0	I/O O — — I	J	Tristate	109	—	42	—	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
PI[10]	PCR[138]	AF0 AF1 AF2 AF3 —	GPIO[138] CAN6RX — — ADC0_S[18]	SIUL FLEXCAN_6 — — ADC_0	I/O I — — I	J	Tristate	110	—	43	—	—	—
PI[11]	PCR[139]	AF0 AF1 AF2 AF3 — —	GPIO[139] CAN7RX — — ADC0_S[19] SIN_3	SIUL FLEXCAN_7 — — ADC_0 DSPI_3	I/O I — — I I	J	Tristate	111	—	44	—	—	—
PI[12]	PCR[140]	AF0 AF1 AF2 AF3 —	GPIO[140] CS0_3 — — ADC0_S[20]	SIUL DSPI_3 — — ADC_0	I/O I/O — — I	J	Tristate	112	—	—	—	—	—
PI[13]	PCR[141]	AF0 AF1 AF2 AF3 —	GPIO[141] CS1_3 — — ADC0_S[21]	SIUL DSPI_3 — — ADC_0	I/O O — — I	J	Tristate	113	—	—	—	—	—
PI[14]	PCR[142]	AF0 AF1 AF2 AF3 — —	GPIO[142] — — — ADC0_S[22] SIN_4	SIUL — — — ADC_0 DSPI_4	I/O — — — I I	J	Tristate	76	—	—	—	—	—
PI[15]	PCR[143]	AF0 AF1 AF2 AF3 —	GPIO[143] CS0_4 — — ADC0_S[23]	SIUL DSPI_4 — — ADC_0	I/O I/O — — I	J	Tristate	75	—	—	—	—	—

Port pin	PCR register	Alternate function ¹	Function	Peripheral	I/O Direction ²	Pad type	RESET config. ³	Pin number ⁴					
								60L7 176 LQFP	60L5 144 LQFP	60L1 64 LQFP	50L5 144 LQFP	50L3 100 LQFP	50L1 64 LQFP
Port J													
PJ[0]	PCR[144]	AF0 AF1 AF2 AF3 —	GPIO[144] CS1_4 — — ADC0_S[24]	SIUL DSPI_4 — — ADC_0	I/O O — — I	J	Tristate	74	—	—	—	—	—
PJ[1]	PCR[145]	AF0 AF1 AF2 AF3 — —	GPIO[145] — — — ADC0_S[25] SIN_5	SIUL — — — ADC_0 DSPI_5	I/O — — — I I	J	Tristate	73	—	—	—	—	—
PJ[2]	PCR[146]	AF0 AF1 AF2 AF3 —	GPIO[146] CS0_5 — — ADC0_S[26]	SIUL DSPI_5 — — ADC_0	I/O I/O — — I	J	Tristate	72	—	—	—	—	—
PJ[3]	PCR[147]	AF0 AF1 AF2 AF3 —	GPIO[147] CS1_5 — — ADC0_S[27]	SIUL DSPI_5 — — ADC_0	I/O O — — I	J	Tristate	71	—	—	—	—	—
PJ[4]	PCR[148]	AF0 AF1 AF2 AF3	GPIO[148] SCK_5 E1UC[18] —	SIUL DSPI_5 eMIOS_1 —	I/O I/O I/O —	M	Tristate	5	—	—	—	—	—
PJ[5]	PCR[149]	—	MDO0	NDI	O	F	Tristate	—	—	—	—	—	—
PJ[6]	PCR[150]	—	MDO1	NDI	O	F	Tristate	—	—	—	—	—	—
PJ[7]	PCR[151]	—	MDO2	NDI	O	F	Tristate	—	—	—	—	—	—
PJ[8]	PCR[152]	—	MDO3	NDI	O	F	Tristate	—	—	—	—	—	—
PJ[9]	PCR[153]	AF0	MSEO	NDI	O	F	Output	—	—	—	—	—	—
PJ[10]	PCR[154]	AF0	MCKO	NDI	O	F	Output	—	—	—	—	—	—
PJ[11]	PCR[155]	AF0	EVTI	NDI	I	F	Input, weak pull-up	—	—	—	—	—	—
PJ[12]	PCR[156]	AF0	EVTO	NDI	O	F	Output	—	—	—	—	—	—
PJ[13]	PCR[157]	AF0	SWCLK	SWD	I	M	Input	—	—	—	—	—	—
PJ[14]	PCR[158]	AF0	SWIO	SWD	I/O	M	Tristate	—	—	—	—	—	—

¹ Alternate functions are chosen by setting the values of the PCR.PA bitfields inside the SIUL module. PCR.PA = 00 -> AF0; PCR.PA = 01 -> AF1; PCR.PA = 10 -> AF2; PCR.PA = 11 -> AF3. This is intended to select the output functions; to use one of the input functions, the PCR.IBE bit must be written to '1', regardless of the values selected in the PCR.PA bitfields. For this reason, the value corresponding to an input only function is reported as "—".

² Multiple inputs are routed to all respective modules internally. The input of some modules must be configured by setting the values of the PSMIO.PADSELx bitfields inside the SIUL module.

³ The RESET configuration applies during and after reset.

⁴ 50L3 refers to CCFC2012BC50L3, whose package is benchmarked with SPC560B50L3*; 60L5 refers to CCFC2012BC60L5, whose package is benchmarked with SPC560B60L5C6E0; 60L7 refers to CCFC2012BC60L7, whose package is benchmarked with SPC560B64L7C(B)6E0X; 50L1 refers to CCFC2012BC50L1, whose package is benchmarked with SPC560B50L1*; 50L5 refers to CCFC2012BC50L5, whose package is benchmarked with CCFC2002BC and SPC560B50L5C6E0.

- ⁵ All WKPU pins also support external interrupt capability. See the WKPU chapter for further details.
- ⁶ NMI has higher priority than alternate function. When NMI is selected, the PCR.AF field is ignored.
- ⁷ “Not applicable” because these functions are available only while the device is booting. Refer to the BAM information for details.
- ⁸ Value of PCR.IBE bit must be 0
- ⁹ This wakeup input cannot be used to exit STANDBY mode.
- ¹⁰ Out of reset all the functional pins except PC[0:1] and PH[9:10] are available to the user as GPIO. PC[0:1] are available as JTAG pins (TDI and TDO respectively). PH[9:10] are available as JTAG pins (TCK and TMS respectively). It is up to the user to configure these pins as GPIO when needed.
- ¹¹ PC[1] is a fast/medium pad but is in medium configuration by default. This pad is in Alternate Function 2 mode after reset which has TDO functionality. The reset value of PCR.OBE is '1', but this setting has no impact as long as this pad stays in AF2 mode. After configuring this pad as GPIO (PCR.PA = 0), output buffer is enabled as reset value of PCR.OBE = 1.
- ¹² Not available in 100 LQFP package

3.8 Nexus 2+ pins

Table 6. Nexus 2+ pin descriptions

Port pin	Function	I/O direction	Pad type	Function after reset	Pin number		
					100 LQFP	144 LQFP	
MCKO	Message clock out	O	F	—	—	—	
MDO0	Message data out 0	O	M	—	—	—	
MDO1	Message data out 1	O	M	—	—	—	
MDO2	Message data out 2	O	M	—	—	—	
MDO3	Message data out 3	O	M	—	—	—	
EVTI	Event in	I	M	Pull-up	—	—	
$\overline{\text{EVTO}}$	Event out	O	M	—	—	—	
$\overline{\text{MSEO}}$	Message start/end out	O	M	—	—	—	

4 Electrical characteristics

This section contains electrical characteristics of the device as well as temperature and power considerations.

This product contains devices to protect the inputs against damage due to high static voltages. However, it is advisable to take precautions to avoid application of any voltage higher than the specified maximum rated voltages.

To enhance reliability, unused inputs can be driven to an appropriate logic voltage level (VDD or VSS). This could be done by the internal pull-up and pull-down, which is provided by the product for most general purpose pins.

The parameters listed in the following tables represent the characteristics of the device and its demands on the system.

In the tables where the device logic provides signals with their respective timing characteristics, the symbol “CC” for Controller Characteristics is included in the Symbol column.

In the tables where the external system must provide signals with their respective timing characteristics to the device, the symbol “SR” for System Requirement is included in the Symbol column.

4.1 Parameter classification

The electrical parameters shown in this supplement are guaranteed by various methods. To give the customer a better understanding, the classifications listed in Table 4-1 are used and the parameters are tagged accordingly in the tables where appropriate.

Table 4-1. Parameter classifications

Classification tag	Tag description
P	Those parameters are guaranteed during production testing on each individual device.
C	Those parameters are achieved by the design characterization by measuring a statistically relevant sample size across process variations.
T	Those parameters are achieved by design characterization on a small sample size from typical devices under typical conditions unless otherwise noted. All values shown in the typical column are within this category.
D	Those parameters are derived mainly from simulations.

NOTE

The classification is shown in the column labeled “C” in the parameter tables where appropriate.

4.2 NVUSRO register

Bit values in the Non-Volatile User Options (NVUSRO) Register control portions of the device configuration, namely electrical parameters such as high voltage supply and oscillator margin, as well as digital functionality (watchdog enable/disable after reset). For a detailed description of the NVUSRO register, please refer to the device reference manual.

4.2.1 NVUSRO[PAD3V5V] field description

The DC electrical characteristics are dependent on the PAD3V5V bit value. Table 4-2 shows how NVUSRO[PAD3V5V] controls the device configuration.

Table 4-2. PAD3V5V field description¹

Value ²	Description
0	High voltage supply is 5.0 V
1	High voltage supply is 3.3 V

¹ See the device reference manual for more information on the NVUSRO register.

² Default manufacturing value is '1'. Value can be programmed by customer in Shadow Flash.

4.2.2 NVUSRO[WATCHDOG_EN] field description

The watchdog enable/disable configuration after reset is dependent on the WATCHDOG_EN bit value. Table 4-3 shows how NVUSRO[WATCHDOG_EN] controls the device configuration.

Table 4-3. WATCHDOG_EN field description

Value ¹	Description
0	Disable after reset
1	Enable after reset

¹ Default manufacturing value is '1'. Value can be programmed by customer in Shadow Flash.

4.3 Absolute maximum ratings

Table 4-4. Absolute maximum ratings

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V _{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V _{DD}	SR	Voltage on VDD_HV pins with respect to ground (V _{SS})	—	−0.3	6.0	
V _{SS_LV}	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V _{SS})	—	V _{SS} − 0.1	V _{SS} + 0.1	V
V _{DD_BV}	SR	Voltage on VDD_BV (regulator supply) pin with respect to ground (V _{SS})	—	−0.3	6.0	
V _{SS_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) pins with respect to ground (V _{SS})	—	V _{SS} − 0.1	V _{SS} + 0.1	V
V _{DD_ADC}	SR	Voltage on VDD_HV_ADC0, VDD_HV_ADC1 (ADC reference) pins with respect to ground (V _{SS})	—	−0.3	6.0	
V _{IN}	SR	Voltage on any GPIO pin with respect to ground (V _{SS})	Relative to V _{DD}	V _{DD} − 0.3	V _{DD} + 0.3	V
			Relative to V	−0.3	6.0	
I _{INJPAD}	SR	Injected input current on any pin during overload condition	—	−10	10	mA
I _{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	−50	50	
I _{AVGSEG}	SR	Sum of all the static I/O current within a supply segment	V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	70	mA
			V _{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	64	
T _{STORAGE}	SR	Storage temperature	—	−55	150	°C

NOTE

Stresses exceeding the recommended absolute maximum ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. During overload conditions ($V_{IN} > V_{DD}$ or $V_{IN} < V_{SS}$), the voltage on pins with respect to ground (V_{SS}) must not exceed the recommended values.

4.4 Recommended operating conditions

Table 12. Recommended operating conditions (3.3 V)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V _{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V _{DD} ¹	SR	Voltage on VDD_HV pins with respect to ground (V _{SS})	—	3.0	3.6	V
V _{SS_LV} ²	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V _{SS})	—	V − 0.1	V + 0.1	V
V _{DD_BV} ³	SR	Voltage on VDD_BV pin (regulator supply) with respect to ground (V _{SS})	—	3.0	3.6	V
			Relative to V	V _{DD} − 0.1	V _{DD} + 0.1	
V _{SS_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) pin with respect to ground (V _{SS})	—	V _{SS} − 0.1	V _{SS} + 0.1	V
V _{DD_ADC} ⁴	SR	Voltage on VDD_HV_ADC0, VDD_HV_ADC1 (ADC reference) with respect to ground (V _{SS})	—	3.0 ⁵	3.6	V
			Relative to V _{DD}	V _{DD} − 0.1	V _{DD} + 0.1	
V _{IN}	SR	Voltage on any GPIO pin with respect to ground (V _{SS})	—	V _{SS} − 0.1	—	V
			Relative to V	—	V + 0.1	
I _{INJPAD}	SR	Injected input current on any pin during overload condition	—	−5	5	mA
I _{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	−50	50	
TV _{DD}	SR	V _{DD} slope to ensure correct power up ⁶	—	3.0 ⁷	0.25 V/μs	V/s
T _A C-Grade Part	SR	Ambient temperature under bias	f _{CPU} ≤ 64 MHz ⁸	−40	85	°C
T _J C-Grade Part	SR	Junction temperature under bias	—	−40	110	
T _A V-Grade Part	SR	Ambient temperature under bias	f _{CPU} ≤ 64 MHz ⁸	−40	105	
T _J V-Grade Part	SR	Junction temperature under bias	—	−40	130	
T _A M-Grade Part	SR	Ambient temperature under bias	f _{CPU} ≤ 64 MHz ⁸	−40	125	
T _J M-Grade Part	SR	Junction temperature under bias	—	−40	150	

¹ 100 nF capacitance needs to be provided between each V_{DD}/V_{SS} pair.

² 330 nF capacitance needs to be provided between each V_{DD_LV}/V_{SS_LV} supply pair.

³ 470 nF capacitance needs to be provided between V_{DD_BV} and the nearest V_{SS_LV} (higher value may be needed depending on external regulator characteristics). Supply ramp slope on V_{DD_BV} should always be faster or equal to slope of V_{DD_HV} . Otherwise, device may enter regulator bypass mode if slope on V_{DD_BV} is slower.

⁴ 100 nF capacitance needs to be provided between V_{DD_ADC}/V_{SS_ADC} pair.

⁵ Full electrical specification cannot be guaranteed when voltage drops below 3.0 V. In particular, ADC electrical characteristics and I/Os DC electrical specification may not be guaranteed. When voltage drops below V_{LVDHVL} , device is reset.

⁶ Guaranteed by device validation

⁷ Minimum value of TV_{DD} must be guaranteed until V_{DD} reaches 2.6 V (maximum value of V_{PORH})

⁸ When the FMPLL uses the frequency modulation with a modulation depth of 4% from the center spread frequency, the maximum value of f_{CPU} is 66.56 MHz.

Table 13. Recommended operating conditions (5.0 V)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V_{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V_{DD}^1	SR	Voltage on VDD_HV pins with respect to ground (V_{SS})	—	4.5	5.5	V
		Voltage drop ²		3.0	5.5	
³	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V_{SS})	—	V - 0.1	V + 0.1	V
$V_{DD_BV}^4$	SR	Voltage on VDD_BV pin (regulator supply) with respect to ground (V_{SS})	—	4.5	5.5	V
		Voltage drop ²		3.0	5.5	
		Relative to V_{DD}		3.0	$V_{DD} + 0.1$	
V_{SS_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) pin with respect to ground (V_{SS})	—	$V_{SS} - 0.1$	$V_{SS} + 0.1$	V
$V_{DD_ADC}^5$	SR	Voltage on VDD_HV_ADC0, VDD_HV_ADC1 (ADC reference) with respect to ground (V_{SS})	—	4.5	5.5	V
		Voltage drop ²		3.0	5.5	
		Relative to V_{DD}		$V_{DD} - 0.1$	$V_{DD} + 0.1$	
V_{IN}	SR	Voltage on any GPIO pin with respect to ground (V_{SS})	—	$V_{SS} - 0.1$	—	V
		Relative to V		—	V + 0.1	
I_{INJPAD}	SR	Injected input current on any pin during overload condition	—	-5	5	mA
I_{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	-50	50	
TV_{DD}	SR	V_{DD} slope to ensure correct power up ⁶	—	3.0^7	0.25 V/ μ s	V/s
$T_{A \text{ C-Grade Part}}$	SR	Ambient temperature under bias	$f_{CPU} \leq 64 \text{ MHz}^8$	-40	85	°C
$T_{J \text{ C-Grade Part}}$	SR	Junction temperature under bias	—	-40	110	
$T_{A \text{ V-Grade Part}}$	SR	Ambient temperature under bias	$f_{CPU} \leq 64 \text{ MHz}^8$	-40	105	
$T_{J \text{ V-Grade Part}}$	SR	Junction temperature under bias	—	-40	130	
$T_{A \text{ M-Grade Part}}$	SR	Ambient temperature under bias	$f_{CPU} \leq 64 \text{ MHz}^8$	-40	125	
$T_{J \text{ M-Grade Part}}$	SR	Junction temperature under bias	—	-40	150	

¹ 100 nF capacitance needs to be provided between each V_{DD}/V_{SS} pair.

² Full device operation is guaranteed by design when the voltage drops below 4.5 V down to 3.0 V. However, certain analog electrical characteristics will not be guaranteed to stay within the stated limits.

³ 330 nF capacitance needs to be provided between each V_{DD_LV}/V_{SS_LV} supply pair.

⁴ 470 nF capacitance needs to be provided between V_{DD_BV} and the nearest V_{SS_LV} (higher value may be needed depending on external regulator characteristics). While the supply voltage ramps up, the slope on V_{DD_BV} should be less than $0.9V_{DD_HV}$ in order to ensure the device does not enter regulator bypass mode.

⁵ 100 nF capacitance needs to be provided between V_{DD_ADC}/V_{SS_ADC} pair.

⁶ Guaranteed by device validation

- ⁷ Minimum value of $T_{V_{DD}}$ must be guaranteed until V_{DD} reaches 2.6 V (maximum value of V_{PORH})
- ⁸ When the FMPLL uses the frequency modulation with a modulation depth of 4% from the center spread frequency, the maximum value of f_{CPU} is 66.56 MHz.

NOTE

RAM data retention is guaranteed with V_{DD_LV} not below 1.08 V.

4.5 Thermal characteristics

4.5.1 External ballast resistor recommendations

External ballast resistor on V_{DD_BV} pin helps in reducing the overall power dissipation inside the device. This resistor is required only when maximum power consumption exceeds the limit imposed by package thermal characteristics.

As stated in [Table 14](#) LQFP thermal characteristics, considering a thermal resistance of 144 LQFP as 48.3 °C/W, at ambient temperature $T_A = 125$ °C, the junction temperature T_j will cross 150 °C if the total power dissipation is greater than $(150 - 125)/48.3 = 517$ mW. Therefore, the total device current I_{DDMAX} at 125 °C/5.5 V must not exceed 94.1 mA (i.e., PD/V_{DD}). Assuming an average $I_{DD}(V_{DD_HV})$ of 15–20 mA consumption typically during device RUN mode, the LV domain consumption $I_{DD}(V_{DD_BV})$ is thus limited to $I_{DDMAX} - I_{DD}(V_{DD_HV})$, i.e., 80 mA.

Therefore, respecting the maximum power allowed as explained in [4.5.2, Package thermal characteristics](#), it is recommended to use this resistor only in the 125 °C/5.5 V operating corner as per the following guidelines:

- If $I_{DD}(V_{DD_BV}) < 80$ mA, then no resistor is required.
- If $80 \text{ mA} < I_{DD}(V_{DD_BV}) < 90$ mA, then 4 Ω resistor can be used.
- If $I_{DD}(V_{DD_BV}) > 90$ mA, then 8 Ω resistor can be used.

Using resistance in the range of 4–8 Ω, the gain will be around 10–20% of total consumption on V_{DD_BV} . For example, if 8 Ω resistor is used, then power consumption when $I_{DD}(V_{DD_BV})$ is 110 mA is equivalent to power consumption when $I_{DD}(V_{DD_BV})$ is 90 mA (approximately) when resistor not used.

In order to ensure correct power up, the minimum V_{DD_BV} to be guaranteed is 30 mV. If the supply ramp is slower than this value, then LVDHV3B monitoring ballast supply V_{DD_BV} pin gets triggered leading to device reset. Until the supply reaches certain threshold, this low voltage detector (LVD) generates destructive reset event in the system. This threshold depends on the maximum $I_{DD}(V_{DD_BV})$ possible across the external resistor.

4.5.2 Package thermal characteristics

Table 14. LQFP thermal characteristics¹

Symbol	C	D	Parameter	Conditions ²	Pin count	Value			Unit
						Min	Typ	Max	
$R_{\theta JA}$	CC	D	Thermal resistance, junction-to-ambient natural convection ³	Single-layer board — 1s	100	—	—	64	°C/W
					144	—	—	64	
					176	—	—	64	
				Four-layer board — 2s2p	100	—	—	49.7	
					144	—	—	48.3	
					176	—	—	47.3	
$R_{\theta JB}$	CC	D	Thermal resistance, junction-to-board ⁴	Single-layer board — 1s	100	—	—	36	°C/W
					144	—	—	38	
					176	—	—	38	
				Four-layer board — 2s2p	100	—	—	33.6	
					144	—	—	33.4	
					176	—	—	33.4	

Symbol	C	Parameter	Conditions ²	Pin count	Value			Unit
					Min	Typ	Max	
R _{θJC}	CC	Thermal resistance, junction-to-case ⁵	Single-layer board — 1s	100	—	—	23	°C/W
				144	—	—	23	
				176	—	—	23	
			Four-layer board — 2s2p	100	—	—	19.8	
				144	—	—	19.2	
				176	—	—	18.8	

¹ Thermal characteristics are targets based on simulation.

² V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C.

³ Junction-to-ambient thermal resistance determined per JEDEC JESD51-3 and JESD51-6. Thermal test board meets JEDEC specification for this package. When Greek letters are not available, the symbols are typed as R_{thJA} and R_{thJMA}.

⁴ Junction-to-board thermal resistance determined per JEDEC JESD51-8. Thermal test board meets JEDEC specification for the specified package. When Greek letters are not available, the symbols are typed as R_{thJB}.

⁵ Junction-to-case at the top of the package determined using MIL-STD 883 Method 1012.1. The cold plate temperature is used for the case temperature. Reported value includes the thermal resistance of the interface layer. When Greek letters are not available, the symbols are typed as R_{thJC}.

4.5.3 Power considerations

The average chip-junction temperature, T_J, in degrees Celsius, may be calculated using [Equation 1](#):

$$T_J = T_A + (P_D \times R_{\theta JA}) \quad \text{Eqn. 1}$$

Where:

T_A is the ambient temperature in °C.

R_{θJA} is the package junction-to-ambient thermal resistance, in °C/W.

P_D is the sum of P_{INT} and P_{I/O} (P_D = P_{INT} + P_{I/O}).

P_{INT} is the product of I_{DD} and V_{DD}, expressed in watts. This is the chip internal power.

P_{I/O} represents the power dissipation on input and output pins; user determined.

Most of the time for the applications, P_{I/O} < P_{INT} and may be neglected. On the other hand, P_{I/O} may be significant, if the device is configured to continuously drive external modules and/or memories.

An approximate relationship between P_D and T_J (if P_{I/O} is neglected) is given by:

$$P_D = K / (T_J + 273 \text{ °C}) \quad \text{Eqn. 2}$$

Therefore, solving equations [1](#) and [2](#):

$$K = P_D \times (T_A + 273 \text{ °C}) + R_{\theta JA} \times P_D^2 \quad \text{Eqn. 3}$$

Where:

K is a constant for the particular part, which may be determined from [Equation 3](#) by measuring P_D (at equilibrium) for a known T_A. Using this value of K, the values of P_D and T_J may be obtained by solving equations [1](#) and [2](#) iteratively for any value of T_A.

4.6 I/O pad electrical characteristics

4.6.1 I/O pad types

The device provides four main I/O pad types depending on the associated alternate functions:

- Slow pads—are the most common pads, providing a good compromise between transition time and low electromagnetic emission.
- Medium pads—provide transition fast enough for the serial communication channels with controlled current to reduce electromagnetic emission.
- Fast pads—provide maximum speed. These are used for improved Nexus debugging capability.
- Input only pads—are associated with ADC channels and 32 kHz low power external crystal oscillator providing low input leakage.

Medium and Fast pads can use slow configuration to reduce electromagnetic emission, at the cost of reducing AC performance.

4.6.2 I/O input DC characteristics

[Table 15](#) provides input DC electrical characteristics as described in [Figure 6](#).

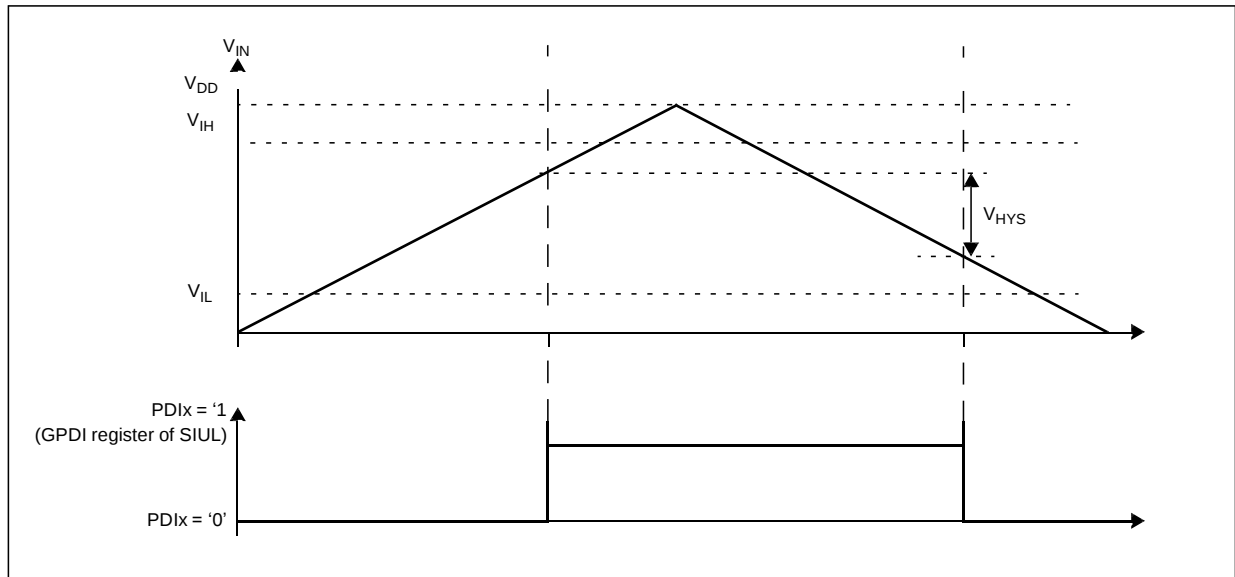


Figure 6. I/O input DC electrical characteristics definition

Table 15. I/O input DC electrical characteristics

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
V _{IH}	SR	P	Input high level CMOS (Schmitt Trigger)	—		0.65V _{DD}	—	V _{DD} + 0.4	V
V _{IL}	SR	P	Input low level CMOS (Schmitt Trigger)	—		-0.4	—	0.35V _{DD}	
V _{HYS}	CC	C	Input hysteresis CMOS (Schmitt Trigger)	—		0.1V _{DD}	—	—	
I _{LKG}	CC	D	Digital input leakage	No injection on adjacent pin	T _A = -40 °C	—	2	200	nA
		T _A = 25 °C			—	2	200		
		T _A = 85 °C			—	5	300		
		T _A = 105 °C			—	12	500		
		T _A = 125 °C			—	70	1000		
W _{FI} ²	SR	P	Wakeup input filtered pulse	—		—	—	40	ns
W _{NFI} ²	SR	P	Wakeup input not filtered pulse	—		1000	—	—	ns

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

² In the range from 40 to 1000 ns, pulses can be filtered or not filtered, according to operating temperature and voltage.

4.6.3 I/O output DC characteristics

The following tables provide DC characteristics for bidirectional pads:

- [Table 16](#) provides weak pull figures. Both pull-up and pull-down resistances are supported.

- Table 17 provides output driver characteristics for I/O pads when in SLOW configuration.
- Table 18 provides output driver characteristics for I/O pads when in MEDIUM configuration.
- Table 19 provides output driver characteristics for I/O pads when in FAST configuration.

Table 16. I/O pull-up/pull-down DC electrical characteristics

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
I _{WPU}	CC	P	Weak pull-up current absolute value	V _{IN} = V _{IL} , V _{DD} = 5.0 V ± 10%	PAD3V5V = 0	10	—	150	μA
		PAD3V5V = 1 ²			10	—	250		
		C			V _{IN} = V _{IL} , V _{DD} = 3.3 V ± 10%	PAD3V5V = 1	10	—	
I _{WPD}	CC	P	Weak pull-down current absolute value	V _{IN} = V _{IH} , V _{DD} = 5.0 V ± 10%	PAD3V5V = 0	10	—	150	μA
		PAD3V5V = 1			10	—	250		
		C			V _{IN} = V _{IH} , V _{DD} = 3.3 V ± 10%	PAD3V5V = 1	10	—	

¹ $V_{DD} = 3.3 \text{ V} \pm 10\% / 5.0 \text{ V} \pm 10\%$, $T_A = -40$ to 125°C , unless otherwise specified.

² The configuration PAD3V5 = 1 when $V_{DD} = 5 \text{ V}$ is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 17. SLOW configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
V _{OH}	CC	P	Output high level SLOW configuration	Push Pull	I _{OH} = −2 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	0.8V _{DD}	—	—	V
		I _{OH} = −2 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 1 ²			0.8V _{DD}	—	—		
		I _{OH} = −1 mA, V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)			V _{DD} − 0.8	—	—		
V _{OL}	CC	P	Output low level SLOW configuration	Push Pull	I _{OL} = 2 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	0.1V _{DD}	V
		I _{OL} = 2 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 1 ²			—	—	0.1V _{DD}		
		I _{OL} = 1 mA, V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)			—	—	0.5		

¹ $V_{DD} = 3.3 \text{ V} \pm 10\% / 5.0 \text{ V} \pm 10\%$, $T_A = -40$ to 125°C , unless otherwise specified

- ² The configuration PAD3V5 = 1 when $V_{DD} = 5\text{ V}$ is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 18. MEDIUM configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
V _{OH}	CC	C	Output high level MEDIUM configuration	Push Pull	I _{OH} = -3.8 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	0.8V _{DD}	—	—	V
		P			I _{OH} = -2 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	0.8V _{DD}	—	—	
		C			I _{OH} = -1 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 1 ²	0.8V _{DD}	—	—	
		C			I _{OH} = -1 mA, V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	V _{DD} - 0.8	—	—	
		C			I _{OH} = -100 μA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	0.8V _{DD}	—	—	
V _{OL}	CC	C	Output low level MEDIUM configuration	Push Pull	I _{OL} = 3.8 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	0.2V _{DD}	V
		P			I _{OL} = 2 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	0.1V _{DD}	
		C			I _{OL} = 1 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 1 ²	—	—	0.1V _{DD}	
		C			I _{OL} = 1 mA, V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	
		C			I _{OL} = 100 μA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	0.1V _{DD}	

¹ $V_{DD} = 3.3\text{ V} \pm 10\% / 5.0\text{ V} \pm 10\%$, $T_A = -40$ to $125\text{ }^\circ\text{C}$, unless otherwise specified

² The configuration PAD3V5 = 1 when $V_{DD} = 5\text{ V}$ is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 19. FAST configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
V _{OH}	CC	P	Output high level FAST configuration	Push Pull	I _{OH} = -14 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	0.8V _{DD}	—	—	V
		C			I _{OH} = -7 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 1 ²	0.8V _{DD}	—	—	
		C			I _{OH} = -11 mA, V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	V _{DD} - 0.8	—	—	
V _{OL}	CC	P	Output low level FAST configuration	Push Pull	I _{OL} = 14 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	0.1V _{DD}	V
		C			I _{OL} = 7 mA, V _{DD} = 5.0 V ± 10%, PAD3V5V = 1 ²	—	—	0.1V _{DD}	
		C			I _{OL} = 11 mA, V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

² The configuration PAD3V5 = 1 when V_{DD} = 5 V is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

4.6.4 Output pin transition times

Table 20. Output pin transition times

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
t _{tr}	CC	D	Output transition time output pin ² SLOW configuration	C _L = 25 pF	V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	50	ns
		T		C _L = 50 pF		—	—	100	
		D		C _L = 100 pF		—	—	125	
		D		C _L = 25 pF	V _{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	50	
		T		C _L = 50 pF		—	—	100	
		D		C _L = 100 pF		—	—	125	

Table 20. Output pin transition times (continued)

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
t _{tr}	CC	D	Output transition time output pin ² MEDIUM configuration	C _L = 25 pF	V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 SIUL.PCRx.SRC = 1	—	—	10	ns
		C _L = 50 pF		—		—	20		
		C _L = 100 pF		—		—	40		
		D		C _L = 25 pF	V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 SIUL.PCRx.SRC = 1	—	—	12	
		T		C _L = 50 pF		—	—	25	
		D		C _L = 100 pF		—	—	40	
t _{tr}	CC	D	Output transition time output pin ² FAST configuration	C _L = 25 pF	V _{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	4	ns
		C _L = 50 pF		—		—	6		
		C _L = 100 pF		—		—	12		
		C _L = 25 pF		V _{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	4		
		C _L = 50 pF			—	—	7		
		C _L = 100 pF			—	—	12		

¹ $V_{DD} = 3.3 \text{ V} \pm 10\%$ / $5.0 \text{ V} \pm 10\%$, $T_A = -40$ to 125°C , unless otherwise specified

² C_L includes device and package capacitances ($C_{PKG} < 5 \text{ pF}$).

4.6.5 I/O pad current specification

The I/O pads are distributed across the I/O supply segment. Each I/O supply segment is associated to a V_{DD}/V_{SS} supply pair as described in Table 21. Table 22 provides I/O consumption figures.

In order to ensure device reliability, the average current of the I/O on a single segment should remain below the I_{AVGSEG} maximum value.

Table 21. I/O supply segments

Package	Supply segment							
	1	2	3	4	5	6	7	8
176 LQFP	pin7 – pin27	pin28 – pin57	pin59 – pin85	pin86 – pin123	pin124 – pin150	pin151 – pin6	—	—
144 LQFP	pin20 – pin49	pin51 – pin99	pin100 – pin122	pin 123 – pin19	—	—	—	—
100 LQFP	pin16 – pin35	pin37 – pin69	pin70 – pin83	pin84 – pin15	—	—	—	—

Table 22. I/O consumption

Symbol	C	Parameter	Conditions ¹		Value			Unit	
					Min	Typ	Max		
I_{SWTSLW}^2	CC	D	Dynamic I/O current for SLOW configuration	$C_L = 25 \text{ pF}$	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$	—	—	20	mA
					$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$	—	—	16	
I_{SWTMED}^2	CC	D	Dynamic I/O current for MEDIUM configuration	$C_L = 25 \text{ pF}$	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$	—	—	29	mA
					$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$	—	—	17	
I_{SWTFST}^2	CC	D	Dynamic I/O current for FAST configuration	$C_L = 25 \text{ pF}$	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$	—	—	110	mA
					$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$	—	—	50	
I_{RMSSLW}	CC	D	Root mean square I/O current for SLOW configuration	$C_L = 25 \text{ pF}$, 2MHz	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$	—	—	2.3	mA
				$C_L = 25 \text{ pF}$, 4MHz		—	—	3.2	
				$C_L = 100 \text{ pF}$, 2MHz		—	—	6.6	
				$C_L = 25 \text{ pF}$, 2MHz	$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$	—	—	1.6	
				$C_L = 25 \text{ pF}$, 4MHz		—	—	2.3	
				$C_L = 100 \text{ pF}$, 2MHz		—	—	4.7	
I_{RMSMED}	CC	D	Root mean square I/O current for MEDIUM configuration	$C_L = 25 \text{ pF}$, 13MHz	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$	—	—	6.6	mA
				$C_L = 25 \text{ pF}$, 40MHz		—	—	13.4	
				$C_L = 100 \text{ pF}$, 13MHz		—	—	18.3	
				$C_L = 25 \text{ pF}$, 13MHz	$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$	—	—	5	
				$C_L = 25 \text{ pF}$, 40MHz		—	—	8.5	
				$C_L = 100 \text{ pF}$, 13MHz		—	—	11	
I_{RMSFST}	CC	D	Root mean square I/O current for FAST configuration	$C_L = 25 \text{ pF}$, 40MHz	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$	—	—	22	mA
				$C_L = 25 \text{ pF}$, 64MHz		—	—	33	
				$C_L = 100 \text{ pF}$, 40MHz		—	—	56	
				$C_L = 25 \text{ pF}$, 40MHz	$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$	—	—	14	
				$C_L = 25 \text{ pF}$, 64MHz		—	—	20	
				$C_L = 100 \text{ pF}$, 40MHz		—	—	35	
I_{AVGSEG}	SR	D	Sum of all the static I/O current within a supply segment	$V_{\text{DD}} = 5.0 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 0$		—	—	70	mA
				$V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$, $\text{PAD3V5V} = 1$		—	—	65	

¹ $V_{\text{DD}} = 3.3 \text{ V} \pm 10\%$ / $5.0 \text{ V} \pm 10\%$, $T_A = -40$ to 125°C , unless otherwise specified

² Stated maximum values represent peak consumption that lasts only a few ns during I/O transition.

Table 23 provides the weight of concurrent switching I/Os.

Due to the dynamic current limitations, the sum of the weight of concurrent switching I/Os on a single segment must not exceed 100% to ensure device functionality.

Table 23. I/O weight¹

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC ² = 1	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
6	4	4	PB[3]	5%	—	6%	—	13%	—	15%	—
			PC[9]	4%	—	5%	—	13%	—	15%	—
			PC[14]	4%	—	4%	—	13%	—	15%	—
			PC[15]	3%	4%	4%	4%	12%	18%	15%	16%
	—	—	PJ[4]	3%	4%	3%	3%	—	—	—	—
1	—	—	PH[15]	2%	3%	3%	3%	—	—	—	—
	—	—	PH[13]	3%	4%	3%	4%	—	—	—	—
	—	—	PH[14]	3%	4%	4%	4%	—	—	—	—
	—	—	PI[6]	4%	—	4%	—	—	—	—	—
	—	—	PI[7]	4%	—	4%	—	—	—	—	—
	4	—	PG[5]	4%	—	5%	—	10%	—	12%	—
		—	PG[4]	4%	6%	5%	5%	9%	13%	11%	12%
		—	PG[3]	4%	—	5%	—	9%	—	11%	—
		—	PG[2]	4%	6%	5%	5%	9%	12%	10%	11%
	4	4	PA[2]	4%	—	5%	—	8%	—	10%	—
			PE[0]	4%	—	5%	—	8%	—	9%	—
			PA[1]	4%	—	5%	—	8%	—	9%	—
			PE[1]	4%	6%	5%	6%	7%	10%	9%	9%
			PE[8]	4%	6%	5%	6%	7%	10%	8%	9%
			PE[9]	4%	—	5%	—	6%	—	8%	—
			PE[10]	4%	—	5%	—	6%	—	7%	—
			PA[0]	4%	6%	5%	5%	6%	8%	7%	7%
			PE[11]	4%	—	5%	—	5%	—	6%	—

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC ² = 1	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
2	1	—	PG[9]	9%	—	10%	—	9%	—	10%	—
		—	PG[8]	9%	—	11%	—	9%	—	11%	—
		1	PC[11]	9%	—	11%	—	9%	—	11%	—
			PC[10]	9%	13%	11%	12%	9%	13%	11%	12%
		—	PG[7]	9%	—	11%	—	9%	—	11%	—
		—	PG[6]	10%	14%	11%	12%	10%	14%	11%	12%
		1	PB[0]	10%	14%	12%	12%	10%	14%	12%	12%
			PB[1]	10%	—	12%	—	10%	—	12%	—
		—	PF[9]	10%	—	12%	—	10%	—	12%	—
		—	PF[8]	10%	14%	12%	13%	10%	14%	12%	13%
		—	PF[12]	10%	15%	12%	13%	10%	15%	12%	13%
		1	PC[6]	10%	—	12%	—	10%	—	12%	—
			PC[7]	10%	—	12%	—	10%	—	12%	—
		—	PF[10]	10%	14%	11%	12%	10%	14%	11%	12%
		—	PF[11]	9%	—	11%	—	9%	—	11%	—
		1	PA[15]	8%	12%	10%	10%	8%	12%	10%	10%
		—	PF[13]	8%	—	10%	—	8%	—	10%	—
		1	PA[14]	8%	11%	9%	10%	8%	11%	9%	10%
			PA[4]	7%	—	9%	—	7%	—	9%	—
			PA[13]	7%	10%	8%	9%	7%	10%	8%	9%
			PA[12]	7%	—	8%	—	7%	—	8%	—

Table 23. I/O weight¹ (continued)

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		² SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
3	2	2	PB[9]	1%	—	1%	—	1%	—	1%	—
			PB[8]	1%	—	1%	—	1%	—	1%	—
			PB[10]	5%	—	6%	—	6%	—	7%	—
		—	PF[0]	5%	—	6%	—	6%	—	8%	—
		—	PF[1]	5%	—	6%	—	7%	—	8%	—
		—	PF[2]	6%	—	7%	—	7%	—	9%	—
		—	PF[3]	6%	—	7%	—	8%	—	9%	—
		—	PF[4]	6%	—	7%	—	8%	—	10%	—
		—	PF[5]	6%	—	7%	—	9%	—	10%	—
		—	PF[6]	6%	—	7%	—	9%	—	11%	—
		—	PF[7]	6%	—	7%	—	9%	—	11%	—
	—	—	PJ[3]	6%	—	7%	—	—	—	—	—
	—	—	PJ[2]	6%	—	7%	—	—	—	—	—
	—	—	PJ[1]	6%	—	7%	—	—	—	—	—
	—	—	PJ[0]	6%	—	7%	—	—	—	—	—
	—	—	PI[15]	6%	—	7%	—	—	—	—	—
	—	—	PI[14]	6%	—	7%	—	—	—	—	—
	2	2	PD[0]	1%	—	1%	—	1%	—	1%	—
			PD[1]	1%	—	1%	—	1%	—	1%	—
			PD[2]	1%	—	1%	—	1%	—	1%	—
			PD[3]	1%	—	1%	—	1%	—	1%	—
			PD[4]	1%	—	1%	—	1%	—	1%	—
			PD[5]	1%	—	1%	—	1%	—	1%	—
			PD[6]	1%	—	1%	—	1%	—	2%	—
			PD[7]	1%	—	1%	—	1%	—	2%	—

Table 23. I/O weight¹ (continued)

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC ² = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
4	2	2	PD[8]	1%	—	1%	—	1%	—	2%	—
			PB[4]	1%	—	1%	—	1%	—	2%	—
			PB[5]	1%	—	1%	—	1%	—	2%	—
			PB[6]	1%	—	1%	—	1%	—	2%	—
			PB[7]	1%	—	1%	—	1%	—	2%	—
			PD[9]	1%	—	1%	—	1%	—	2%	—
			PD[10]	1%	—	1%	—	1%	—	2%	—
			PD[11]	1%	—	1%	—	1%	—	2%	—
4	—	—	PB[11]	1%	—	1%	—	—	—	—	—
	—	—	PD[12]	11%	—	13%	—	—	—	—	—
	2	2	PB[12]	11%	—	13%	—	15%	—	17%	—
			PD[13]	11%	—	13%	—	14%	—	17%	—
			PB[13]	11%	—	13%	—	14%	—	17%	—
			PD[14]	11%	—	13%	—	14%	—	17%	—
			PB[14]	11%	—	13%	—	14%	—	16%	—
			PD[15]	11%	—	13%	—	13%	—	16%	—
			PB[15]	11%	—	13%	—	13%	—	15%	—
	—	—	PI[8]	10%	—	12%	—	—	—	—	—
	—	—	PI[9]	10%	—	12%	—	—	—	—	—
	—	—	PI[10]	10%	—	12%	—	—	—	—	—
	—	—	PI[11]	10%	—	12%	—	—	—	—	—
	—	—	PI[12]	10%	—	12%	—	—	—	—	—
	—	—	PI[13]	10%	—	11%	—	—	—	—	—
	2	2	PA[3]	9%	—	11%	—	11%	—	13%	—
			PG[13]	9%	13%	11%	11%	10%	14%	12%	13%
			PG[12]	9%	13%	10%	11%	10%	14%	12%	12%
			PH[0]	6%	8%	7%	7%	6%	9%	7%	8%
			PH[1]	6%	8%	7%	7%	6%	8%	7%	7%
			PH[2]	5%	7%	6%	6%	5%	7%	6%	7%
			PH[3]	5%	7%	5%	6%	5%	7%	6%	6%
			PG[1]	4%	—	5%	—	4%	—	5%	—
	—	—	PG[0]	4%	5%	4%	5%	4%	5%	4%	5%

Table 23. I/O weight¹ (continued)

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC ² = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
5	3	—	PF[15]	4%	—	4%	—	4%	—	4%	—
		—	PF[14]	4%	6%	5%	5%	4%	6%	5%	5%
		—	PE[13]	4%	—	5%	—	4%	—	5%	—
		3	PA[7]	5%	—	6%	—	5%	—	6%	—
			PA[8]	5%	—	6%	—	5%	—	6%	—
			PA[9]	6%	—	7%	—	6%	—	7%	—
			PA[10]	6%	—	8%	—	6%	—	8%	—
			PA[11]	8%	—	9%	—	8%	—	9%	—
			PE[12]	8%	—	9%	—	8%	—	9%	—
			PG[14]	8%	—	9%	—	8%	—	9%	—
			PG[15]	8%	11%	9%	10%	8%	11%	9%	10%
			PE[14]	8%	—	9%	—	8%	—	9%	—
			PE[15]	8%	11%	9%	10%	8%	11%	9%	10%
			PG[10]	8%	—	9%	—	8%	—	9%	—
			PG[11]	7%	11%	9%	9%	7%	11%	9%	9%
		—	PH[11]	7%	10%	9%	9%	—	—	—	—
		—	PH[12]	7%	10%	8%	9%	—	—	—	—
		—	PI[5]	7%	—	8%	—	—	—	—	—
		—	PI[4]	7%	—	8%	—	—	—	—	—
	3	3	PC[3]	6%	—	8%	—	6%	—	8%	—
			PC[2]	6%	8%	7%	7%	6%	8%	7%	7%
			PA[5]	6%	8%	7%	7%	6%	8%	7%	7%
			PA[6]	5%	—	6%	—	5%	—	6%	—
			PH[10]	5%	7%	6%	6%	5%	7%	6%	6%
			PC[1]	5%	19%	5%	13%	5%	19%	5%	13%

Table 23. I/O weight¹ (continued)

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC ² = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
6	4	4	PC[0]	6%	9%	7%	8%	7%	10%	8%	8%
			PH[9]	7%	—	8%	—	7%	—	9%	—
			PE[2]	7%	10%	8%	9%	8%	11%	9%	10%
			PE[3]	7%	10%	9%	9%	8%	12%	10%	10%
			PC[5]	7%	11%	9%	9%	8%	12%	10%	11%
			PC[4]	8%	11%	9%	10%	9%	13%	10%	11%
			PE[4]	8%	11%	9%	10%	9%	13%	11%	12%
			PE[5]	8%	11%	10%	10%	9%	14%	11%	12%
		—	PH[4]	8%	12%	10%	10%	10%	14%	12%	12%
		—	PH[5]	8%	—	10%	—	10%	—	12%	—
		—	PH[6]	8%	12%	10%	11%	10%	15%	12%	13%
		—	PH[7]	9%	12%	10%	11%	11%	15%	13%	13%
		—	PH[8]	9%	12%	10%	11%	11%	16%	13%	14%
		4	PE[6]	9%	12%	10%	11%	11%	16%	13%	14%
			PE[7]	9%	12%	10%	11%	11%	16%	14%	14%
	—	—	PI[3]	9%	—	10%	—	—	—	—	—
	—	—	PI[2]	9%	—	10%	—	—	—	—	—
	—	—	PI[1]	9%	—	10%	—	—	—	—	—
	—	—	PI[0]	9%	—	10%	—	—	—	—	—
	4	4	PC[12]	8%	12%	10%	11%	12%	18%	15%	16%
			PC[13]	8%	—	10%	—	13%	—	15%	—
			PC[8]	8%	—	10%	—	13%	—	15%	—
			PB[2]	8%	11%	9%	10%	13%	18%	15%	16%

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

² SRC: "Slew Rate Control" bit in SIU_PCRx

4.6.6 RESET electrical characteristics

The device implements a dedicated bidirectional $\overline{\text{RESET}}$ pin.

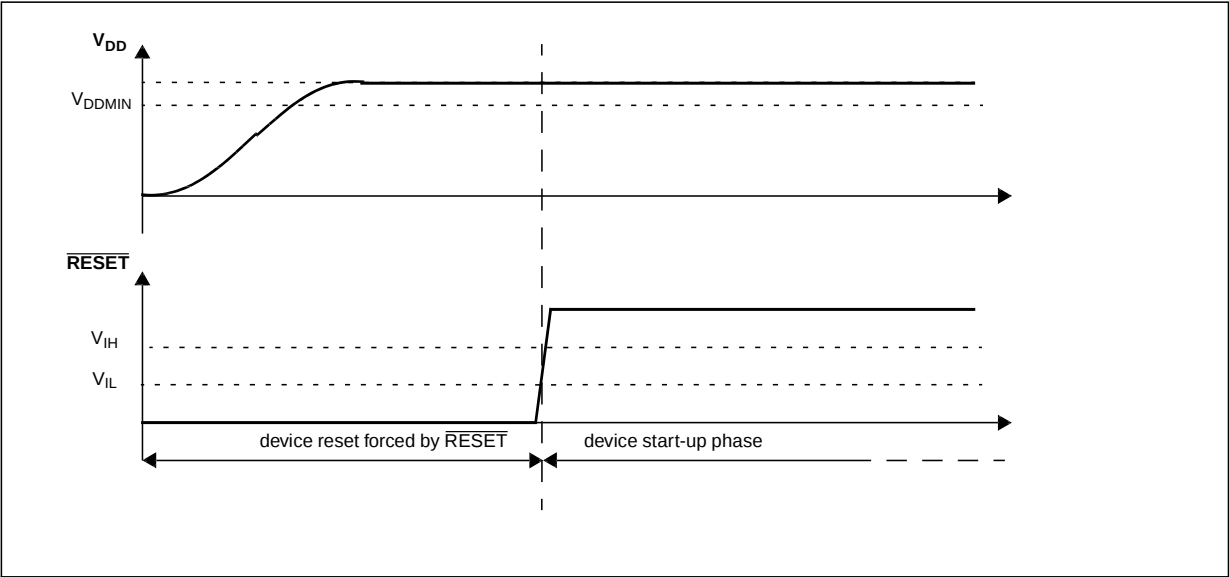


Figure 7. Start-up reset requirements

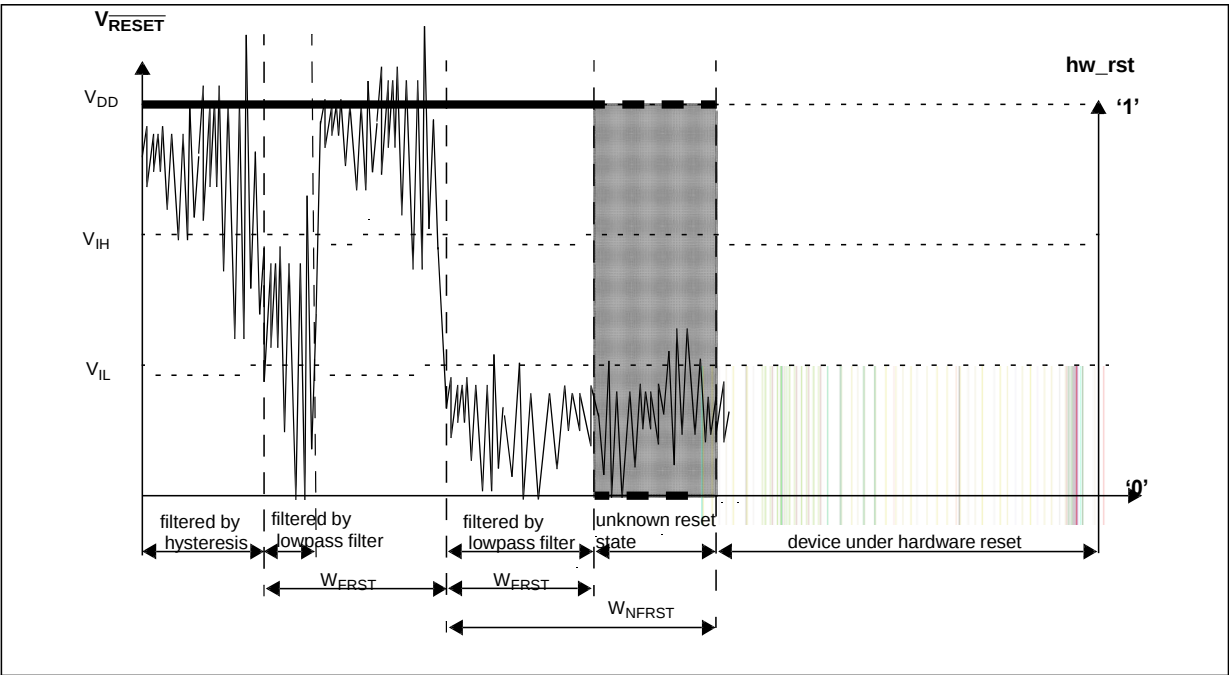


Figure 8. Noise filtering on reset signal

Table 24. Reset electrical characteristics

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
V_{IH}	SR	P	Input High Level CMOS (Schmitt Trigger)	—	—	$0.65V_{DD}$	V
V_{IL}	SR	P	Input low Level CMOS (Schmitt Trigger)	—	—	-0.4	V
V_{HYS}	CC	C	Input hysteresis CMOS (Schmitt Trigger)	—	—	$0.1V_{DD}$	V
V_{OL}	CC	P	Output low level	Push Pull, $I_{OL} = 2$ mA, $V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$
				Push Pull, $I_{OL} = 1$ mA, $V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 1 ²	—	—	$0.1V_{DD}$
				Push Pull, $I_{OL} = 1$ mA, $V_{DD} = 3.3$ V $\pm 10\%$, PAD3V5V = 1 (recommended)	—	—	0.5
t_{tr}	CC	D	Output transition time output pin ³ MEDIUM configuration	$C_L = 25$ pF, $V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 0	—	—	10
				$C_L = 50$ pF, $V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 0	—	—	20
				$C_L = 100$ pF, $V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 0	—	—	40
				$C_L = 25$ pF, $V_{DD} = 3.3$ V $\pm 10\%$, PAD3V5V = 1	—	—	12
				$C_L = 50$ pF, $V_{DD} = 3.3$ V $\pm 10\%$, PAD3V5V = 1	—	—	25
				$C_L = 100$ pF, $V_{DD} = 3.3$ V $\pm 10\%$, PAD3V5V = 1	—	—	40
W_{FRST}	SR	P	RESET input filtered pulse	—	—	40	ns
W_{NFRST}	SR	P	RESET input not filtered pulse	—	1000	—	ns
$ I_{WPUL} $	CC	P	Weak pull-up current absolute value	$V_{DD} = 3.3$ V $\pm 10\%$, PAD3V5V = 1	10	—	150
		D		$V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 0	10	—	150
		P		$V_{DD} = 5.0$ V $\pm 10\%$, PAD3V5V = 1 ⁴	10	—	250

¹ $V_{DD} = 3.3$ V $\pm 10\%$ / 5.0 V $\pm 10\%$, $T_A = -40$ to 125 °C, unless otherwise specified

² This is a transient configuration during power-up, up to the end of reset PHASE2 (refer to RGM module section of the device reference manual).

³ C_L includes device and package capacitance ($C_{PKG} < 5$ pF).

⁴ The configuration PAD3V5 = 1 when $V_{DD} = 5$ V is only transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

4.7 Power management electrical characteristics

4.7.1 Voltage regulator electrical characteristics

The device implements an internal voltage regulator to generate the low voltage core supply V_{DD_LV} from the high voltage ballast supply V_{DD_BV} . The regulator itself is supplied by the common I/O supply V_{DD} . The following supplies are involved:

- **HV:** High voltage external power supply for voltage regulator module. This must be provided externally through V_{DD} power pin.
- **BV:** High voltage external power supply for internal ballast module. This must be provided externally through V_{DD_BV} power pin. Voltage values should be aligned with V_{DD} .
- **LV:** Low voltage internal power supply for core, FMPLL and Flash digital logic. This is generated by the internal voltage regulator but provided outside to connect stability capacitor. It is further split into four main domains to ensure noise isolation between critical LV modules within the device:
 - **LV_COR:** Low voltage supply for the core. It is also used to provide supply for FMPLL through double bonding.
 - **LV_CFLA:** Low voltage supply for code flash module. It is supplied with dedicated ballast and shorted to LV_COR through double bonding.
 - **LV_DFLA:** Low voltage supply for data flash module. It is supplied with dedicated ballast and shorted to LV_COR through double bonding.
 - **LV_PLL:** Low voltage supply for FMPLL. It is shorted to LV_COR through double bonding.

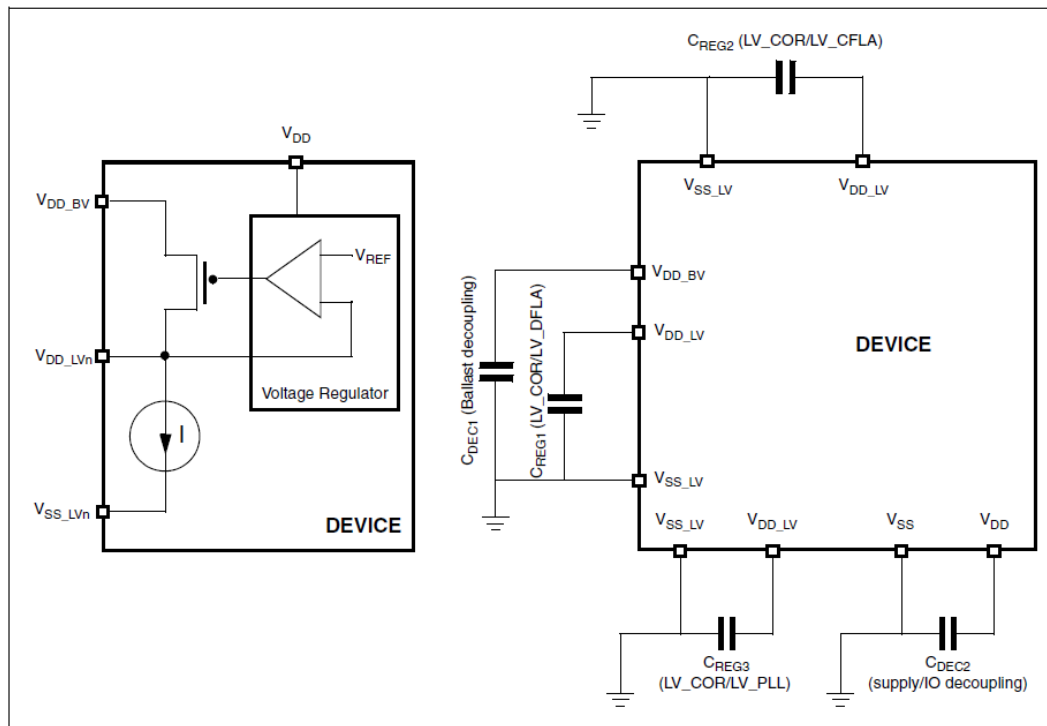


Figure 9. Voltage regulator capacitance connection

The internal voltage regulator requires external capacitance (C_{REGn}) to be connected to the device in order to provide a stable low voltage digital supply to the device. Capacitances should be placed on the board as near as possible to the associated pins. Care should also be taken to limit the serial inductance of the board to less than 5 nH.

Each decoupling capacitor must be placed between each of the three VDD_LV/VSS_LV supply pairs to ensure stable voltage (see 4.4, Recommended operating conditions).

Table 25. Voltage regulator electrical characteristics

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
C _{REGn}	SR	—	Internal voltage regulator external capacitance	200	—	500	nF
R _{REG}	SR	—	Stability capacitor equivalent serial resistance	—	—	0.2	Ω
C _{DEC1}	SR	—	Decoupling capacitance ² ballast	V _{DD_BV} /V _{SS_LV} pair: V _{DD_BV} = 4.5 V to 5.5 V	470 ⁴	—	nF
				V _{DD_BV} /V _{SS_LV} pair: V _{DD_BV} = 3 V to 3.6 V		—	
C _{DEC2}	SR	—	Decoupling capacitance regulator supply	V _{DD} /V _{SS} pair	10	100	nF
V _{MREG}	CC	T	Main regulator output voltage	Before exiting from reset	—	1.32	V
				After trimming	1.08	—	
I _{MREG}	SR	—	Main regulator current provided to V _{DD_LV} domain	—	—	150	mA
I _{MREGINT}	CC	D	Main regulator module current consumption	I _{MREG} = 200 mA	—	—	mA
				I _{MREG} = 0 mA	—	—	
I _{LPREGINT}	CC	D	Low-power regulator module current consumption	I _{LPREG} = 15 mA; T _A = 55 °C	—	—	μA
				I _{LPREG} = 0 mA; T _A = 55 °C	—	5	
V _{ULPREG}	CC	P	Ultra low power regulator output voltage	After trimming	1.16	1.28	V
I _{ULPREG}	SR	—	Ultra low power regulator current provided to V _{DD_LV} domain	—	—	5	mA
I _{ULPREGINT}	CC	D	Ultra low power regulator module current consumption	I _{ULPREG} = 5 mA; T _A = 55 °C	—	—	μA
				I _{ULPREG} = 0 mA; T _A = 55 °C	—	2	
I _{DD_BV}	CC	D	In-rush average current on V _{DD_BV} during power-up ⁵	—	—	300 ⁶	mA

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

² This capacitance value is driven by the constraints of the external voltage regulator supplying the V_{DD_BV} voltage. A typical value is in the range of 470 nF.

³ This value is acceptable to guarantee operation from 4.5 V to 5.5 V

⁴ External regulator and capacitance circuitry must be capable of providing I_{DD_BV} while maintaining supply V_{DD_BV} in operating range.

⁵ In-rush average current is seen only for short time during power-up and on standby exit (maximum 20 μs, depending on external capacitances to be loaded).

⁶ The duration of the in-rush current depends on the capacitance placed on LV pins. BV decoupling capacitors must be sized accordingly. Refer to I_{MREG} value for minimum amount of current to be provided in cc.

4.7.2 Low voltage detector electrical characteristics

The device implements a power-on reset (POR) module to ensure correct power-up initialization, as well as five low voltage detectors (LVDs) to monitor the V_{DD} and the V_{DD_LV} voltage while device is supplied:

- POR monitors V_{DD} during the power-up phase to ensure device is maintained in a safe reset state (refer to RGM Destructive Event Status (RGM_DES) Register flag F_POR in device reference manual)
- LVDHV3 monitors V_{DD} to ensure device reset below minimum functional supply (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD27 in device reference manual)
- LVDHV3B monitors V_{DD_BV} to ensure device reset below minimum functional supply (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD27_VREG in device reference manual)
- LVDHV5 monitors V_{DD} when application uses device in the $5.0\text{ V} \pm 10\%$ range (refer to RGM Functional Event Status (RGM_FES) Register flag F_LVD45 in device reference manual)
- LVDLVCOR monitors power domain No. 1 (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD12_PD1 in device reference manual)
- LVDLVBKP monitors power domain No. 0 (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD12_PD0 in device reference manual)

NOTE

When enabled, power domain No. 2 is monitored through LVDLVBKP.

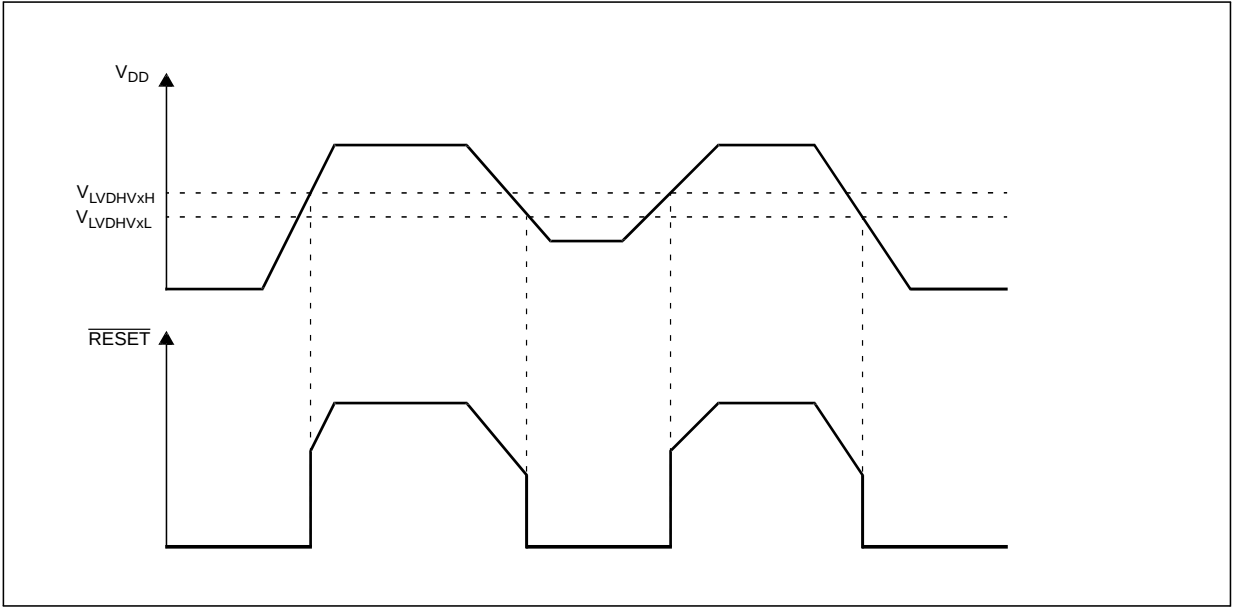


Figure 10. Low voltage detector vs reset

Table 26. Low voltage detector electrical characteristics

Symbol		C	Parameter	Conditions ¹	Value			Unit
					Min	Typ	Max	
V _{PORUP}	SR	P	Supply for functional POR module	T _A = 25 °C, after trimming	1.0	—	5.5	V
V _{PORH}	CC	P	Power-on reset threshold		1.5	—	2.6	
V _{LVDHV3H}	CC	T	LVDHV3 low voltage detector high threshold		—	—	2.95	
V _{LVDHV3L}	CC	P	LVDHV3 low voltage detector low threshold		2.6	—	2.9	
V _{LVDHV3BH}	CC	P	LVDHV3B low voltage detector high threshold		—	—	2.95	
V _{LVDHV3BL}	CC	P	LVDHV3B low voltage detector low threshold		2.6	—	2.9	
V _{LVDHV5H}	CC	T	LVDHV5 low voltage detector high threshold		—	—	4.5	
V _{LVDHV5L}	CC	P	LVDHV5 low voltage detector low threshold		3.8	—	4.4	
V _{LVDLVCORL}	CC	P	LVDLVCOR low voltage detector low threshold		1.08	—	1.16	
V _{LVDLVBKPL}	CC	P	LVDLVBKP low voltage detector low threshold		1.08	—	1.16	

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

4.8 Power consumption

Table 27 provides DC electrical characteristics for significant application modes. These values are indicative values; actual consumption depends on the application.

Table 27. Power consumption on VDD_BV and VDD_HV

Symbol	C	Parameter	Conditions ¹	Value			Unit		
				Min	Typ	Max			
I _{DDMAX} ²	CC	D	RUN mode maximum average current	—		—	115	140 ³	mA
I _{DDRUN} ⁴	CC	T	RUN mode typical average current ⁵	f _{CPU} = 8 MHz		—	12	—	mA
		f _{CPU} = 16 MHz		—	27	—			
		f _{CPU} = 32 MHz		—	43	—			
		f _{CPU} = 48 MHz		—	56	100			
		f _{CPU} = 64 MHz		—	70	125			
I _{DDHALT}	CC	C	HALT mode current ⁶	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	10	18	mA
		P			T _A = 125 °C	—	17	28	
I _{DDSTOP}	CC	P	STOP mode current ⁷	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	350	900 ⁸	μA
		D			T _A = 55 °C	—	750	—	
		D			T _A = 85 °C	—	2	7	mA
		D			T _A = 105 °C	—	4	10	
		P			T _A = 125 °C	—	7	14	

Table 27. Power consumption on VDD_BV and VDD_HV (continued)

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
I _{DDSTDBY2}	CC	P	STANDBY2 mode current ⁹	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	30	100	μA
		D			T _A = 55 °C	—	75	—	
		D			T _A = 85 °C	—	180	700	
		D			T _A = 105 °C	—	315	1000	
		P			T _A = 125 °C	—	560	1700	
I _{DDSTDBY1}	CC	T	STANDBY1 mode current ¹⁰	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	20	60	μA
		D			T _A = 55 °C	—	45	—	
		D			T _A = 85 °C	—	100	350	
		D			T _A = 105 °C	—	165	500	
		D			T _A = 125 °C	—	280	900	

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

² Running consumption does not include I/Os toggling which is highly dependent on the application. The given value is thought to be a worst case value with all peripherals running, and code fetched from code flash while modify operation ongoing on data flash. Notice that this value can be significantly reduced by application: switch off not used peripherals (default), reduce peripheral frequency through internal prescaler, fetch from RAM most used functions, use low power mode when possible.

³ Higher current may be sunk by device during power-up and standby exit. Please refer to in-rush average current in Table 25.

⁴ RUN current measured with typical application with accesses on both Flash and RAM.

⁵ Only for the "P" classification: Data and Code Flash in Normal Power. Code fetched from RAM: Serial IPs CAN and LIN in loop back mode, DSPI as Master, PLL as system clock (4 x Multiplier) peripherals on (eMIOS/CTU/ADC) and running at max frequency, periodic SW/WDG timer reset enabled.

⁶ Data Flash Power Down. Code Flash in Low Power. SIRC 128 kHz and FIRC 16 MHz on. 10 MHz XTAL clock. FlexCAN: instances: 0, 1, 2 ON (clocked but not reception or transmission), instances: 4, 5, 6 clocks gated. LINFlex: instances: 0, 1, 2 ON (clocked but not reception or transmission), instance: 3 to 9 clocks gated. eMIOS: instance: 0 ON (16 channels on PA[0]–PA[11] and PC[12]–PC[15]) with PWM 20 kHz, instance: 1 clock gated. DSPI: instance: 0 (clocked but no communication), instance: 1 to 5 clocks gated. RTC/API ON. PIT ON. STM ON. ADC1 OFF. ADC0 ON but no conversion except two analog watchdogs.

⁷ Only for the "P" classification: No clock, FIRC 16 MHz off, SIRC 128 kHz on, PLL off, HPVreg off, ULPVreg/LPVreg on. All possible peripherals off and clock gated. Flash in power down mode.

⁸ When going from RUN to STOP mode and the core consumption is > 6 mA, it is normal operation for the main regulator module to be kept on by the on-chip current monitoring circuit. This is most likely to occur with junction temperatures exceeding 125 °C and under these circumstances, it is possible for the current to initially exceed the maximum STOP specification by up to 2 mA. After entering stop, the application junction temperature will reduce to the ambient level and the main regulator will be automatically switched off when the load current is below 6 mA.

⁹ Only for the "P" classification: ULPreg on, HP/LPVreg off, 32 KB RAM on, device configured for minimum consumption, all possible modules switched off.

¹⁰ ULPreg on, HP/LPVreg off, 8 KB RAM on, device configured for minimum consumption, all possible modules switched off.

4.9 Flash memory electrical characteristics

4.9.1 Program/erase characteristics

Table 28 shows the program and erase characteristics.

Table 28. Program and erase specifications

Symbol		C	Parameter	Conditions	Value				Unit
					Min	Typ ¹	Initial max ²	Max ³	
t _{dwprogram}	CC	C	Double word (64 bits) program time ⁴	Code Flash	—	18	50	500	μs
				Data Flash		22			
t _{16Kpperase}		16 KB block preprogram and erase time	Code Flash	—	200	500	5000	ms	
			Data Flash		300				
t _{32Kpperase}		32 KB block preprogram and erase time	Code Flash	—	300	600	5000	ms	
			Data Flash		400				
t _{128Kpperase}		128 KB block preprogram and erase time	Code Flash	—	600	1300	7500	ms	
			Data Flash		800				
t _{esus}		D	Erase Suspend Latency	—	—	—	30	30	μs
t _{ESRT}		C	Erase Suspend Request Rate ⁵	Code Flash	20	—	—	—	ms
				Data Flash	10	—	—	—	

¹ Typical program and erase times assume nominal supply values and operation at 25 °C. All times are subject to change pending device characterization.

² Initial factory condition: < 100 program/erase cycles, 25 °C, typical supply voltage.

³ The maximum program and erase times occur after the specified number of program/erase cycles. These maximum values are characterized but not guaranteed.

⁴ Actual hardware programming times. This does not include software overhead.

⁵ Time between erase suspend resume and the next erase suspend request

Table 29. Flash module life

Symbol		C	Parameter	Conditions	Value			Unit
					Min	Typ	Max	
P/E	CC	C	—	Code Flash	10,000	—	—	cycles
				Data Flash	10,000		—	
Retention	CC	C	Minimum data retention at 85 °C average ambient temperature ¹	Blocks with 0–1,000 P/E cycles	20	—	—	years
				Blocks with 1,001–10,000 P/E cycles	10	—	—	years
				Blocks with 10,001–100,000 P/E cycles	5	—	—	years

¹ Ambient temperature averaged over duration of application, not to exceed recommended product operating temperature range.

ECC circuitry provides correction of single bit faults and is used to improve further automotive reliability results. Some units will experience single bit corrections throughout the life of the product with no impact to product reliability.

Table 30. Flash read access timing

Symbol		C	Parameter	Conditions ¹	Max	Unit
f _{READ}	CC	P	Maximum frequency for Flash reading	2 wait states	64	MHz
		C		2 wait state	40	
		C		2 wait states	20	

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

4.9.2 Flash power supply DC characteristics

Table 31 shows the power supply DC characteristics on external supply.

Table 31. Flash power supply DC electrical characteristics

Symbol		Parameter	Conditions ¹		Value			Unit
					Min	Typ	Max	
I _{CFREAD}	CC	Sum of the current consumption on V _{DD_HV} and V _{DD_BV} on read access	Flash module read f _{CPU} = 64 MHz	Code Flash	—	—	33	mA
I _{DFREAD}				Data Flash	—	—	33	
I _{CFMOD}	CC	Sum of the current consumption on V _{DD_HV} and V _{DD_BV} on matrix modification (program/erase)	Program/Erase on-going while reading Flash registers f _{CPU} = 64 MHz	Code Flash	—	—	52	mA
I _{DFMOD}				Data Flash	—	—	33	
I _{CFLPW}	CC	Sum of the current consumption on V _{DD_HV} and V _{DD_BV} during Flash low power mode	—	Code Flash	—	—	1.1	mA
I _{DFLPW}				Data Flash	—	—	900	
I _{CFPWD}	CC	Sum of the current consumption on V _{DD_HV} and V _{DD_BV} during Flash power down mode	—	Code Flash	—	—	150	μA
I _{DFPWD}				Data Flash	—	—	150	

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

4.9.3 Start-up/Switch-off timings

Table 32. Start-up time/Switch-off time

Symbol		C	Parameter	Conditions ¹	Value			Unit
					Min	Typ	Max	
t _{FLARSTEXIT}	CC	T	Delay for Flash module to exit reset mode	—	—	—	125	μs
t _{FLALPEXIT}	CC	T	Delay for Flash module to exit low-power mode	—	—	—	0.5	
t _{FLAPDEXIT}	CC	T	Delay for Flash module to exit power-down mode	—	—	—	30	
t _{FLALPENTRY}	CC	T	Delay for Flash module to enter low-power mode	—	—	—	0.5	
t _{FLAPDENTRY}	CC	T	Delay for Flash module to enter power-down mode	—	—	—	1.5	

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

4.10 Electromagnetic compatibility (EMC) characteristics

Susceptibility tests are performed on a sample basis during product characterization.

4.10.1 Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore, it is recommended that the user apply EMC software optimization and prequalification tests in relation with the EMC level requested for the application.

- Software recommendations – The software flowchart must include the management of runaway conditions such as:
 - Corrupted program counter
 - Unexpected reset

— Critical data corruption (control registers...)

- Prequalification trials – Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the reset pin or the oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring.

4.10.2 Electromagnetic interference (EMI)

The product is monitored in terms of emission based on a typical application. This emission test conforms to the IEC61967-1 standard, which specifies the general conditions for EMI measurements.

Table 33. EMI radiated emission measurement^{1,2}

Symbol	C	Parameter	Conditions	Value			Unit
				Min	Typ	Max	
—	SR	—	Scan range	0.150		1000	MHz
f _{CPU}	SR	—	Operating frequency	—	64	—	MHz
V _{DD_LV}	SR	—	LV operating voltages	—	1.1	—	V
S _{EMI}	CC	T	Peak level V _{DD} = 5 V, T _A = 25 °C, LQFP144 package Test conforming to IEC 61967-2, f _{OSC} = 8 MHz/f _{CPU} = 64 MHz	No PLL frequency modulation	—	—	18 dBμV
				± 2% PLL frequency modulation	—	—	14 dBμV

¹ EMI testing and I/O port waveforms per IEC 61967-1, -2, -4

² For information on conducted emission and susceptibility measurement (norm IEC 61967-4), please contact your local marketing representative.

4.10.3 Absolute maximum ratings (electrical sensitivity)

Based on two different tests (ESD and LU) using specific measurement methods, the product is stressed in order to determine its performance in terms of electrical sensitivity.

4.10.3.1 Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts×(n + 1) supply pin). This test conforms to the AEC-Q100-002/-003/-011 standard.

Table 34. ESD absolute maximum ratings^{1,2}

Symbol	Ratings	Conditions	Class	Max value ³	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (Human Body Model)	$T_A = 25\text{ }^{\circ}\text{C}$ conforming to AEC-Q100-002	H1C	2000	V
$V_{ESD(MM)}$	Electrostatic discharge voltage (Machine Model)	$T_A = 25\text{ }^{\circ}\text{C}$ conforming to AEC-Q100-003	M2	200	
$V_{ESD(CDM)}$	Electrostatic discharge voltage (Charged Device Model)	$T_A = 25\text{ }^{\circ}\text{C}$ conforming to AEC-Q100-011	C3A	500 750 (corners)	

¹ All ESD testing is in conformity with CDF-AEC-Q100 Stress Test Qualification for Automotive Grade Integrated Circuits.

² A device will be defined as a failure if after exposure to ESD pulses the device no longer meets the device specification requirements. Complete DC parametric and functional testing shall be performed per applicable device specification at room temperature followed by hot temperature, unless specified otherwise in the device specification.

³ Data based on characterization results, not tested in production

4.10.3.2 Static latch-up (LU)

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin. These tests are compliant with the EIA/JESD 78 IC latch-up standard.

Table 35. Latch-up results

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	$T_A = 125\text{ }^{\circ}\text{C}$ conforming to JESD 78	II level A

4.11 Fast external crystal oscillator (8 to 20 MHz) electrical characteristics

The device provides an oscillator/resonator driver. [Figure 11](#) describes a simple model of the internal oscillator driver and provides an example of a connection for an oscillator or a resonator.

[Table 36](#) provides the parameter description of 8 MHz to 20 MHz crystals used for the design simulations.

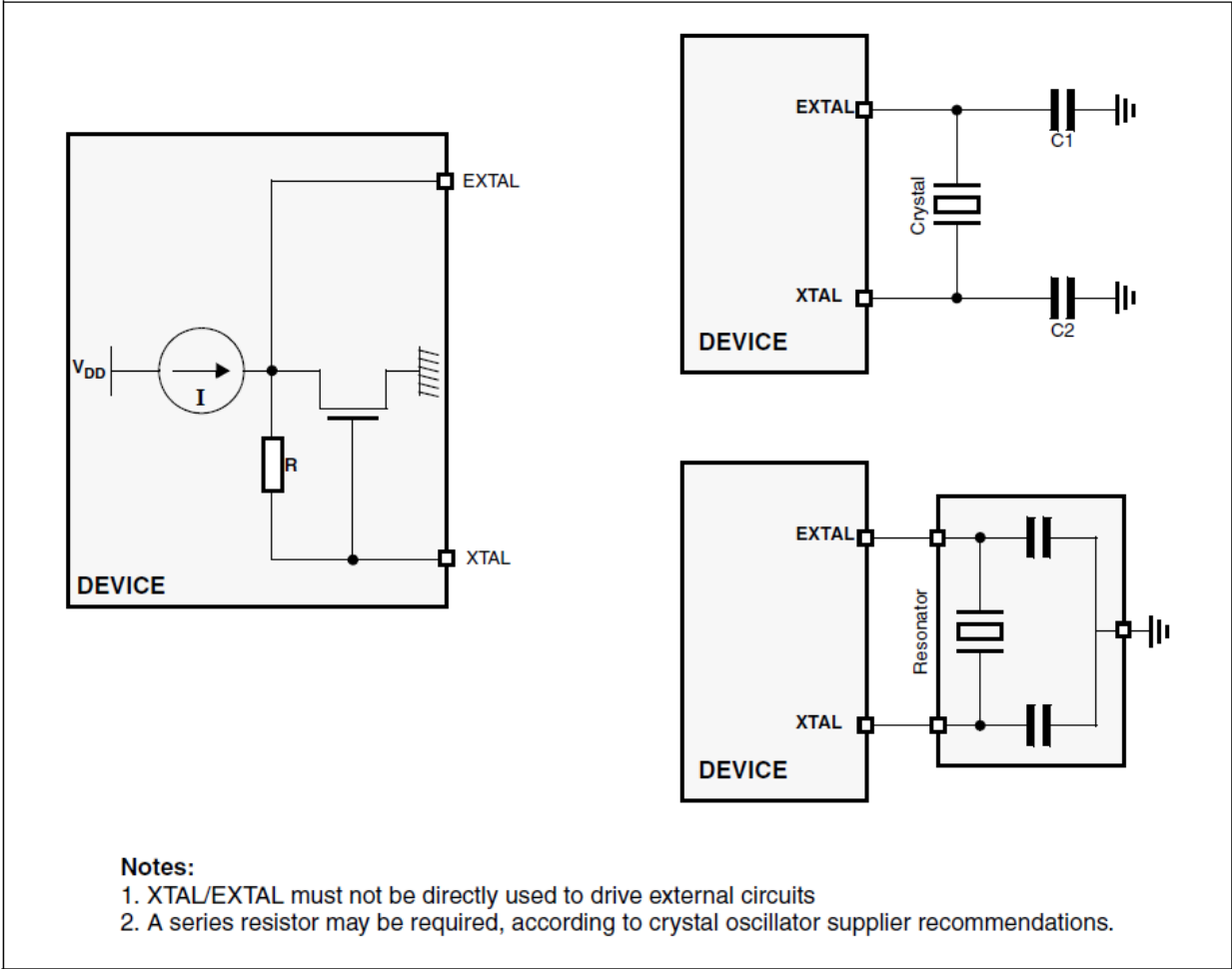


Figure 11. Crystal oscillator and resonator connection scheme

Table 36. Crystal description

Nominal frequency (MHz)	NDK crystal reference	Crystal equivalent series resistance ESR (Ω)	Crystal motional capacitance (C _m) fF	Crystal motional inductance (L _m) mH	Load on xtal _{in} /xtal _{out} C1 = C2 ¹	Shunt capacitance between xtal _{out} and xtal _{in} C0 ² (pF)
8	NX5032GA	300	2.46	160.7	17	3.01
10		150	2.93	86.6	15	2.91
12		120	3.11	56.5	15	2.93
16		120	3.90	25.3	10	3.00

¹ The values specified for C1 and C2 are the same as used in simulations. It should be ensured that the testing includes all the parasitics (from the board, probe, crystal, etc.) as the AC / transient behavior depends upon them.

² The value of C0 specified here includes 2 pF additional capacitance for parasitics (to be seen with bond-pads, package, etc.).

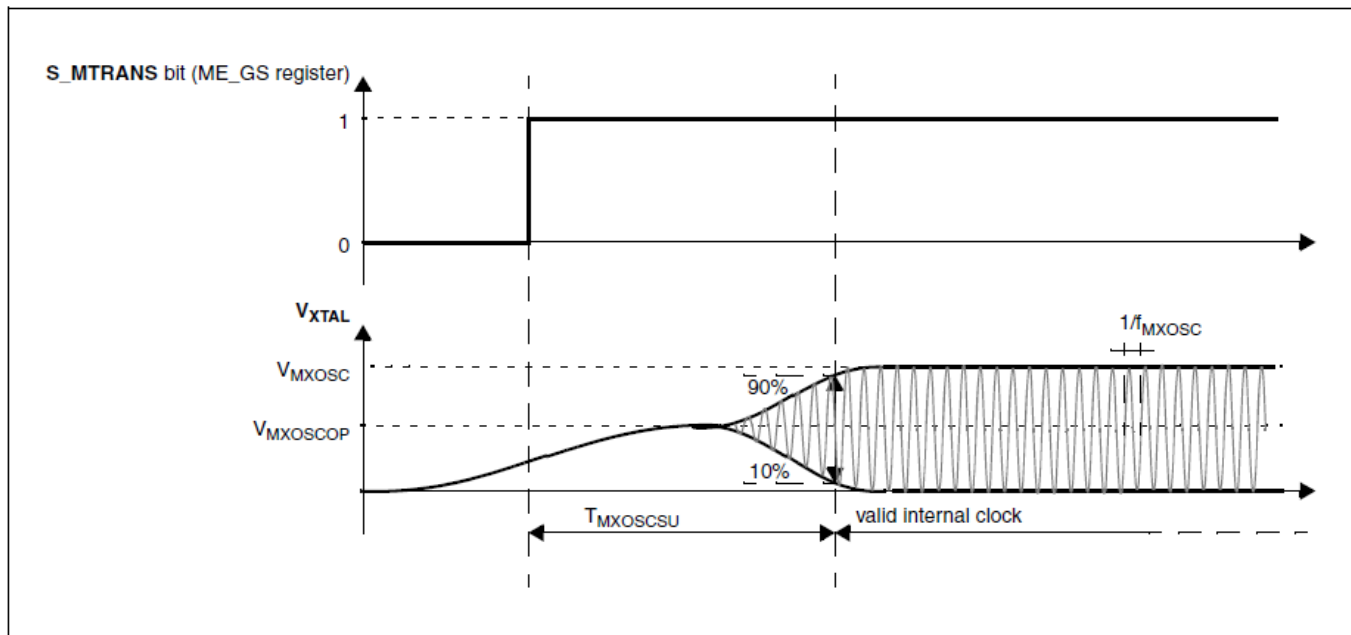


Figure 12. Fast external crystal oscillator (8 to 20 MHz) timing diagram

Table 37. Fast external crystal oscillator (8 to 20 MHz) electrical characteristics

Symbol		C	Parameter	Conditions ¹	Value			Unit
					Min	Typ	Max	
f _{FXOSC}	SR	—	Fast external crystal oscillator frequency	—	8.0	—	20.0	MHz
g _{mFXOSC}	CC	C	Fast external crystal oscillator transconductance	V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 OSCILLATOR_MARGIN = 0	2.2	—	8.2	mA/V
	CC	P		V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 OSCILLATOR_MARGIN = 0	2.0	—	7.4	
	CC	C		V _{DD} = 3.3 V ± 10%, PAD3V5V = 1 OSCILLATOR_MARGIN = 1	2.7	—	9.7	
	CC	C		V _{DD} = 5.0 V ± 10%, PAD3V5V = 0 OSCILLATOR_MARGIN = 1	2.5	—	9.2	
V _{FXOSC}	CC	T	Oscillation amplitude at XTAL	f _{OSC} = 8 MHz, OSCILLATOR_MARGIN = 0	1.3	—	—	V
				f _{OSC} = 20 MHz, OSCILLATOR_MARGIN = 1	1.3	—	—	
V _{FXOSCOP}	CC	C	Oscillation operating point	—	—	0.95	—	V
I _{FXOSC} ²	CC	T	Fast external crystal oscillator consumption	—	—	2	3	mA

Table 37. Fast external crystal oscillator (8 to 20 MHz) electrical characteristics (continued)

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
t_{FXOSCSU}	CC	T	Fast external crystal oscillator start-up time	—	—	4	ms
V_{IH}	SR	P	Input high level CMOS (Schmitt Trigger)	$0.65V_{\text{DD}}$	—	$V_{\text{DD}} + 0.4$	V
V_{IL}	SR	P	Input low level CMOS (Schmitt Trigger)	-0.4	—	$0.35V_{\text{DD}}$	V

¹ $V_{\text{DD}} = 3.3 \text{ V} \pm 10\% / 5.0 \text{ V} \pm 10\%$, $T_{\text{A}} = -40$ to 125 °C, unless otherwise specified.

² Stated values take into account only analog module consumption but not the digital contributor (clock tree and enabled peripherals).

³ The default start-up time of external crystal oscillator is 4ms which can be configured by software if the developers need.

4.12 Slow external crystal oscillator (32 kHz) electrical characteristics

The device provides a low power oscillator/resonator driver.

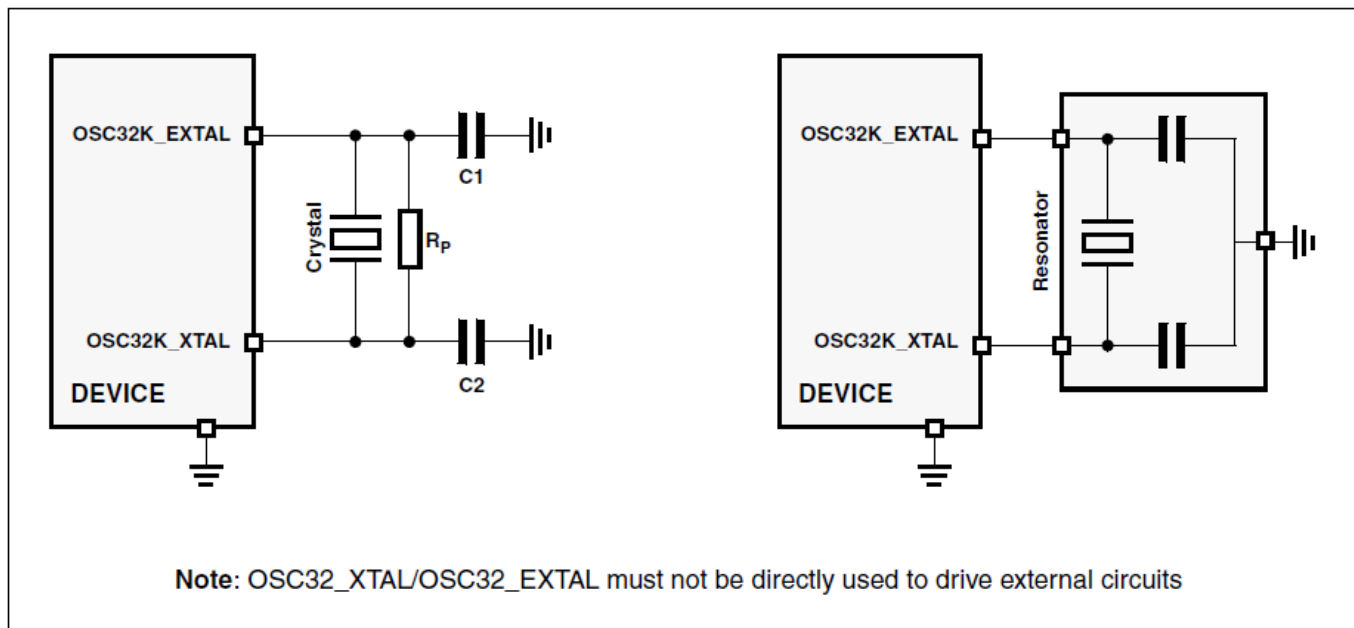


Figure 13. Crystal oscillator and resonator connection scheme

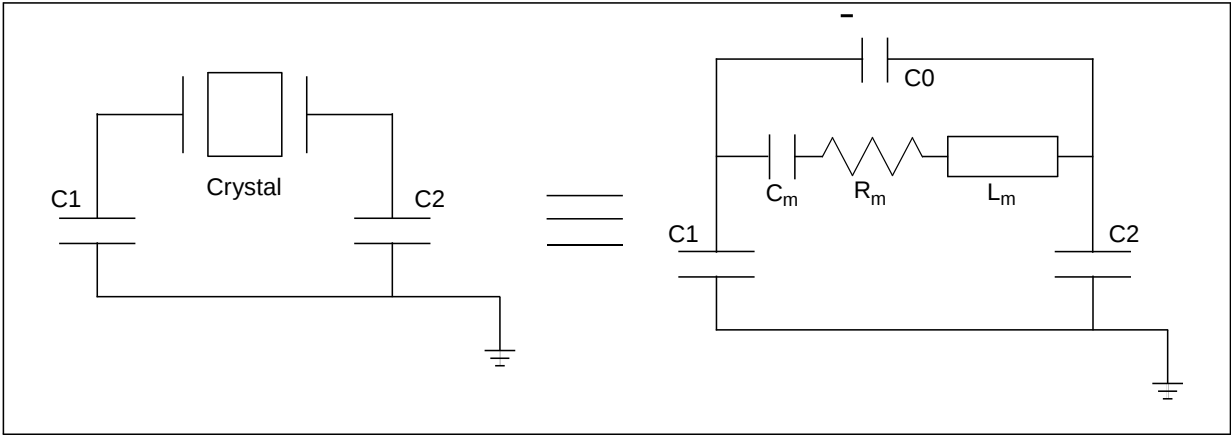


Figure 14. Equivalent circuit of a quartz crystal

Table 38. Crystal motional characteristics¹

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
L_m	Motional inductance	—	—	11.796	—	KH
C_m	Motional capacitance	—	—	2	—	fF
C1/C2	Load capacitance at OSC32K_XTAL and OSC32K_EXTAL with respect to ground ²	—	18	—	28	pF
R_m ³	Motional resistance	AC coupled at C0 = 2.85 pF ⁴	—	—	65	kΩ
		AC coupled at C0 = 4.9 pF ⁴	—	—	50	
		AC coupled at C0 = 7.0 pF ⁴	—	—	35	
		AC coupled at C0 = 9.0 pF ⁴	—	—	30	

¹ The crystal used is Epson Toyocom MC306.

² This is the recommended range of load capacitance at OSC32K_XTAL and OSC32K_EXTAL with respect to ground. It includes all the parasitics due to board traces, crystal and package.

³ Maximum ESR (R_m) of the crystal is 50 kΩ

⁴ C0 Includes a parasitic capacitance of 2.0 pF between OSC32K_XTAL and OSC32K_EXTAL pins.

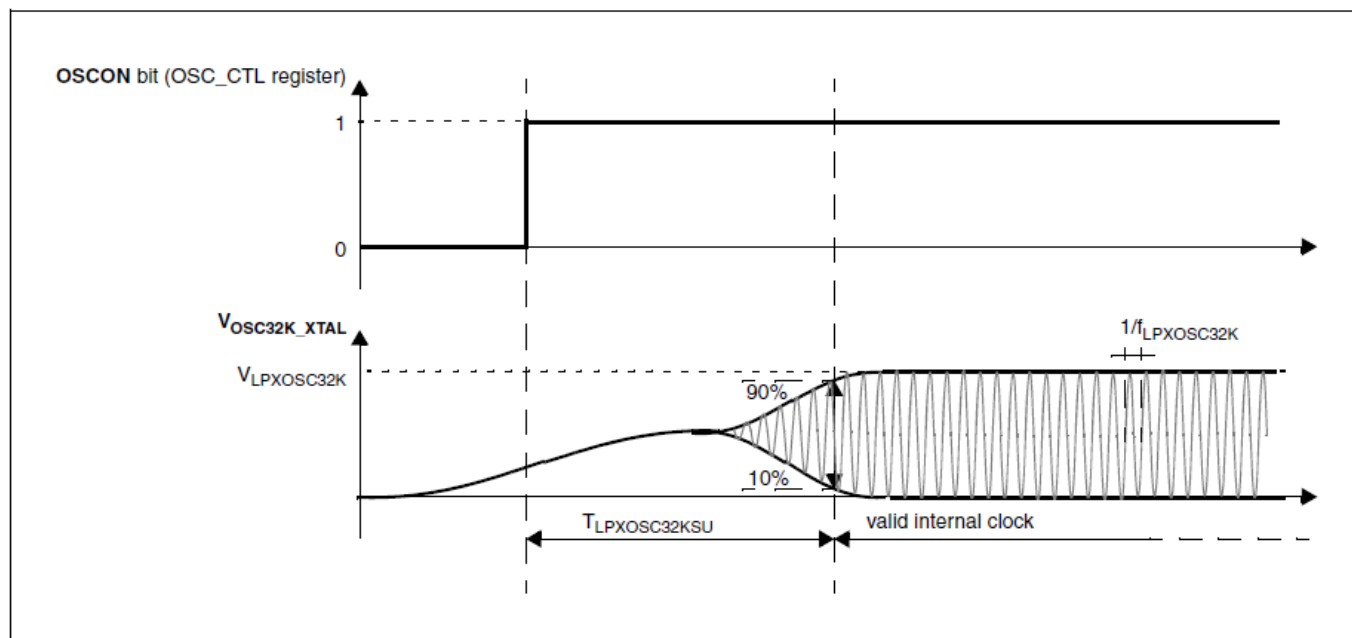


Figure 15. Slow external crystal oscillator (32 kHz) timing diagram

Table 39. Slow external crystal oscillator (32 kHz) electrical characteristics

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
f_{SXOSC}	SR	—	Slow external crystal oscillator frequency	32	32.768	40	kHz
V_{SXOSC}	CC	T	Oscillation amplitude	—	2.1	—	V
$I_{SXOSCBIAS}$	CC	T	Oscillation bias current	—	2.5	—	μA
I_{SXOSC}	CC	T	Slow external crystal oscillator consumption	—	—	8	μA
$t_{SXOSCSU}$	CC	T	Slow external crystal oscillator start-up time	—	—	2 ²	s

¹ $V_{DD} = 3.3 V \pm 10\% / 5.0 V \pm 10\%$, $T_A = -40$ to $125^\circ C$, unless otherwise specified. Values are specified for no neighbor GPIO pin activity. If oscillator is enabled (OSC32K_XTAL and OSC32K_EXTAL pins), neighboring pins should not toggle.

² Start-up time has been measured with EPSON TOYOCOM MC306 crystal. Variation may be seen with other crystal.

4.13 FMPLL electrical characteristics

The device provides a frequency modulated phase locked loop (FMPLL) module to generate a fast system clock from the main oscillator driver.

Table 40. FMPLL electrical characteristics

Symbol		C	Parameter	Conditions ¹	Value			Unit
					Min	Typ	Max	
f _{PLLIN}	SR	—	FMPLL reference clock ²	—	4	—	64	MHz
Δ _{PLLIN}	SR	—	FMPLL reference clock duty cycle ²	—	40	—	60	%
f _{PLLOUT}	CC	P	FMPLL output clock frequency	—	16	—	120	MHz
f _{VCO} ³	CC	P	VCO frequency without frequency modulation	—	256	—	512	MHz
		P	VCO frequency with frequency modulation	—	245.76	—	532.48	
f _{CPU}	SR	—	System clock frequency	—	—	—	120	MHz
f _{FREE}	CC	P	Free-running frequency	—	20	—	150	MHz
t _{LOCK}	CC	P	FMPLL lock time	Stable oscillator (f _{PLLIN} = 16 MHz)		40	150	μs
Δt _{STJIT}	CC	—	FMPLL short term jitter ⁴	f _{sys} maximum	−4	—	4	%
Δt _{LTJIT}	CC	—	FMPLL long term jitter	f _{PLLCLK} at 64 MHz, 4000 cycles	—	—	10	ns
I _{PLL}	CC	C	FMPLL consumption	T _A = 25 °C	—	—	4	mA

¹ $V_{DD} = 3.3\text{ V} \pm 10\% / 5.0\text{ V} \pm 10\%$, $T_A = -40$ to $125\text{ }^{\circ}\text{C}$, unless otherwise specified

² PLLIN clock retrieved directly from FXOSC clock. Input characteristics are granted when oscillator is used in functional mode. When bypass mode is used, oscillator input clock should verify f_{PLLIN} and Δ_{PLLIN} .

³ Frequency modulation is considered $\pm 4\%$.

⁴ Short term jitter is measured on the clock rising edge at cycle n and n+4.

4.14 Fast internal RC oscillator (16 MHz) electrical characteristics

The device provides a 16 MHz main internal RC oscillator. This is used as the default clock at the power-up of the device.

Table 41. Fast internal RC oscillator (16 MHz) electrical characteristics

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
f_{FIRC}	CC	P	Fast internal RC oscillator high frequency	$T_A = 25\text{ °C}$, trimmed			MHz
	SR			—	16	—	
$I_{FIRC RUN}^{2,}$	CC	T	Fast internal RC oscillator high frequency current in running mode	$T_A = 25\text{ °C}$, trimmed	—	200	μA
$I_{FIRC PWD}$	CC	D	Fast internal RC oscillator high frequency current in power down mode	$T_A = 25\text{ °C}$	—	10	μA

Table 41. Fast internal RC oscillator (16 MHz) electrical characteristics (continued)

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
I _{FIRCSTOP}	CC	T	Fast internal RC oscillator high frequency and system clock current in stop mode	T _A = 25 °C	sysclk = off	—	500	—	μA
					sysclk = 2 MHz	—	600	—	
					sysclk = 4 MHz	—	700	—	
					sysclk = 8 MHz	—	900	—	
					sysclk = 16 MHz	—	1250	—	
t _{FIRCSU}	CC	C	Fast internal RC oscillator start-up time	V _{DD} = 5.0 V ± 10%		—	1.1	2.0	μs
Δ _{FIRCPRE}	CC	C	Fast internal RC oscillator precision after software trimming of f _{FIRC}	T _A = 25 °C		–1	—	1	%
Δ _{FIRCTRIM}	CC	C	Fast internal RC oscillator trimming step	T _A = 25 °C		—	1.6		%
Δ _{FIRCVAR}	CC	C	Fast internal RC oscillator variation over temperature and supply with respect to f _{FIRC} at T _A = 25 °C in high-frequency configuration	—		–5	—	5	%

¹ $V_{DD} = 3.3\text{ V} \pm 10\% / 5.0\text{ V} \pm 10\%$, $T_A = -40\text{ to }125\text{ °C}$, unless otherwise specified

² This does not include consumption linked to clock tree toggling and peripherals consumption when RC oscillator is ON.

4.15 Slow internal RC oscillator (128 kHz) electrical characteristics

The device provides a 128 kHz low power internal RC oscillator. This can be used as the reference clock for the RTC module.

Table 42. Slow internal RC oscillator (128 kHz) electrical characteristics

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
f _{SIRC}	CC P	Slow internal RC oscillator low frequency	T _A = 25 °C, trimmed	—	128	—	kHz
	SR —		—	100	—	150	
I _{SIRC} ²	CC C	Slow internal RC oscillator low frequency current	T _A = 25 °C, trimmed	—	—	5	μA
t _{SIRCSU}	CC P	Slow internal RC oscillator start-up time	T _A = 25 °C, V _{DD} = 5.0 V ± 10%	—	8	12	μs
Δ _{SIRCPRE}	CC C	Slow internal RC oscillator precision after software trimming of f _{SIRC}	T _A = 25 °C	-2	—	2	%
Δ _{SIRCTRM}	CC C	Slow internal RC oscillator trimming step	—	—	2.7	—	
Δ _{SIRCVAR}	CC C	Slow internal RC oscillator variation in temperature and supply with respect to f _{SIRC} at T _A = 55 °C in high frequency configuration	High frequency configuration	-10	—	10	%

¹ V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified

² This does not include consumption linked to clock tree toggling and peripherals consumption when RC oscillator is ON.

4.16 ADC electrical characteristics

4.16.1 Introduction

The device provides two Successive Approximation Register (SAR) analog-to-digital converters (10-bit and 12-bit).

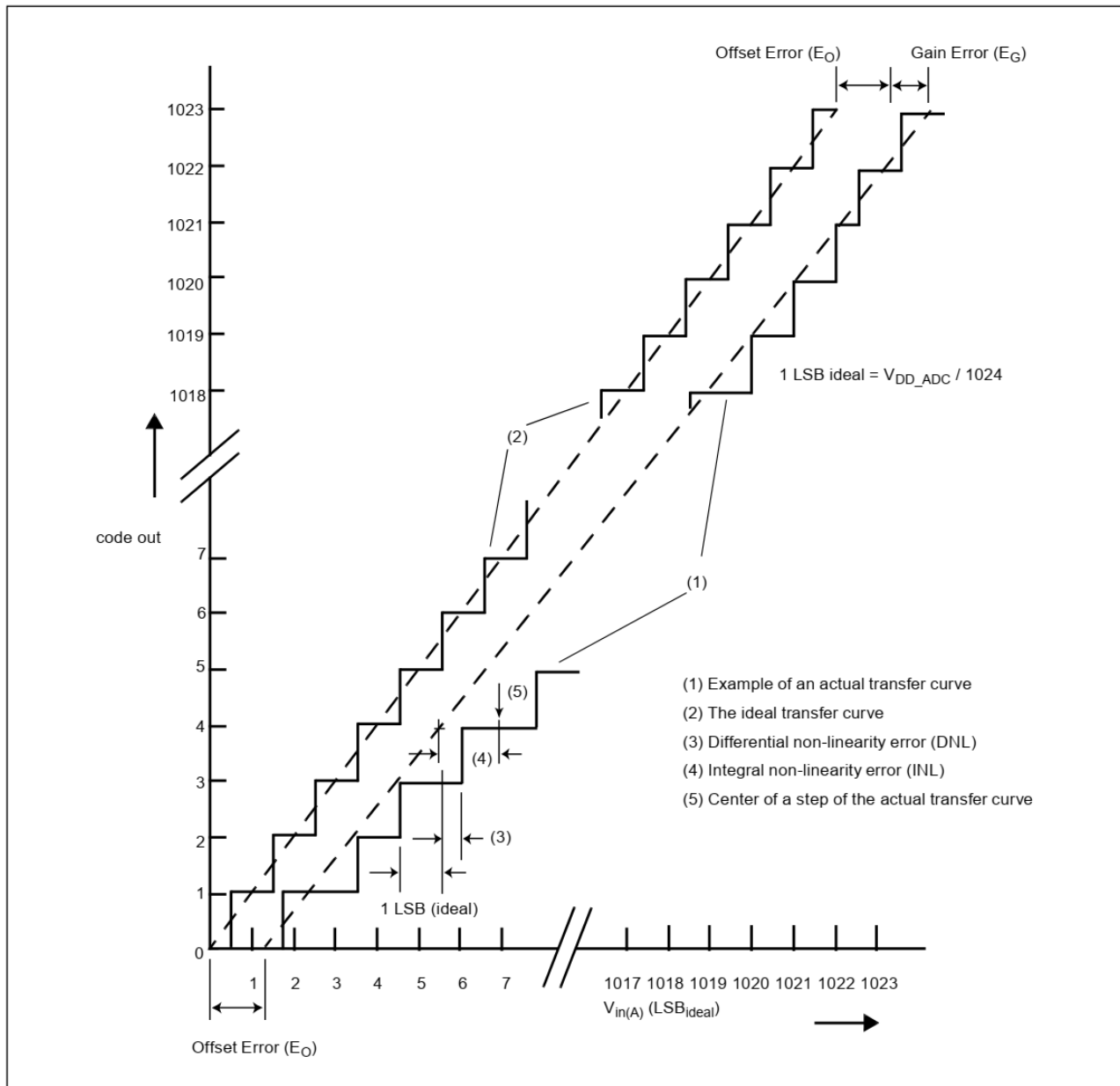


Figure 16. ADC_0 characteristic and error definitions

4.16.2 Input impedance and ADC accuracy

In the following analysis, the input circuit corresponding to the precise channels is considered.

To preserve the accuracy of the A/D converter, it is necessary that analog input pins have low AC impedance. Placing a capacitor with good high frequency characteristics at the input pin of the device can be effective: the capacitor should be as large as possible, ideally infinite. This capacitor contributes to attenuating the noise present on the input pin; furthermore, it sources charge during the sampling phase, when the analog signal source is a high-impedance source.

A real filter can typically be obtained by using a series resistance with a capacitor on the input pin (simple RC filter). The RC filtering may be limited according to the value of source impedance of the transducer or circuit supplying the analog signal to be measured. The filter at the input pins must be designed taking into account the dynamic characteristics of the input signal (bandwidth) and the equivalent input impedance of the ADC itself.

In fact a current sink contributor is represented by the charge sharing effects with the sampling capacitance: being C_S and C_{p2} substantially two switched capacitances, with a frequency equal to the conversion rate of the ADC, it can be seen as a

resistive path to ground. For instance, assuming a conversion rate of 1 MHz, with $C_S + C_{P2}$ equal to 3 pF, a resistance of 330 k Ω is obtained ($R_{EQ} = 1 / (f_c \times (C_S + C_{P2}))$), where f_c represents the conversion rate at the considered channel). To minimize the error induced by the voltage partitioning between this resistance (sampled voltage on $C_S + C_{P2}$) and the sum of $R_S + R_F$, the external circuit must be designed to respect the Equation 4:

$$V_A \cdot \frac{R_S + R_F}{R_{EQ}} < \frac{1}{2} LSB \quad \text{Eqn. 4}$$

Equation 4 generates a constraint for external network design, in particular on a resistive path.

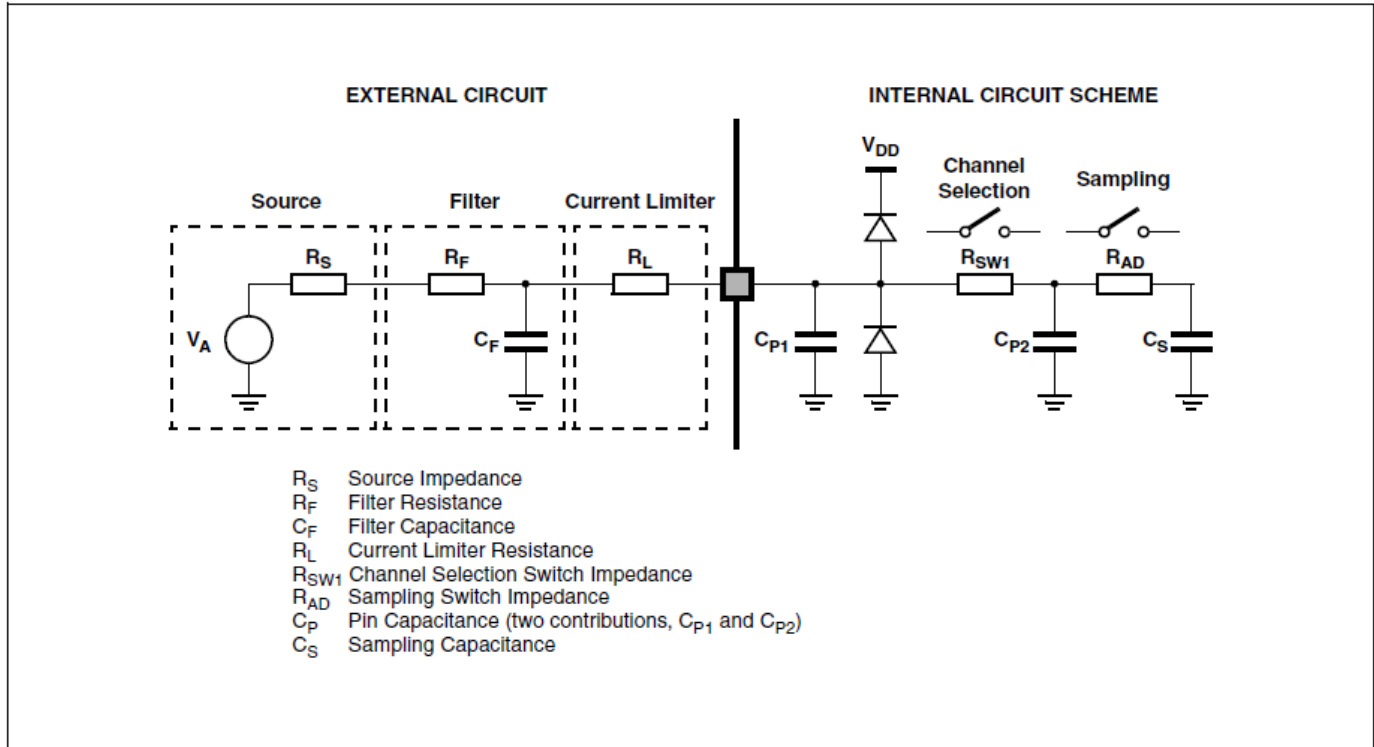


Figure 17. Input equivalent circuit (precise channels)

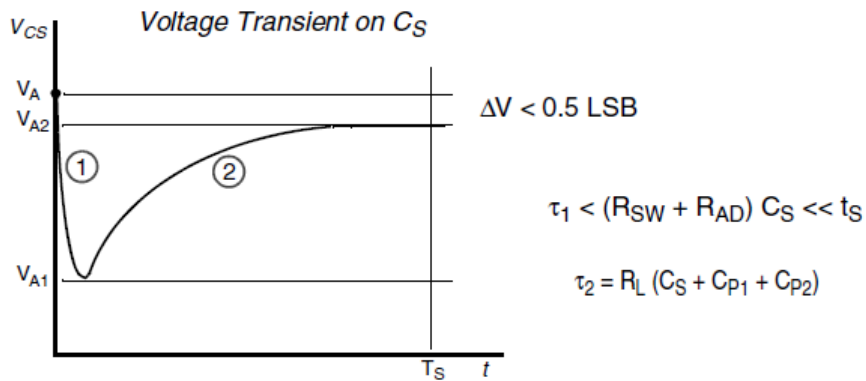
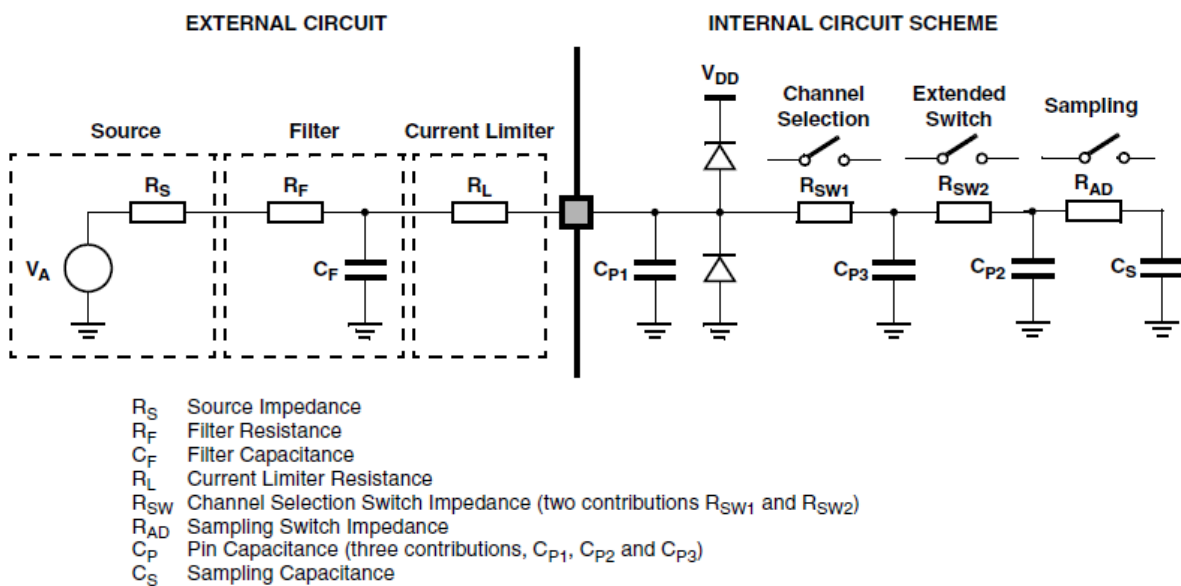


Figure 18. Input equivalent circuit (extended channels)

A second aspect involving the capacitance network shall be considered. Assuming the three capacitances C_F , C_{P1} and C_{P2} are initially charged at the source voltage V_A (refer to the equivalent circuit reported in Figure 17): A charge sharing phenomenon is installed when the sampling phase is started (A/D switch close).

Figure 19. Transient behavior during sampling phase



In particular two different transient periods can be distinguished:

1. A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

$$\tau_1 = (R_{SW} + R_{AD}) \cdot \frac{C_P \cdot C_S}{C_P + C_S}$$

Eqn. 5

Equation 5 can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time t_s is always much longer than the internal time constant:

Eqn. 6

$$\tau_1 < (R_{SW} + R_{AD}) \cdot C_S \ll t_s$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S , determining a new value of the voltage V_{A1} on the capacitance according to Equation 7:

Eqn. 7

$$V_{A1} \cdot (C_S + C_{P1} + C_{P2}) = V_A \cdot (C_{P1} + C_{P2})$$

2. A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L : again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Eqn. 8

$$\tau_2 < R_L \cdot (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time t_s , a constraints on R_L sizing is obtained:

ADC_0 (10-bit)**Eqn. 9**

$$8.5 \cdot \tau_2 = 8.5 \cdot R_L \cdot (C_S + C_{P1} + C_{P2}) < t_s$$

ADC_1 (12-bit)**Eqn. 10**

$$10 \cdot \tau_2 = 10 \cdot R_L \cdot (C_S + C_{P1} + C_{P2}) < t_s$$

Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1} , C_{P2} and C_S , then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1} . Equation 11 must be respected (charge balance assuming now C_S already charged at V_{A1}):

Eqn. 11

$$V_{A2} \cdot (C_S + C_{P1} + C_{P2} + C_F) = V_A \cdot C_F + V_{A1} \cdot (C_{P1} + C_{P2} + C_S)$$

The two transients above are not influenced by the voltage source that, due to the presence of the $R_F C_F$ filter, is not able to provide the extra charge to compensate the voltage drop on C_S with respect to the ideal source V_A ; the time constant $R_F C_F$ of the filter is very high with respect to the sampling time (t_s). The filter is typically designed to act as antialiasing.

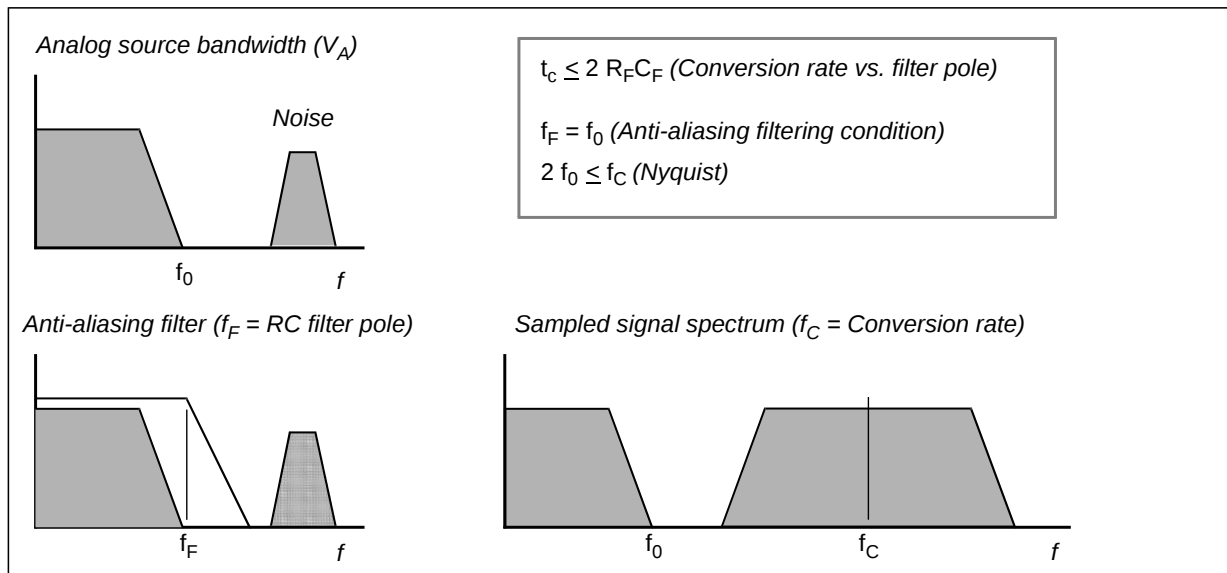


Figure 20. Spectral representation of input signal

Calling f_0 the bandwidth of the source signal (and as a consequence the cut-off frequency of the antialiasing filter, f_F), according to the Nyquist theorem the conversion rate f_C must be at least $2f_0$; it means that the constant time of the filter is greater than or at least equal to twice the conversion period (t_c). Again the conversion period t_c is longer than the sampling time t_s , which is just a portion of it, even when fixed channel continuous conversion mode is selected (fastest conversion rate at a specific channel): in

conclusion it is evident that the time constant of the filter $R_F C_F$ is definitively much higher than the sampling time t_s , so the charge level on C_S cannot be modified by the analog signal source during the time in which the sampling switch is closed.

The considerations above lead to impose new constraints on the external circuit, to reduce the accuracy error due to the voltage drop on C_S ; from the two charge balance equations above, it is simple to derive [Equation 12](#) between the ideal and real sampled voltage on C_S :

Eqn. 12

$$\frac{V_{A2}}{V_A} = \frac{C_{P1} + C_{P2} + C_F}{C_{P1} + C_{P2} + C_F + C_S}$$

From this formula, in the worst case (when V_A is maximum, that is for instance 5 V), assuming to accept a maximum error of half a count, a constraint is evident on C_F value:

ADC_0 (10-bit)

Eqn. 13

$$C_F > 2048 \cdot C_S$$

ADC_1 (12-bit)

Eqn. 14

$$C_F > 8192 \cdot C_S$$

4.16.3 ADC electrical characteristics

Table 43. ADC input leakage current

Symbol	C	Parameter	Conditions	Value			Unit
				Min	Typ	Max	
I _{LKG}	CC	D	Input leakage current	T _A = -40 °C	No current injection on adjacent pin	—	nA
						1	
						70	
						—	
						1	
						70	
		D		T _A = 25 °C		—	
						1	
						70	
						—	
						3	
						100	
		D		T _A = 85 °C		—	
						8	
						200	
						—	
						45	
						400	
		D		T _A = 105 °C		—	
						8	
						200	
						—	
						45	
						400	
		P		T _A = 125 °C		—	
						45	
						400	
						—	
						45	
						400	

Table 44. ADC_0 conversion characteristics (10-bit ADC_0)

Symbol		C	Parameter	Conditions ¹	Value			Unit
					Min	Typ	Max	
V _{SS_ADC0}	SR	—	Voltage on VSS_HV_ADC0 (ADC_0 reference) pin with respect to ground (V _{SS}) ²	—	−0.1	—	0.1	V
V _{DD_ADC0}	SR	—	Voltage on VDD_HV_ADC pin (ADC reference) with respect to ground (V _{SS})	—	V _{DD} − 0.1	—	V _{DD} + 0.1	V
V _{AINx}	SR	—	Analog input voltage ³	—	V _{SS_ADC0} − 0.1	—	V _{DD_ADC0} + 0.1	V
I _{ADC0pwd}	SR	—	ADC_0 consumption in power down mode	—	—	—	50	μA
I _{ADC0run}	SR	—	ADC_0 consumption in running mode	—	—	—	5	mA
f _{ADC0}	SR	—	ADC_0 analog frequency	—	6	—	32 + 4%	MHz
Δ _{ADC0_SYS}	SR	—	ADC_0 digital clock duty cycle (ipg_clk)	ADCLKSEL = 1 ⁴	45	—	55	%
t _{ADC0_PU}	SR	—	ADC_0 power up delay	—	—	—	1.5	μs
t _{ADC0_S}	CC	T	Sampling time ⁵	f _{ADC} = 32 MHz, INPSAMP = 17	0.5	—	42	μs
				f _{ADC} = 6 MHz, INPSAMP = 255	—	—		

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
t _{ADC0_C}	CC	P	Conversion time ⁶	f _{ADC} = 32 MHz, INPCMP = 2		0.625	—	—	μs
C _S	CC	D	ADC_0 input sampling capacitance	—		—	—	3	pF
C _{P1}	CC	D	ADC_0 input pin capacitance 1	—		—	—	3	pF
C _{P2}	CC	D	ADC_0 input pin capacitance 2	—		—	—	1	pF
C _{P3}	CC	D	ADC_0 input pin capacitance 3	—		—	—	1	pF
R _{SW1}	CC	D	Internal resistance of analog source	—		—	—	3	kΩ
R _{SW2}	CC	D	Internal resistance of analog source	—		—	—	2	kΩ
R _{AD}	CC	D	Internal resistance of analog source	—		—	—	2	kΩ
I _{INJ}	SR	—	Input current Injection	Current injection on one ADC_0 input, different from the converted one	V _{DD} = 3.3 V ± 10%	−5	—	5	mA
					V _{DD} = 5.0 V ± 10%	−5	—	5	
INL	CC	T	Absolute integral nonlinearity	No overload		—	0.5	1.5	LSB
DNL	CC	T	Absolute differential nonlinearity	No overload		—	0.5	1.0	LSB
E _O	CC	T	Absolute offset error	—		—	0.5	—	LSB
E _G	CC	T	Absolute gain error	—		—	0.6	—	LSB
TUEP	CC	P	Total unadjusted error ⁷ for precise channels, input only pins	Without current injection		−2	0.6	2	LSB
		With current injection		−3	—	3			
TUEX	CC	T	Total unadjusted error ⁷ for extended channel	Without current injection		−3	1	3	LSB
		With current injection		−4		4			

¹ $V_{\text{DD}} = 3.3 \text{ V} \pm 10\% / 5.0 \text{ V} \pm 10\%$, $T_A = -40$ to 125°C , unless otherwise specified.

² Analog and digital V_{SS} **must** be common (to be tied together externally).

³ V_{AINx} may exceed $V_{\text{SS_ADC0}}$ and $V_{\text{DD_ADC0}}$ limits, remaining on absolute maximum ratings, but the results of the conversion will be clamped respectively to 0x000 or 0x3FF.

⁴ Duty cycle is ensured by using system clock without prescaling. When $\text{ADCLKSEL} = 0$, the duty cycle is ensured by internal divider by 2.

⁵ During the sampling time the input capacitance C_S can be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach its final voltage level within $t_{\text{ADC0_S}}$. After the end of the sampling time $t_{\text{ADC0_S}}$, changes of the analog input voltage have no effect on the conversion result. Values for the sampling clock $t_{\text{ADC0_S}}$ depend on programming.

⁶ This parameter does not include the sampling time $t_{\text{ADC0_S}}$, but only the time for determining the digital result and the time to load the result's register with the conversion result.

⁷ Total Unadjusted Error: The maximum error that occurs without adjusting Offset and Gain errors. This error is a combination of Offset, Gain and Integral Linearity errors.

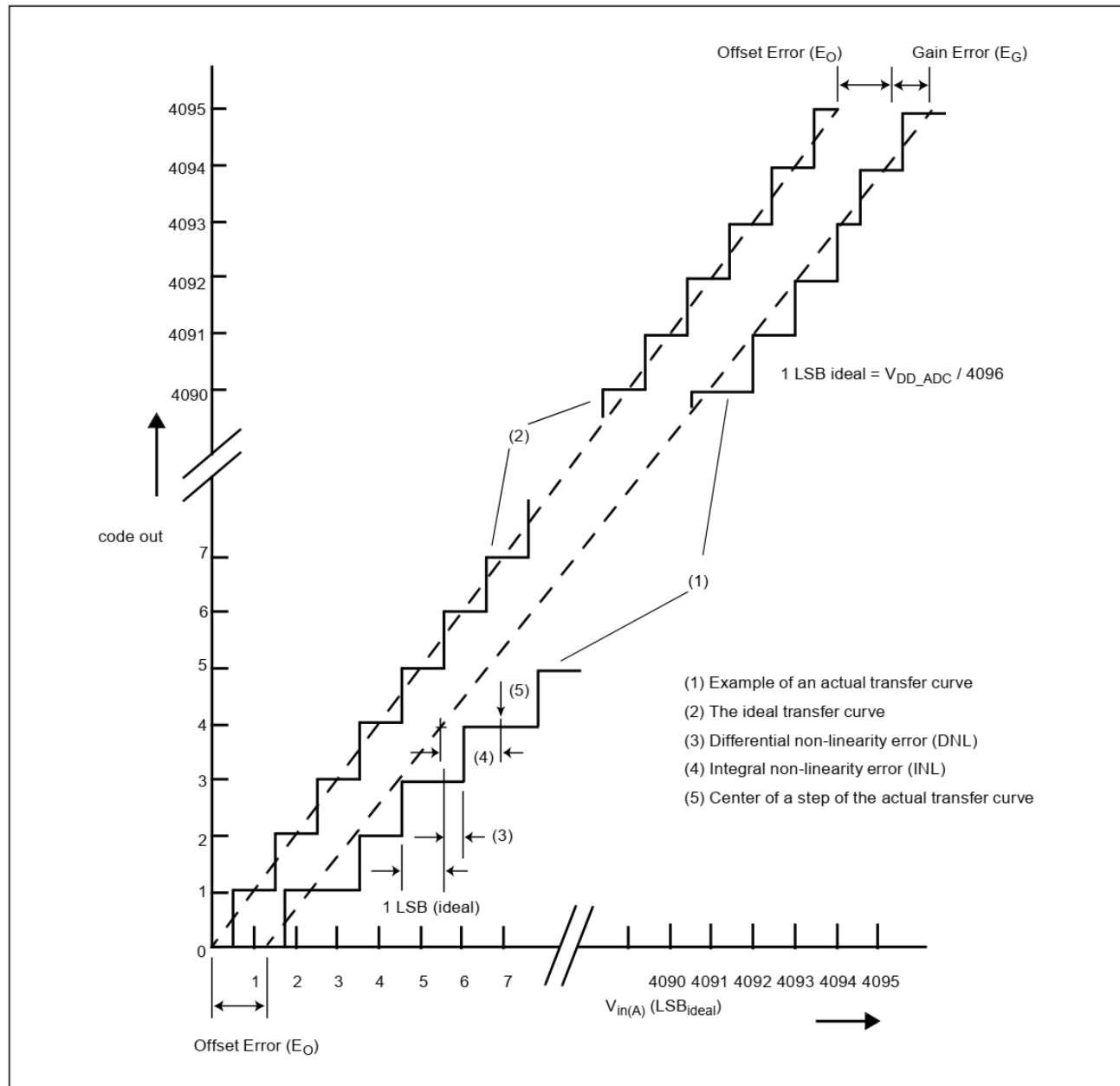


Figure 21. ADC_1 characteristic and error definitions
Table 45. ADC_1 conversion characteristics (12-bit ADC_1)

Symbol	C	Parameter	Conditions ¹	Value			Unit
				Min	Typ	Max	
V_{SS_ADC1}	SR	—	—	−0.1	—	0.1	V
V_{DD_ADC1}	SR	—	—	$V_{DD} - 0.1$	—	$V_{DD} + 0.1$	V

Table 45. ADC_1 conversion characteristics (12-bit ADC_1) (continued)

Symbol	C		Parameter	Conditions ¹	Value			Unit
					Min	Typ	Max	
V _{AINx}	SR	—	Analog input voltage ³	—	V _{SS_ADC1} - 0.1	—	V _{DD_ADC1} + 0.1	V
I _{ADC1pwd}	SR	—	ADC_1 consumption in power down mode	—	—	—	50	μA
I _{ADC1run}	SR	—	ADC_1 consumption in running mode	—	—	—	6	mA
f _{ADC1}	SR	—	ADC_1 analog frequency	V _{DD} = 3.3 V	3.33	—	20 + 4%	MHz
				V _{DD} = 5 V	3.33	—	32 + 4%	
t _{ADC1_PU}	SR	—	ADC_1 power up delay	—	—	—	1.5	μs
t _{ADC1_S}	CC	T	Sampling time ⁴ V _{DD} = 3.3 V	f _{ADC1} = 20 MHz, INPSAMP = 12	600	—	—	ns
			Sampling time ⁴ V _{DD} = 5.0 V	f _{ADC1} = 32 MHz, INPSAMP = 17	500	—	—	
			Sampling time ⁴ V _{DD} = 3.3 V	f _{ADC1} = 3.33 MHz, INPSAMP = 255	—	—	76.2	μs
			Sampling time ⁴ V _{DD} = 5.0 V	f _{ADC1} = 3.33 MHz, INPSAMP = 255	—	—	76.2	
t _{ADC1_C}	CC	P	Conversion time ⁵ V _{DD} = 3.3 V	f _{ADC1} = 20 MHz, INPCMP = 0	2.4	—	—	μs
			Conversion time ⁵ V _{DD} = 5.0 V	f _{ADC1} = 32 MHz, INPCMP = 0	1.5	—	—	μs
			Conversion time ⁵ V _{DD} = 3.3 V	f _{ADC1} = 13.33 MHz, INPCMP = 0	—	—	3.6	μs
			Conversion time ⁵ V _{DD} = 5.0 V	f _{ADC1} = 13.33 MHz, INPCMP = 0	—	—	3.6	μs
Δ _{ADC1_SYS}	SR	—	ADC_1 digital clock duty cycle	ADCLKSEL = 1 ⁶	45	—	55	%
C _S	CC	D	ADC_1 input sampling capacitance	—	—	—	5	pF
C _{P1}	CC	D	ADC_1 input pin capacitance 1	—	—	—	3	pF
C _{P2}	CC	D	ADC_1 input pin capacitance 2	—	—	—	1	pF
C _{P3}	CC	D	ADC_1 input pin capacitance 3	—	—	—	1.5	pF
R _{SW1}	CC	D	Internal resistance of analog source	—	—	—	1	kΩ
R _{SW2}	CC	D	Internal resistance of analog source	—	—	—	2	kΩ
R _{AD}	CC	D	Internal resistance of analog source	—	—	—	0.3	kΩ

Table 45. ADC_1 conversion characteristics (12-bit ADC_1) (continued)

Symbol		C	Parameter	Conditions ¹		Value			Unit
						Min	Typ	Max	
I_{INJ}	SR	—	Input current Injection	Current injection on one ADC_1 input, different from the converted one	$V_{DD} = 3.3\text{ V} \pm 10\%$	–5	—	5	mA
					$V_{DD} = 5.0\text{ V} \pm 10\%$	–5	—	5	
INLP	CC	T	Absolute integral nonlinearity – Precise channels	No overload		—	1	3	LSB
INLX	CC	T	Absolute integral nonlinearity – Extended channels	No overload		—	1.5	5	LSB
DNL	CC	T	Absolute differential nonlinearity	No overload		—	0.5	1	LSB
E_O	CC	T	Absolute offset error	—		—	2	—	LSB
E_G	CC	T	Absolute gain error	—		—	2	—	LSB
TUEP ⁷	CC	P	Total unadjusted error for precise channels, input only pins	Without current injection		–6	—	6	LSB
		T		With current injection		–8	—	8	
TUEX ⁷	CC	T	Total unadjusted error for extended channel	Without current injection		–10	—	10	LSB
		T		With current injection		–12	—	12	

¹ $V_{DD} = 3.3\text{ V} \pm 10\%$ / $5.0\text{ V} \pm 10\%$, $T_A = -40$ to $125\text{ }^{\circ}\text{C}$, unless otherwise specified

² Analog and digital V_{SS} **must** be common (to be tied together externally).

³ V_{AINx} may exceed V_{SS_ADC1} and V_{DD_ADC1} limits, remaining on absolute maximum ratings, but the results of the conversion will be clamped respectively to 0x000 or 0xFFF.

⁴ During the sampling time the input capacitance C_S can be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach its final voltage level within t_{ADC1_S} . After the end of the sampling time t_{ADC1_S} , changes of the analog input voltage have no effect on the conversion result. Values for the sampling clock t_{ADC1_S} depend on programming.

⁵ This parameter does not include the sampling time t_{ADC1_S} , but only the time for determining the digital result and the time to load the result's register with the conversion result.

⁶ Duty cycle is ensured by using system clock without prescaling. When $ADCLKSEL = 0$, the duty cycle is ensured by internal divider by 2.

⁷ Total Unadjusted Error: The maximum error that occurs without adjusting Offset and Gain errors. This error is a combination of Offset, Gain and Integral Linearity errors.

4.17 On-chip peripherals

4.17.1 Current consumption

Table 46. On-chip peripherals current consumption¹

Symbol	C	T	Parameter	Conditions		Typical value ²	Unit
I _{DD_BV} (CAN)	CC	T	CAN (FlexCAN) supply current on V _{DD_BV}	Bitrate: 500 Kbyte/s	Total (static + dynamic) consumption: • FlexCAN in loop-back mode • XTAL at 8 MHz used as CAN engine clock source • Message sending period is 580 μs	8 * f _{periph} + 85	μA
				Bitrate: 125 Kbyte/s		8 * f _{periph} + 27	
I _{DD_BV} (eMIOS)	CC	T	eMIOS supply current on V _{DD_BV}	Static consumption: • eMIOS channel OFF • Global prescaler enabled		29 * f _{periph}	μA
				Dynamic consumption: • It does not change varying the frequency (0.003 mA)		3	
I _{DD_BV} (SCI)	CC	T	SCI (LINFlex) supply current on V _{DD_BV}	Total (static + dynamic) consumption: • LIN mode • Baudrate: 20 Kbyte/s		5 * f _{periph} + 31	μA
I _{DD_BV} (SPI)	CC	T	SPI (DSPI) supply current on V _{DD_BV}	Ballast static consumption (only clocked)		1	μA
				Ballast dynamic consumption (continuous communication): • Baudrate: 2 Mbit/s • Transmission every 8 μs • Frame: 16 bits		16 * f _{periph}	
I _{DD_BV} (ADC_0/ADC_1)	CC	T	ADC_0/ADC_1 supply current on V _{DD_BV}	V _{DD} = 5.5 V	Ballast static consumption (no conversion) ³	41 * f _{periph}	μA
					Ballast dynamic consumption (continuous conversion) ³	46 * f _{periph}	
I _{DD_HV} _ADC0	CC	T	ADC_0 supply current on V _{DD_HV} _ADC0	V _{DD} = 5.5 V	Analog static consumption (no conversion)	200	μA
					Analog dynamic consumption (continuous conversion)	3	mA
I _{DD_HV} _ADC1	CC	T	ADC_1 supply current on V _{DD_HV} _ADC1	V _{DD} = 5.5 V	Analog static consumption (no conversion)	300 * f _{periph}	μA
					Analog dynamic consumption (continuous conversion)	4	mA
I _{DD_HV} (FLASH)	CC	T	CFlash + DFlash supply current on V _{DD_HV}	V _{DD} = 5.5 V	—	12	mA
I _{DD_HV} (PLL)	CC	T	PLL supply current on V _{DD_HV}	V _{DD} = 5.5 V	—	30 * f _{periph}	μA

- 1 Operating conditions: $T_A = 25^\circ\text{C}$, $f_{\text{periph}} = 8\text{ MHz to } 64\text{ MHz}$
 - 2 f_{periph} is an absolute value.
 - 3 During the conversion the total current consumption is given from the sum of the static and dynamic consumption
i.e. $(41 + 46) * f_{\text{periph}}$
-

4.17.2 DSPI characteristics

Table 47. DSPI characteristics¹

No.	Symbol	C		Parameter		DSPI0/DSPI1/DSPI3/DSPI5			DSPI2/DSPI4			Unit
						Min	Typ	Max	Min	Typ	Max	
1	t_{SCK}	SR	D	SCK cycle time	Master mode (MTFE = 0)	125	—	—	333	—	—	ns
			D		Slave mode (MTFE = 0)	125	—	—	333	—	—	
			D		Master mode (MTFE = 1)	83	—	—	125	—	—	
			D		Slave mode (MTFE = 1)	83	—	—	125	—	—	
—	f_{DSPI}	SR	D	DSPI digital controller frequency		—	—	f_{CPU}	—	—	f_{CPU}	MHz
—	Δt_{CSC}	CC	D	Internal delay between pad associated to SCK and pad associated to CSn in master mode for CSn1->0	Master mode	—	—	130^2	—	—	15^3	ns
—	Δt_{ASC}	CC	D	Internal delay between pad associated to SCK and pad associated to CSn in master mode for CSn1->1	Master mode	—	—	130^3	—	—	130^3	ns
2	$t_{CSCe_{xt}}^4$	SR	D	CS to SCK delay	Slave mode	32	—	—	32	—	—	ns
3	t_{ASCext}^5	SR	D	After SCK delay	Slave mode	$1/f_{DSPI} + 5$	—	—	$1/f_{DSPI} + 5$	—	—	ns
4	t_{SDC}	CC	D	SCK duty cycle	Master mode	—	$t_{SCK}/2$	—	—	$t_{SCK}/2$	—	ns
		SR	D		Slave mode	$t_{SCK}/2$	—	—	$t_{SCK}/2$	—	—	
5	t_A	SR	D	Slave access time	Slave mode	—	—	$1/f_{DSPI} + 70$	—	—	$1/f_{DSPI} + 130$	ns
6	t_{DI}	SR	D	Slave SOUT disable time	Slave mode	7	—	—	7	—	—	ns
7	t_{PCSC}	SR	D	PCSx to PCSS time	—	0	—	—	0	—	—	ns
8	t_{PASC}	SR	D	PCSS to PCSx time	—	0	—	—	0	—	—	ns
9	t_{SUI}	SR	D	Data setup time for inputs	Master mode	43	—	—	145	—	—	ns
					Slave mode	5	—	—	5	—	—	

Table 47. DSPI characteristics¹ (continued)

No.	Symbol		C	D	Parameter	DSPI0/DSPI1/DSPI3/DSPI5			DSPI2/DSPI4			Unit
						Min	Typ	Max	Min	Typ	Max	
10	t _{HI}	SR	D	Data hold time for inputs	Master mode	0	—	—	0	—	—	ns
					Slave mode	2 ⁶	—	—	2 ⁶	—	—	
11	t _{SUO} ⁷	CC	D	Data valid after SCK edge	Master mode	—	—	32	—	—	50	ns
					Slave mode	—	—	52	—	—	160	
12	t _{HO} ⁷	CC	D	Data hold time for outputs	Master mode	0	—	—	0	—	—	ns
					Slave mode	8	—	—	13	—	—	

¹ Operating conditions: $C_L = 10$ to 50 pF, $Slew_{IN} = 3.5$ to 15 ns

² Maximum value is reached when CSn pad is configured as SLOW pad while SCK pad is configured as MEDIUM. A positive value means that SCK starts before CSn is asserted. DSPI2 has only SLOW SCK available.

³ Maximum value is reached when CSn pad is configured as MEDIUM pad while SCK pad is configured as SLOW. A positive value means that CSn is deasserted before SCK. DSPI0 and DSPI1 have only MEDIUM SCK available.

⁴ The t_{CSC} delay value is configurable through a register. When configuring t_{CSC} (using PCSSCK and CSSCK fields in DSPI_CTARx registers), delay between internal CS and internal SCK must be higher than Δt_{CSC} to ensure positive t_{CSCext} .

⁵ The t_{ASC} delay value is configurable through a register. When configuring t_{ASC} (using PASC and ASC fields in DSPI_CTARx registers), delay between internal CS and internal SCK must be higher than Δt_{ASC} to ensure positive t_{ASCext} .

⁶ This delay value corresponds to SMPL_PT = 00b which is bit field 9 and 8 of DSPI_MCR register.

⁷ SCK and SOUT are configured as MEDIUM pad.

Figure 22. DSPI classic SPI timing — master, CPHA = 0

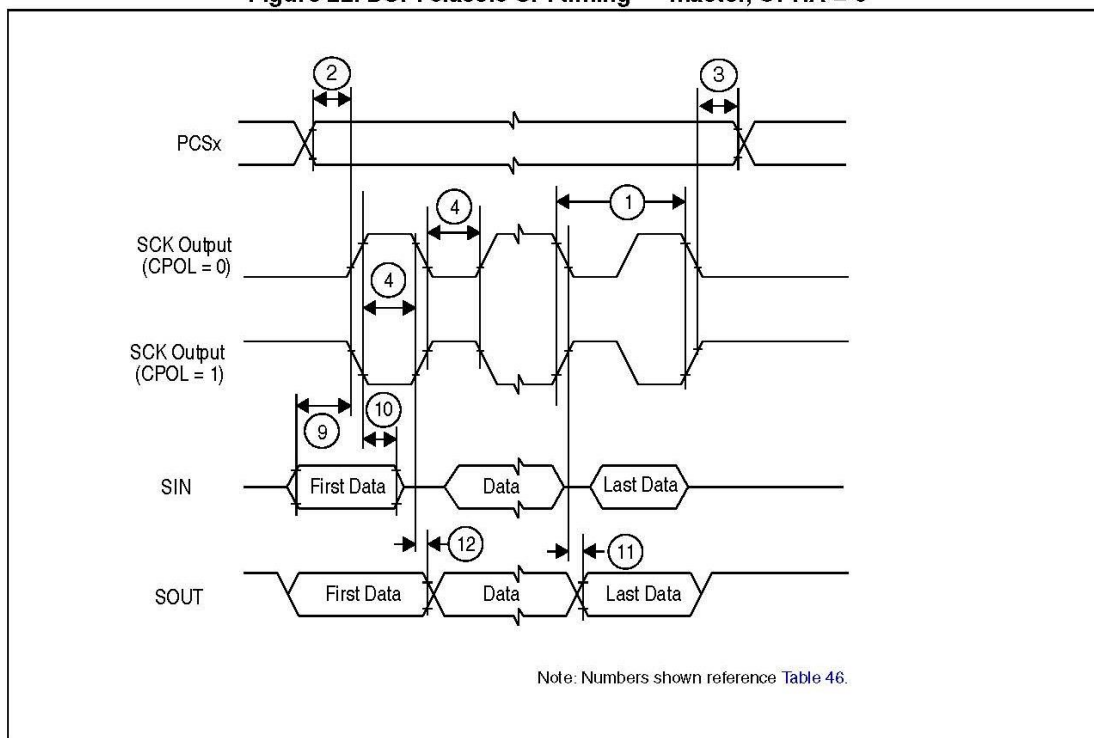


Figure 23. DSPI classic SPI timing — master, CPHA = 1

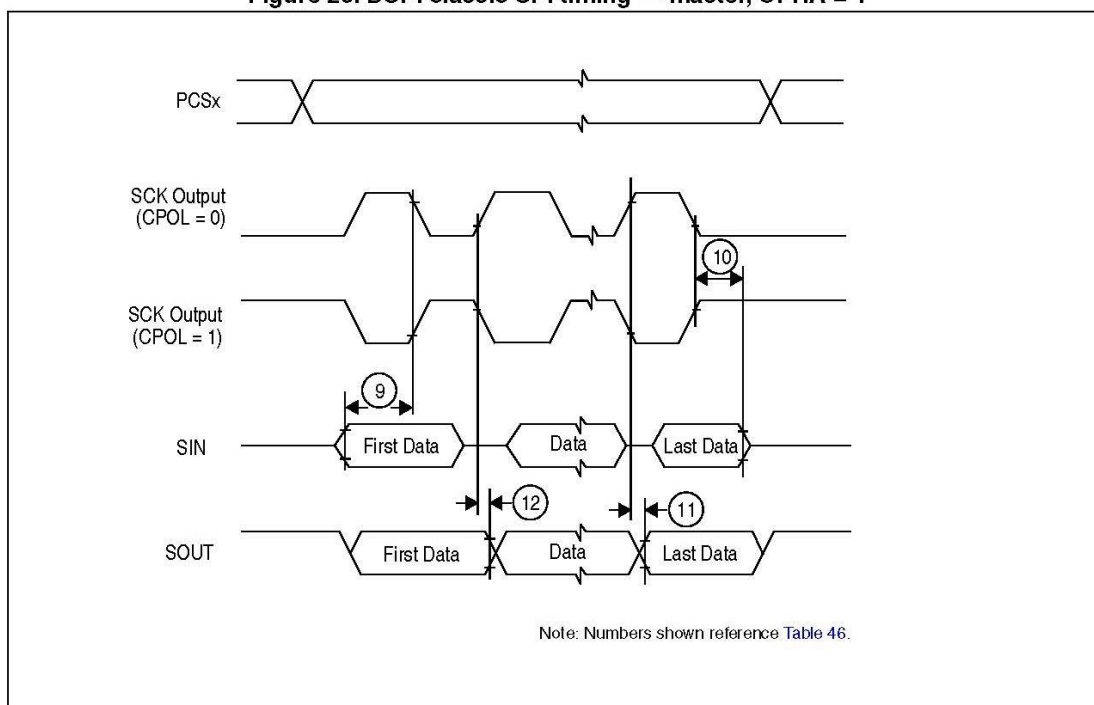


Figure 24. DSPI classic SPI timing — slave, CPHA = 0

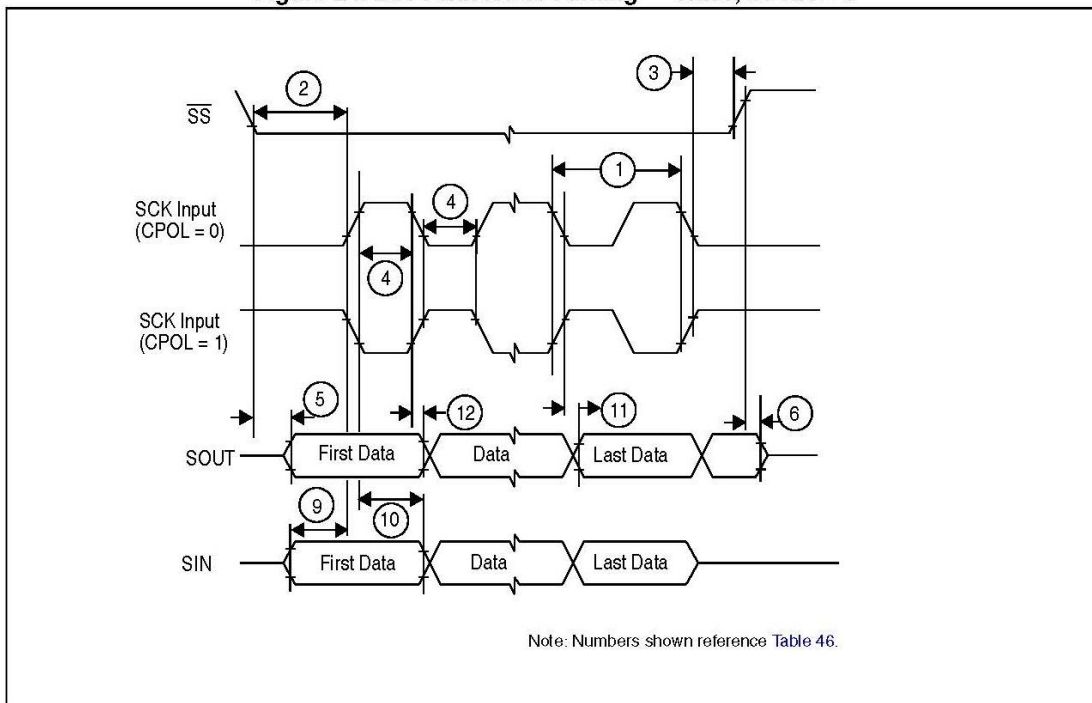


Figure 25. DSPI classic SPI timing — slave, CPHA = 1

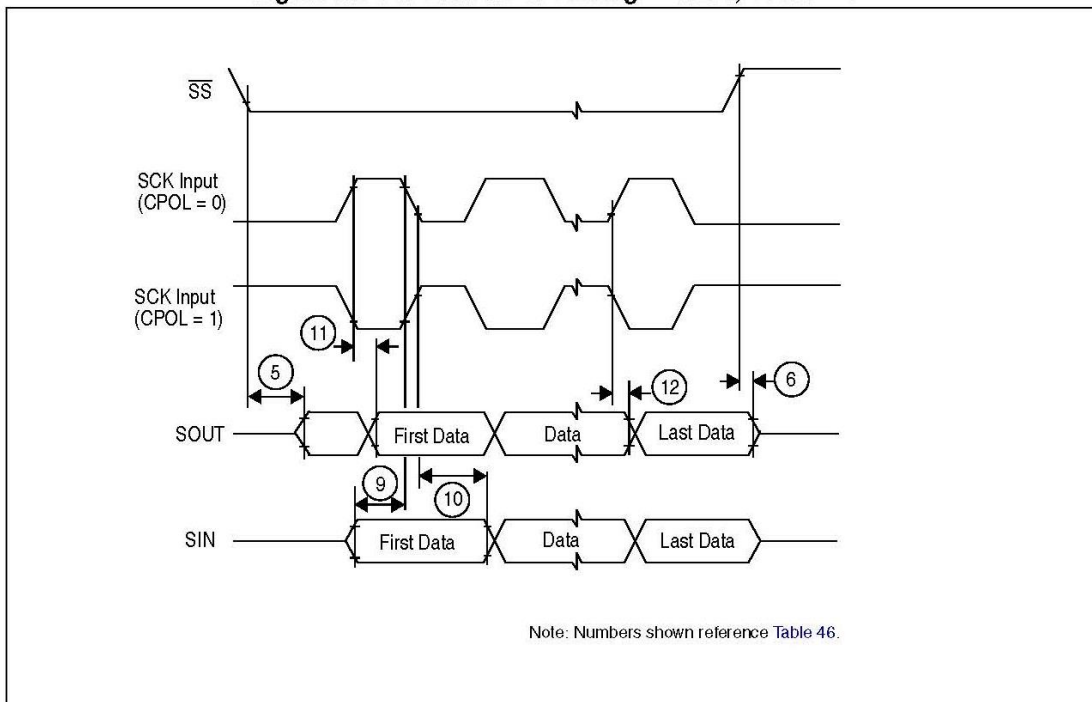


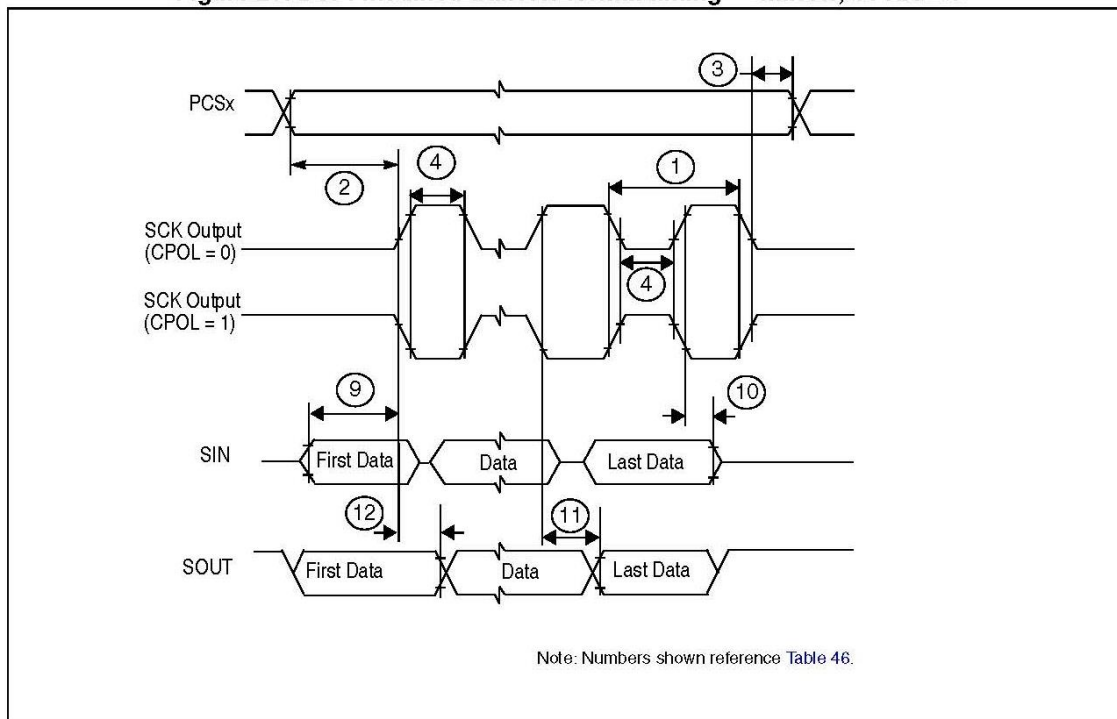
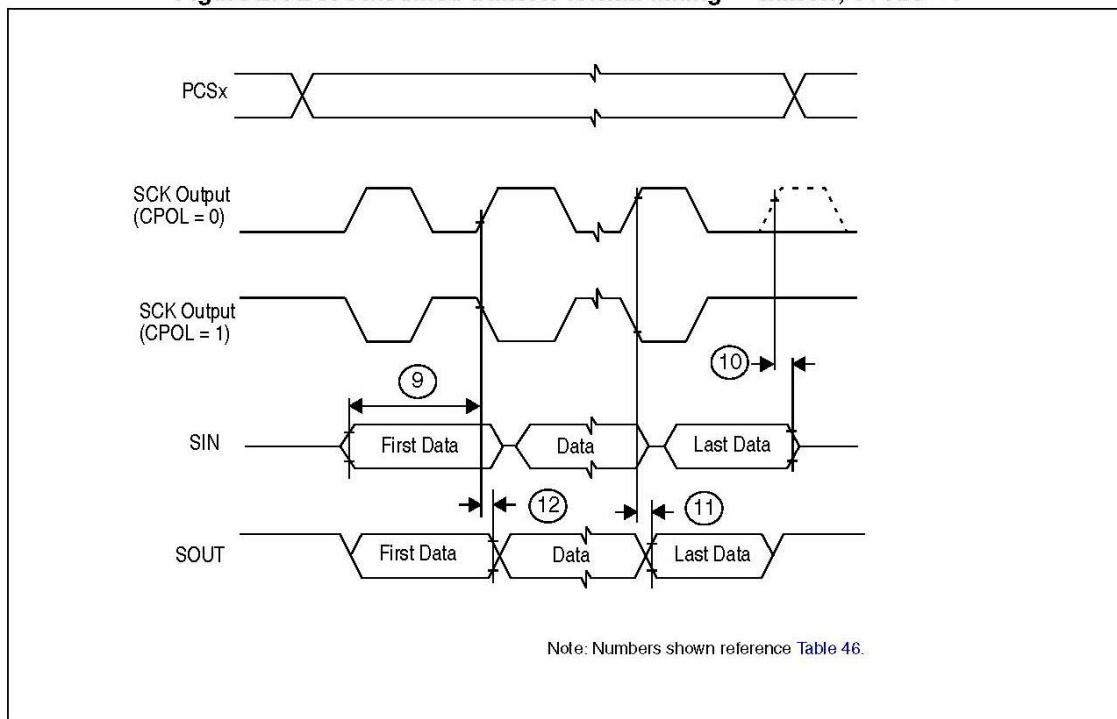
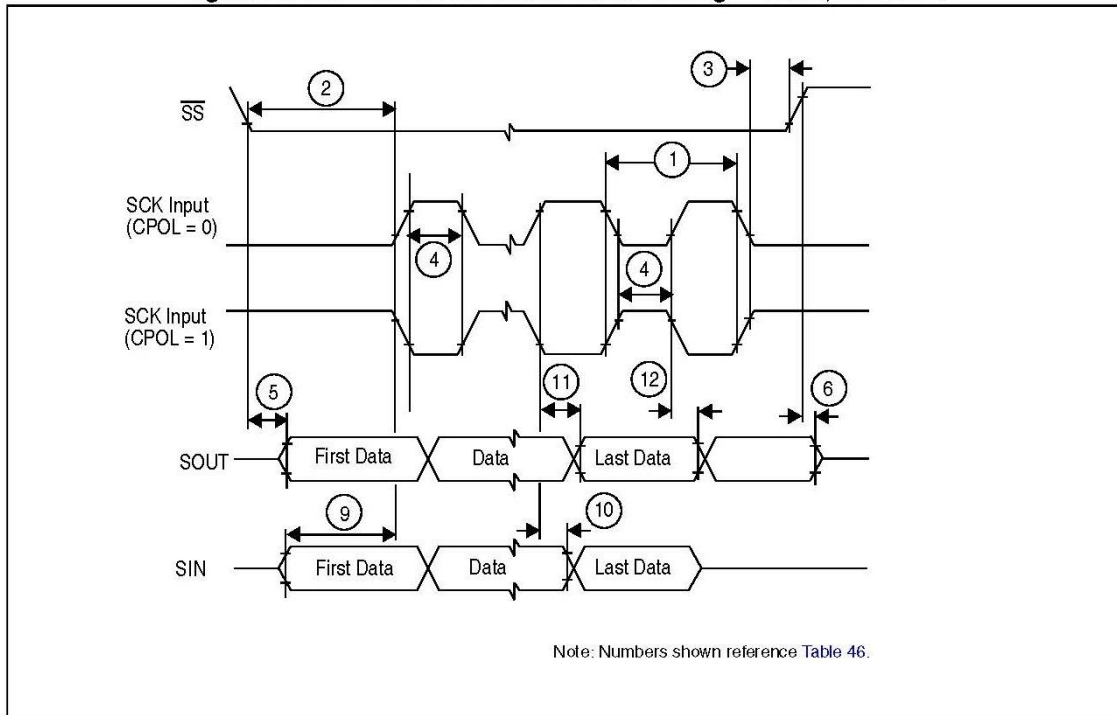
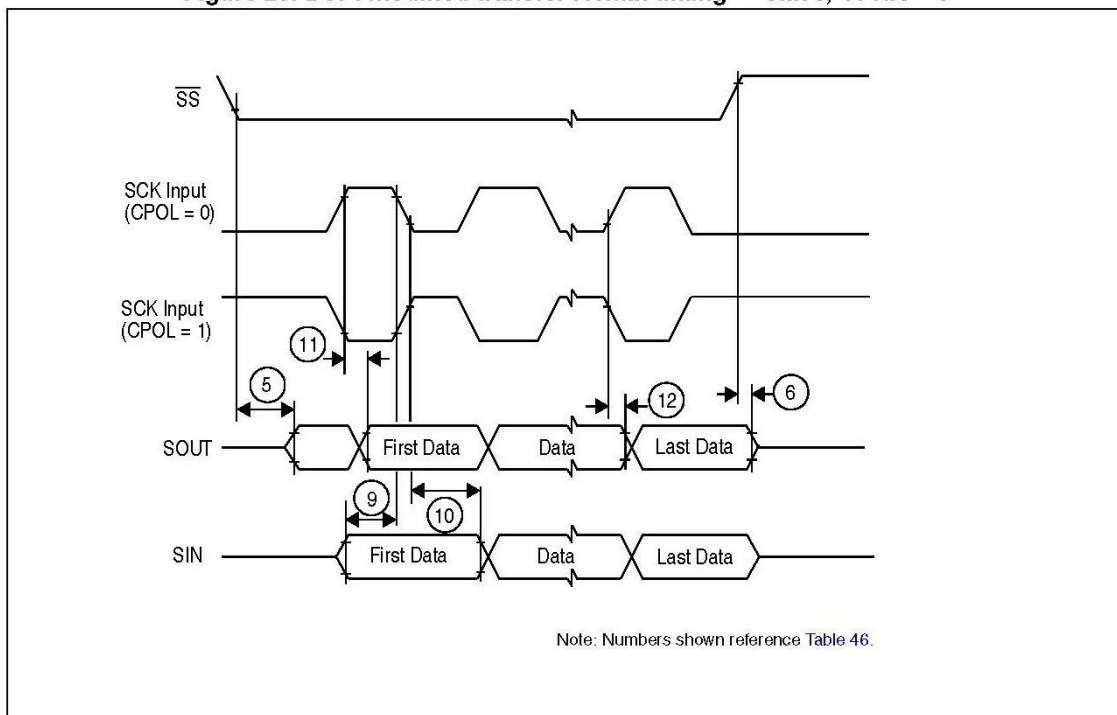
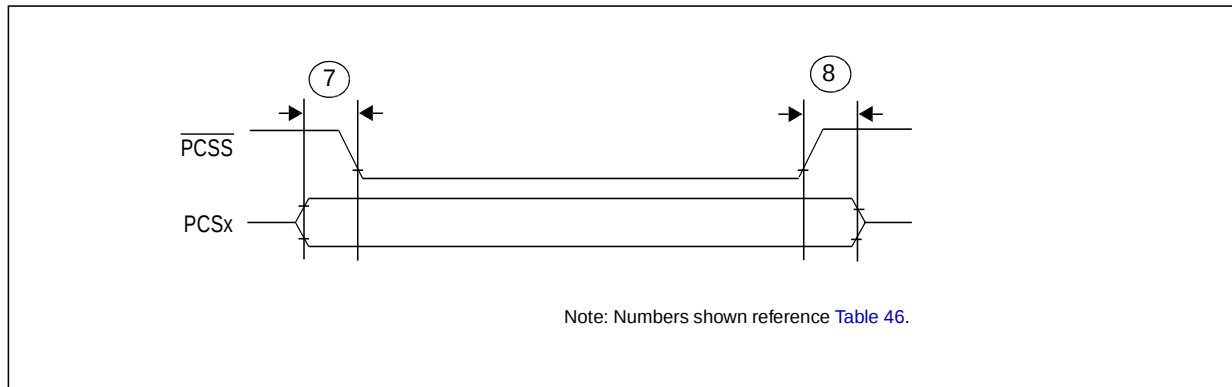
Figure 26. DSPI modified transfer format timing — master, CPHA = 0**Figure 27. DSPI modified transfer format timing — master, CPHA = 1**

Figure 28. DSPI modified transfer format timing — slave, CPHA = 0**Figure 29. DSPI modified transfer format timing — slave, CPHA = 1**

Figure 30. DSPI PCS strobe ($\overline{\text{PCSS}}$) timing

4.17.3 Nexus characteristics

Table 48. Nexus characteristics

No.	Symbol		C	Parameter	Value			Unit
					Min	Typ	Max	
1	t _{TCYC}	CC	D	TCK cycle time	64	—	—	ns
2	t _{MCYC}	CC	D	MCKO cycle time	32	—	—	ns
3	t _{MDOV}	CC	D	MCKO low to MDO data valid	—	—	8	ns
4	t _{MSEOV}	CC	D	MCKO low to MSEO_b data valid	—	—	8	ns
5	t _{EVTOV}	CC	D	MCKO low to EVTO data valid	—	—	8	ns
6	t _{NTDIS}	CC	D	TDI data setup time	15	—	—	ns
	t _{NTMSS}	CC	D	TMS data setup time	15	—	—	ns
7	t _{NTDIH}	CC	D	TDI data hold time	5	—	—	ns
	t _{NTMSH}	CC	D	TMS data hold time	5	—	—	ns
8	t _{TDOV}	CC	D	TCK low to TDO data valid	35	—	—	ns
9	t _{TDOI}	CC	D	TCK low to TDO data invalid	6	—	—	ns

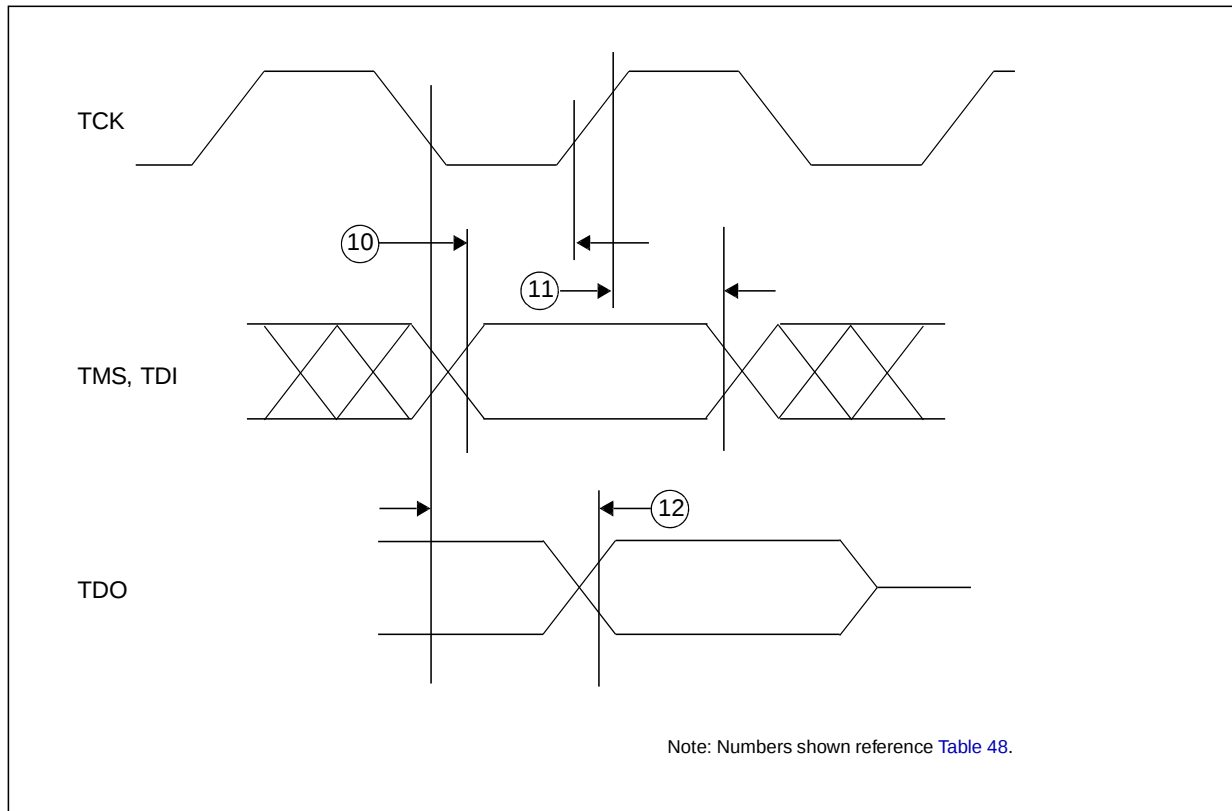


Figure 31. Nexus TDI, TMS, TDO timing

4.17.4 JTAG characteristics

Table 49. JTAG characteristics

No.	Symbol		C	Parameter	Value			Unit
					Min	Typ	Max	
1	t _{JCYC}	CC	D	TCK cycle time	64	—	—	ns
2	t _{TDIS}	CC	D	TDI setup time	15	—	—	ns
3	t _{TDIH}	CC	D	TDI hold time	5	—	—	ns
4	t _{TMSS}	CC	D	TMS setup time	15	—	—	ns
5	t _{TMSH}	CC	D	TMS hold time	5	—	—	ns
6	t _{TDOV}	CC	D	TCK low to TDO valid	—	—	33	ns
7	t _{TDOI}	CC	D	TCK low to TDO invalid	6	—	—	ns

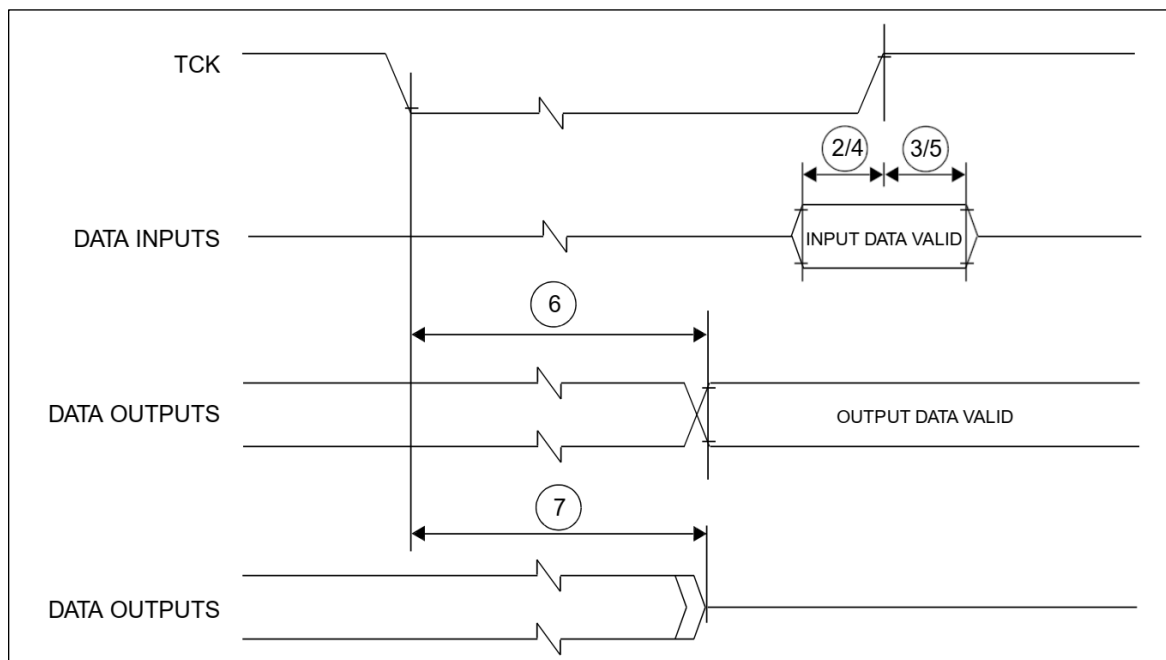


Figure 32. Timing diagram — JTAG boundary scan

5 Differences

This chapter show that points are incompatible with MPC560X.

The user should pay attention to:

5.1 CTU

The register can been changed between two different placements described in Figure 29-2.

The mapping of the channel number value to the corresponding ADC channel is provided in Table 29-4-1 or Table 29-4-2. The function can be changed to Table 29- 4-1 for CCFC2012BC60L7, CCFC2012BC60L5 and CCFC2012BC60L3 or Table 29- 4- 2 for CCFC2012BC50L5 and CCFC2012BC50L3.

5.2 SRAM

The size of SRAM in CCFC2012BC60L7 and CCFC2012BC60L5 is 128Kbyte. The size of SRAM in CCFC2012BC50L5 and CCFC2012BC60L3 is 96Kbyte.

5.3 INTC

Whether the interrupt sources described in Table 18-10 exist actually depends on the existence of the modules in each chip. All interrupt sources above 233 in CCFC2012BC do not exist on MPC560X.

The hard interrupt vector table's base address has to align with **0x0+0x1000*N**.

5.4 MC_CGM

In Modulation Register (CR) of FMPLL, the formula of INC_STEP is different between CCFC2012BC and MPC560X.

The INC_STEP field is the binary equivalent of the value incstep derived from following formula in CCFC2012BC:

$$\text{incstep} = \text{round}\left(\frac{(2^{10} - 1) \times MD \times EMFD}{100 \times 5 \times \text{modperiod}}\right)$$

The INC_STEP field is the binary equivalent of the value incstep derived from following formula in MPC560X:

$$\text{incstep} = \text{round}\left(\frac{(2^{15} - 1) \times MD \times EMFD}{100 \times 5 \times \text{modperiod}}\right)$$

The SELCTL in flowed table are different with MPC560X.

SELCTL	CCFC2012BC SERIES	MPC 560X
0000	8-20 MHz ext. xtal osc.	4-16 MHz ext. xtal osc.
0001	16 MHz int. RC osc	16 MHz int. RC osc
0010	freq. mod. PLL	freq. mod. PLL
0011	RTC clock	system clock
0100	128 KHz int. RC osc	RTC clock
0101	system clock	Reserved

5.5 FlexCAN

CCFC2012BCS has eight FlexCANs. MPC560X has six. In addition, the CANFD interface is be support by CCFC2012BC's CAN interface.

5.6 ADC

In ADC5607, INPCMP & INPSAMP in CTR one bit more than MPC560X.

INPLATCH & OFFSHIFT in CTR is unused.

ADC sampling and conversion timing is different from MPC560X.

FADC must be less than 16MHz.

5.7 PFU

The flash area where 0xFF data has been written to the flash cannot be written to other values. The area can be written only after the flash is erased.

CFLASH size is 1.5MB in CCFC2012BC60L7 and CCFC2012BC60L5.

CFLASH size is 1MB in CCFC201250L5, CCFC2012BC50L3 and CCFC2012BC50L1.

CFLASH size is 1.5MB in MPC560X.

In CCFC2012BC, PFU registers are different with MPC560X.

Registers	CCFC2012BC SERIES	MPC560X
CFLASH_UT0(User test register)	Not exist	Exists
CFLASH_UT1	Not exist	Exists
CFLASH_UT2	Not exist	Exists
CFLASH_UMISR0(User multiple input signature register)	Not exist	Exists
CFLASH_UMISR1	Not exist	Exists
CFLASH_UMISR2	Not exist	Exists
CFLASH_UMISR3	Not exist	Exists
CFLASH_UMISR4	Not exist	Exists
NVPWD0(Nonvolatile private censorship password register)	Not exist	Exists
NVPWD1	Not exist	Exists
NVSCC0(Nonvolatile system censorship control register)	Not exist	Exists
NVSCC0	Not exist	Exists
DLASH_UT0(User test register)	Not exist	Exists
DLASH_UT1	Not exist	Exists
DLASH_UT2	Not exist	Exists
DLASH_UMISR0(User multiple input signature register)	Not exist	Exists
DLASH_UMISR1	Not exist	Exists
DLASH_UMISR2	Not exist	Exists
DLASH_UMISR3	Not exist	Exists
DLASH_UMISR4	Not exist	Exists

In CCFC2012BC, flash read access timing are different with MPC560X.

Frequency	CCFC2012BC SERIES	MPC560X
120 Mhz	4 wait states	-
80 Mhz	3 wait states	-
64 Mhz	2 wait states	2 wait states
40 Mhz	2 wait states	1 wait state
20 Mhz	2 wait states	0 wait states

5.8 XBAR

Misaligned access is not support in CCFC2012BC. Half word access, the address value must be a multiple of 2. As to word access, the address value must be a multiple of 4.

5.9 MC_ME

Wake up from standby mode, the flash mode of me_drun_mc register(DFLAON&CFLAON) must be configured to 11(normal mode) .

5.10 CLOCK

External crystal oscillator does not need external 1 M Ω resistance.

5.11 DSPI

During DSPI continuous operation, only 10 and 01 modes are supported temporarily, and 00 and 01 mode are not supported in the current version.

5.12 WKUP

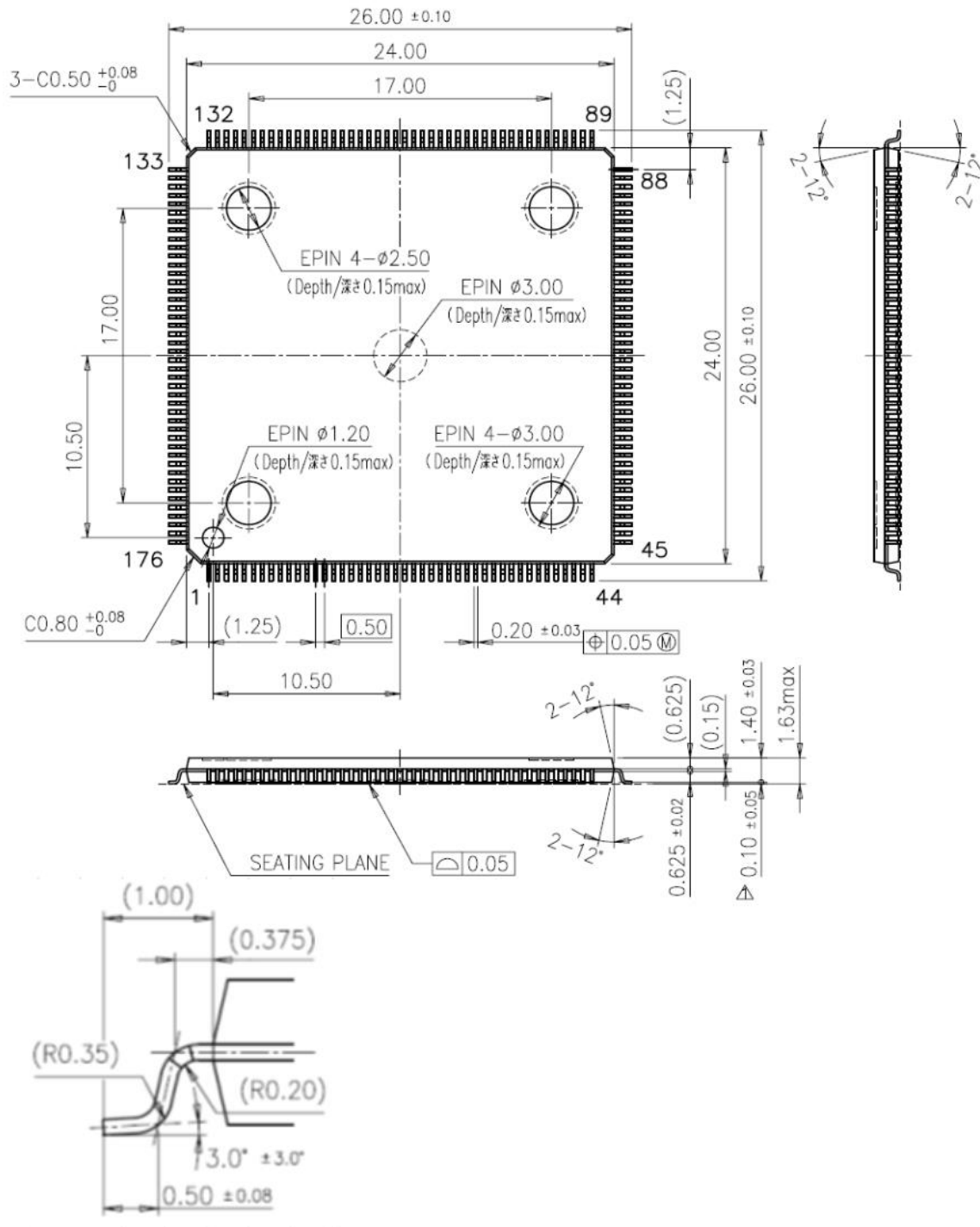
When initializing the configuration operation, all wakeup pin in reference manual *Table 12-1* need to be configured by pulling up or pulling down, otherwise it will cause electric leakage.

5.13 DMA

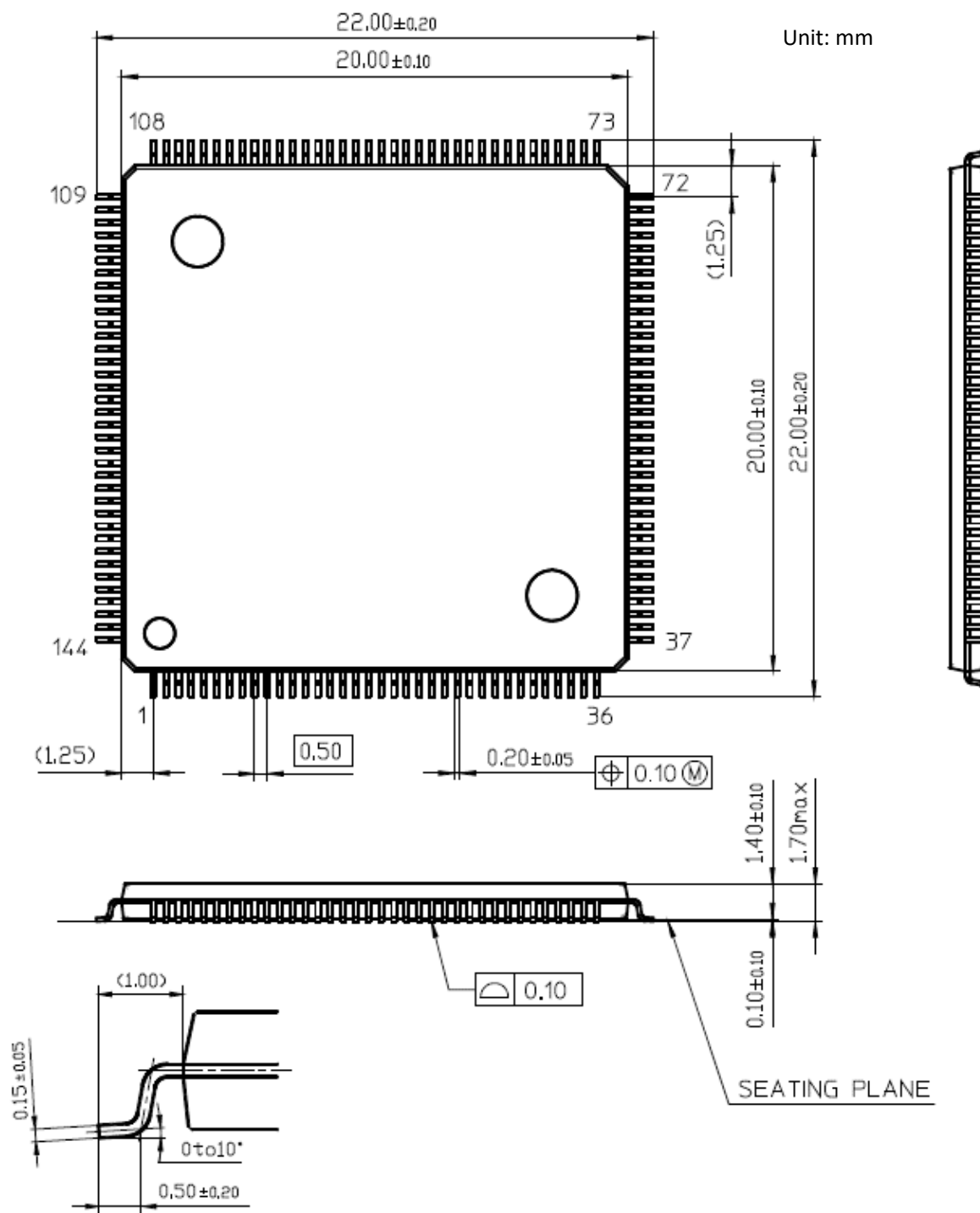
The DMA function is supported by the full range of CCFC2012BC.

6 Package characteristics

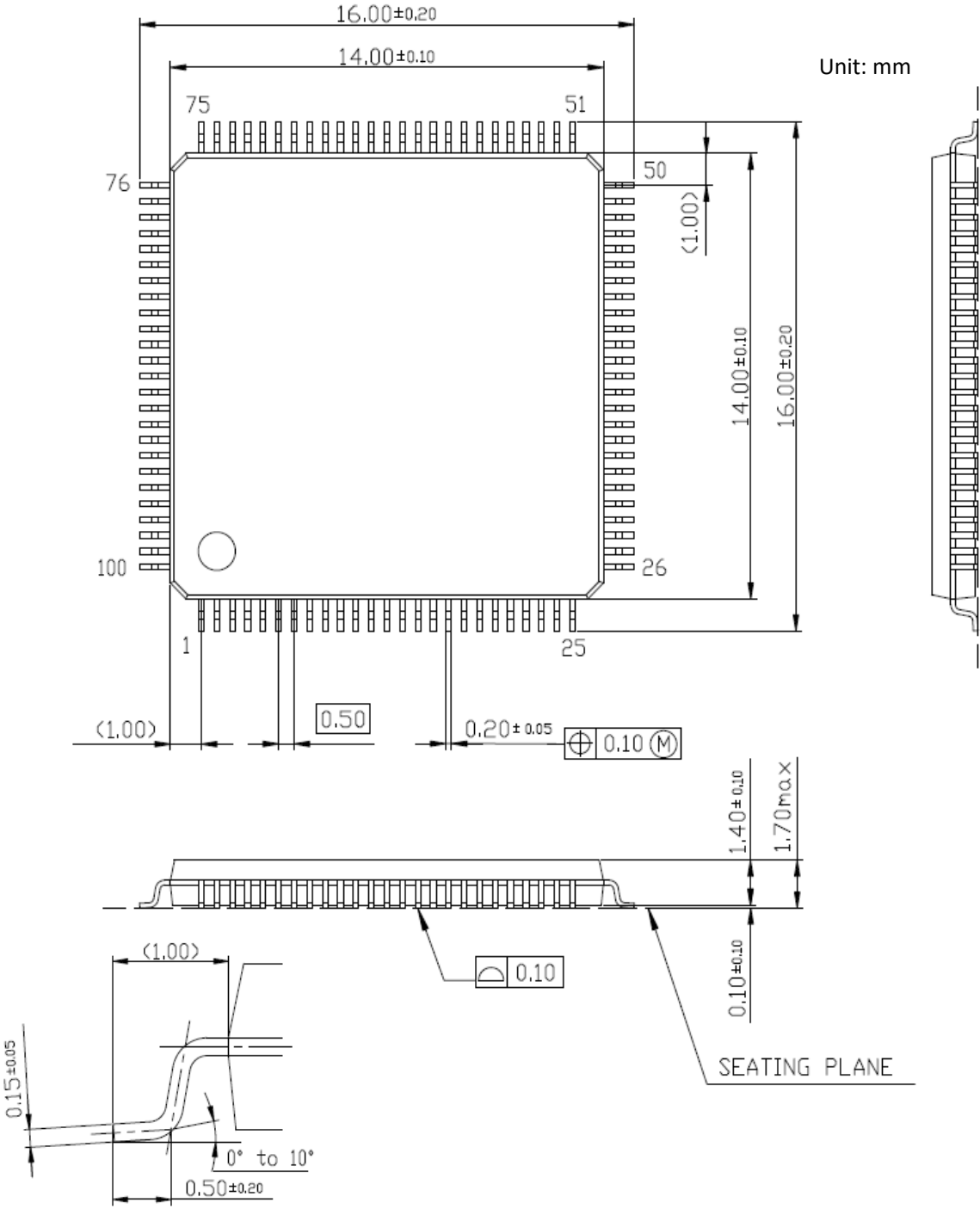
6.1 176 LQFP



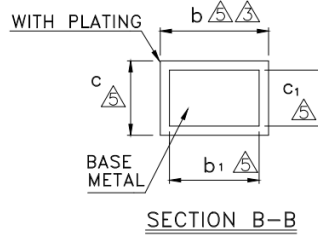
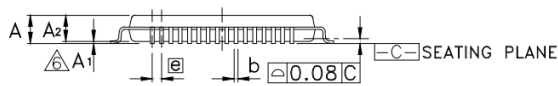
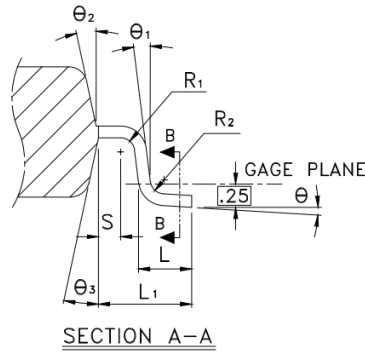
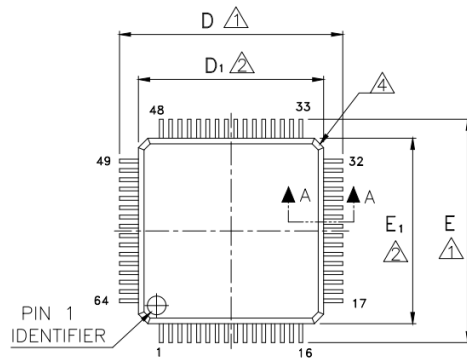
6.2 144 LQFP



6.3 100 LQFP



6.4 64 LQFP



NOTE :

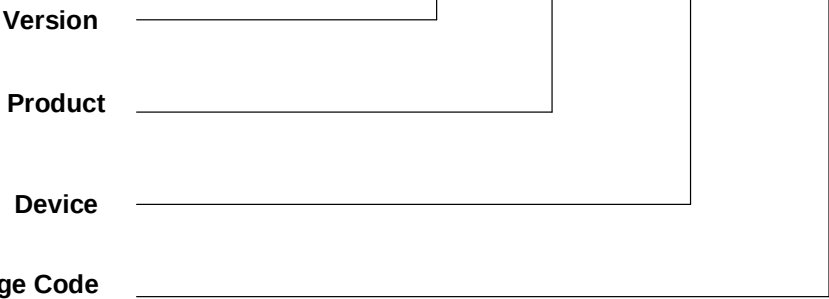
- △ TO BE DETERMINED AT SEATING PLANE $\overline{C-C}$.
- △ DIMENSIONS D1 AND E1 DO NOT INCLUDE MOLD PROTRUSION. D1 AND E1 ARE MAXIMUM PLASTIC BODY SIZE DIMENSIONS INCLUDING MOLD MISMATCH.
- △ DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR CAN NOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT.
- △ EXACT SHAPE OF EACH CORNER IS OPTIONAL.
- △ THESE DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN 0.10 mm AND 0.25 mm FROM THE LEAD TIP.
- △ A1 IS DEFINED AS THE DISTANCE FROM THE SEATING PLANE TO THE LOWEST POINT OF THE PACKAGE BODY.
- 7. CONTROLLING DIMENSION : MILLIMETER.
- 8. REFERENCE DOCUMENT : JEDEC MS-026 , BCD.

Symbol	Dimension in mm			Dimension in inch		
	Min	Norm	Max	Min	Norm	Max
A	—	—	1.60	—	—	0.063
A ₁	0.025	—	0.127	0.001	—	0.005
A ₂	1.35	1.40	1.45	0.053	0.055	0.057
b	0.17	0.22	0.27	0.007	0.009	0.011
b ₁	0.17	0.20	0.23	0.007	0.008	0.009
c	0.09	—	0.20	0.004	—	0.008
c ₁	0.09	—	0.16	0.004	—	0.006
D	12.00 BSC			0.472 BSC		
D ₁	10.00 BSC			0.394 BSC		
E	12.00 BSC			0.472 BSC		
E ₁	10.00 BSC			0.394 BSC		
⌀	0.50 BSC			0.020 BSC		
L	0.45	0.60	0.75	0.018	0.024	0.030
L ₁	1.00 REF			0.039 REF		
R ₁	0.08	—	—	0.003	—	—
R ₂	0.08	—	0.20	0.003	—	0.008
S	0.20	—	—	0.008	—	—
Θ	0°	3.5°	7°	0°	3.5°	7°
Θ ₁	0°	—	—	0°	—	—
Θ ₂	12°TYP			12°TYP		
Θ ₃	12°TYP			12°TYP		

7 Ordering information

Excample Code:

CCFC2012 BC 60 L7



- | | |
|-------------------------------------|--|
| Version
CCFC2012 | Device
60 = 5607
50 = 5604 |
| Product
BC = Body Control | Package Code
L1 = 64 LQFP
L3 = 100 LQFP
L5 = 144 LQFP
L7 = 176 LQFP |

8 Revision history

Table 54 summarizes revisions to this document.

Table 54. Revision history

Revision	Date	Substantive changes
1.0	18-Feb-2022	Initial release(Electrical characteristics shall be subject to the actual measurement)
2.0	6-May-2022	Some points incompatible with the target device are added
2.1	20-May-2022	Add DMA feature supported by the full range products
2.2	24-May-2022	Update ADC Chapter
2.3	10-Jul-2022	Update the i2c source
2.4	25-Jul-2022	Update the V_{MREG} value
3.0	29-Jan-2024	Modify external crystal oscillator frequency range. Add CCFC2012BC60L1 series in Table 1-1

Appendix A Abbreviations

Table 53 lists abbreviations used but not defined elsewhere in this document.

Table 53.
Abbreviations

Abbreviation	Meaning
CMOS	Complementary metal oxide semiconductor
CPHA	Clock phase
CPOL	Clock polarity
CS	Peripheral chip select
EVTO	Event out
MCKO	Message clock out
MDO	Message data out
MSEO	Message start/end out
MTFE	Modified timing format enable
SCK	Serial communications clock
SOUT	Serial data out
TBD	To be defined
TCK	Test clock input
TDI	Test data input
TDO	Test data output
TMS	Test mode select